

Electromagnetic Characterization Of The Basal Plane Region In Sea Ice

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1 Introduction

In electromagnetic pulse measurements of the depth of sea ice, it is found that under certain circumstances the strength of the return from the lower surface of the ice is strongly polarization dependent [1]. This occurs when there are well-developed basal planes, i.e. planes containing brine drainage channels extending approximately 5–20 cm up from the lower surface. The planes are closely spaced and parallel to one another, being perpendicular to a common c axis. Under these circumstances, the lower surface return is a maximum when the electric vector is parallel to the c axis, and a minimum when it is perpendicular to c . In the latter case it is not unusual to lose the lower surface return completely.

In an effort to understand this phenomenon, various models have been proposed, such as equivalent media theories, parallel metallic plates, etc. These have not been successful and, in some cases, have indicated results completely opposite to those observed. Thus, the task at hand is to develop a physically-based model that explains the observed effects and shows the dependence on channel length and spacing, brine concentration, wavelength and polarization. In this report we will present such a model based on the electrical properties of the brine drainage channels and the periodic nature of the basal planes with respect to the c axis. The solution to the scattered field will be determined for plane wave incidence using the appropriate electric field integral equation whose kernel is the Green's function for a periodic infinite array. Upon solving the integral equation numerically by the method of moments, results are obtained for the reflected field which support the observations made in [1].

2 The Model

In the formation of the basal planes, the brine concentration varies from virtually zero at the top of each plane to very high at the lower surface in

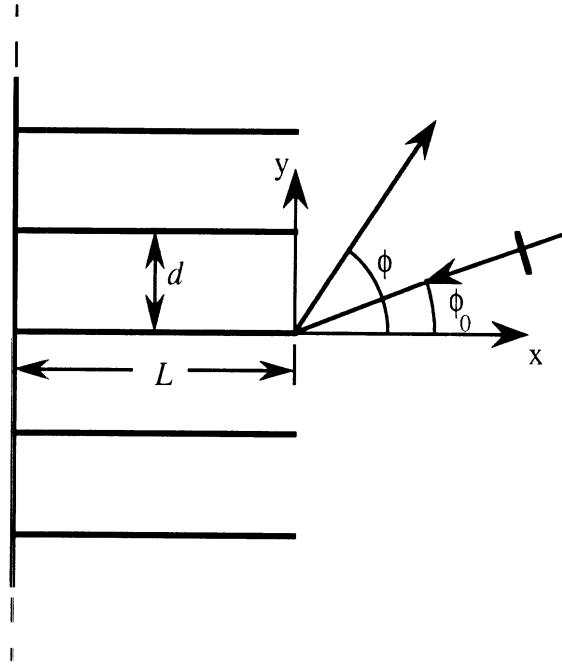


Figure 1: Infinite array of parallel resistive sheets with a perfectly conducting termination.

contact with the underlying sea [1]. It is convenient to think of the brine channels as concentrated in a layer of very small electrical thickness which can be modeled as a resistive sheet. The resistivity of each sheet therefore decreases from high at the top to zero at the bottom. An appropriate model for the basal plane region is then an infinite array of parallel resistive sheets terminated in a perfect conductor simulating the sea surface. When the resistive sheets are equally-spaced and have identical lengths and resistivity profiles, the array becomes a periodic structure, and the electric field integral equation can be readily solved numerically by employing the Green's function for a periodic infinite array.

The model of the basal plane region is shown in Fig. 1 as an infinite array of parallel resistive sheets of length L and separation d terminated in a perfect conductor with the y axis parallel to the c axis of the sea ice. The sheets have identical resistivities tapered quadratically from a maximum at the front ($x = 0$) to zero at the back ($x = -L$). The resistivity in ohms per square of the sheet at, say, $-L \leq x \leq 0$, $y = 0$ is given by

$$R(x, 0) = R_0 \left(\frac{x + L}{L} \right)^2 \quad (1)$$

where R_0 is the specified maximum value of the resistivity at $x = 0$, and from the periodicity of the array in y ,

$$R(x, pd) = R(x, 0), \quad p = 0, \pm 1, \pm 2, \dots$$

The boundary (transition) condition of the total field across each of the sheets is

$$\hat{y} \times \hat{y} \times \bar{E}(x, pd) = -R(x, pd) \bar{J}(x, pd), \quad -L < x \leq 0, \quad (2)$$

where $p = 0, \pm 1, \pm 2, \dots$ and \bar{J} is the total surface current on the sheets. For the perfect conductor in the $x = -L$ plane, the resistivity is zero, and the boundary condition for the total electric field is

$$\hat{x} \times \bar{E}(-L, y) = 0. \quad (3)$$

3 Electromagnetic Field Analysis

Consider the plane wave

$$E_z^i(H_z^i) = e^{-ik(x \cos \phi_0 + y \sin \phi_0)} \quad (4)$$

incident on the infinite array for E -(H -) polarization, where k is the propagation constant of the medium (assumed lossless) and a time factor $e^{-i\omega t}$ has been suppressed. The induced current on the resistive sheets and termination must be periodic in accordance with Floquet's theorem, such that

$$\bar{J}(x, pd) = \bar{J}(x, 0) e^{-ipkd \sin \phi_0}, \quad -L < x \leq 0, \quad (5)$$

and

$$\bar{J}(-L, y + pd) = \bar{J}(-L, y) e^{-ipkd \sin \phi_0} \quad (6)$$

with $p = 0, \pm 1, \pm 2, \dots$. In light of (5) and (6), the scattered fields can be expressed in an integral form where the integration is only taken over one of the cells of the infinite array, say $-L \leq x \leq 0$, $0 \leq y < d$. For E -polarization, the scattered electric field is

$$E_z^s(x, y) = ikZ \int_{l'} J_z(x', y') G(x, y, x', y') dl', \quad (7)$$

and for H -polarization, the scattered magnetic field is

$$H_z^s(x, y) = -\hat{z} \cdot \int_{l'} J_l(x', y') \hat{l} \times \nabla G(x, y, x', y') dl' \quad (8)$$

where Z is the intrinsic impedance of the medium and the path l' of the line integrals is over $-L \leq x \leq 0$, $y = 0$ and $0 \leq y < d$, $x = -L$. The unit vector \hat{l} is tangent to the path of integration l' , and the subscript l denotes the

component in the direction \hat{l} . The Green's function for the periodic infinite array is

$$G = \frac{i}{4} \sum_{p=-\infty}^{\infty} H_0^{(1)} \left(k \sqrt{(x-x')^2 + (y-y'-pd)^2} \right) e^{-ipkd \sin \phi_0} \quad (9)$$

where $H_0^{(1)}$ is the zeroth order Hankel function of the first kind. Since the series is slowly convergent, especially for small d , a more quickly converging form ($d \lesssim 2\lambda$) was derived in [2], giving

$$G = \frac{i}{2d} \sum_{p=-\infty}^{\infty} \frac{1}{k_x} e^{i(k_x|x-x'|+k_y(y-y'))} \quad (10)$$

with

$$\begin{aligned} k_x &= \sqrt{k^2 - k_y^2} \\ k_y &= k(p\lambda/d - \sin \phi_0). \end{aligned}$$

To solve for the unknown surface currents in (7) and (8), an integral equation for the total electric field $\bar{E} = \bar{E}^i + \bar{E}^s$ is derived by applying the boundary conditions (2) and (3). For E -polarization, we obtain

$$\begin{aligned} E_z^i(x, y) &= R(x, y) J_z(x, y) \\ &\quad - ikZ \int_{l'} J_z(x', y') G(x, y, x', y') dl', \end{aligned} \quad (11)$$

and for H -polarization,

$$\begin{aligned} E_l^i(x, y) &= R(x, y) J_l(x, y) \\ &\quad - ikZ \int_{l'} J_l(x', y') \hat{l} \cdot \left(\bar{I} + \frac{1}{k^2} \nabla \nabla \right) G(x, y, x', y') dl' \end{aligned} \quad (12)$$

where (x, y) are taken on l' and, in (12), $E_l^i = (iZ/k) \hat{l} \cdot \nabla \times (\hat{z} H_z^i)$.

Upon solution of (11) and (12), the scattered fields are given by (7) and (8), respectively, and we are interested in the region $x > 0$. Inspection of the Green's function (10) reveals that the scattered field is comprised of reflected waves propagating away from the structure for k_x real ($k_y < \pm k$) and surface waves decaying exponentially away from the structure for k_x imaginary ($k_y > \pm k$). Thus the condition on the index p for a reflected wave to occur is

$$-\frac{d}{\lambda}(1 - \sin \phi_0) < p < \frac{d}{\lambda}(1 + \sin \phi_0), \quad (13)$$

and the p^{th} wave propagates in the direction

$$\phi_p^r = \sin^{-1}(p\lambda/d - \sin \phi_0), \quad (14)$$

taking the principle value of the arcsine measured from the positive x axis. In the far field only the reflected waves are observable, and for E -polarization, the p^{th} reflected wave is

$$E_{zp}^r = -\frac{kZ}{2k_x d} e^{i(k_x x + k_y y)} \int_{l'} J_z(x', y') e^{-i(k_x x' + k_y y')} dl', \quad (15)$$

and likewise for H -polarization,

$$H_{zp}^r = \frac{1}{2d} e^{i(k_x x + k_y y)} \cdot \int_{l'} \hat{l} \cdot \left[\hat{x} \frac{k_y}{k_x} J_x(x', y') - \hat{y} J_y(x', y') \right] e^{-i(k_x x' + k_y y')} dl' \quad (16)$$

where k_x and k_y are evaluated at the p^{th} value of the infinite series.

4 Numerical Implementation

The integral equations (11) and (12) were programmed for numerical solution by the method of moments to determine the unknown surface currents on one cell of the infinite array. The surface currents were discretized by employing a pulse basis function expansion, and the remaining integral of the Green's function (10) over each segment was then evaluated analytically. Applying point matching over l' yielded a set of linear equations for solution, the unknown coefficients being the surface currents on each segment of the expansion. For E -polarization an appropriate upper limit on the infinite series in (10) was found to be $p = p_{max} \simeq 60$ for $d \lesssim 2$, with the convergence improving for decreasing d . For H -polarization the convergence was slowed due to the derivatives of the kernel (see (12)), and an auxiliary series was introduced following the development in [2] to improve the convergence rate such that $p_{max} \simeq 60$, $d \lesssim 2$.

The numerical implementation of the method of moments solution is contained in the code INFARR as listed in the Appendix. The user is prompted to choose either E - or H -polarization and to enter the length L , separation d , maximum resistivity R_0 , and the order of taper of the resistive sheets, along with the upper limit p_{max} and angle of incidence ϕ_0 . Other options include removing the perfectly conducting termination from the solution, specifying a constant profile for $R(x, 0)$ rather than a tapered one, and generating results over a range of L (all other parameters constant). We note that since a singularity exists in the Green's function series if $k_y = \pm k$, a restriction on d and ϕ_0 is that $p\lambda/d - \sin \phi_0 \neq \pm 1$ for any value of p . The user is notified if such a case exists and requested to change either parameter. The sampling rate used is 100 segments per wavelength, and this is increased by

$10(1 - f)$, $f = d, L$ for $d, L < \lambda$ to ensure a proper sampling. The output then is the magnitude and phase of the specular reflected field given by (15) or (16) with $p = 0$, whose direction of propagation is $\phi_0^r = -\phi_0$. The user is also informed of which other reflected waves exist, if any, according to (13) and their direction of propagation (14). The magnitude and phase of these fields could be output as well by a minor change in the code.

5 Results and Discussion

For the results presented here, the case of primary interest was when $d < 1\lambda$ and $\phi_0 = 0$. Consequently, the condition (13) gives $p = 0$ as the only reflected wave. The quadratic resistivity profile in (1) will be assumed, and p_{max} is set to 50.

5.1 *E*-polarization

The reflected field E_{z0}^r was generated by (15) as a function of L , $0.1\lambda \leq L \leq 2\lambda$, for several values of d with $R_0 = 500\Omega$ as shown in Fig. 2 and for several values of R_0 with $d = 0.2\lambda$ as shown in Fig. 3. As L approaches zero, the reflected field becomes -1, which is the reflected field of the perfectly conducting termination. As L increases, the magnitude decreases to an asymptotic value. The asymptotes imply that there is no longer any contribution from the termination at $x = -L$ and are thus attributable to the scattering from the leading edges of the resistive sheets alone. This was verified by comparing the asymptotic values to the reflected fields produced when the termination is removed from the solution. Figure 4 shows this comparison for $d = 0.2\lambda$, $R_0 = 500\Omega$, and the asymptotes are indistinguishable. This can also be shown by comparing the results of Figs. 2 and 3 to the reflected field from resistive sheets of constant resistivity R_0 with the termination removed, for which the results of each case are almost identical as a function of d and R_0 for large L (see Fig. 5). In the limit L approaches infinity the reflected field in (15) is virtually equivalent to that from an infinite array of half planes of constant resistivity R_0 . Therefore, as lossy parallel plate waveguides, the resistive sheets tremendously reduce the contribution from the termination.

We note that the rate of convergence to the asymptotic values decreases with increasing d and R_0 , providing a larger return for $L \lesssim 0.5\lambda$ than for smaller d and R_0 (see Figs. 2 and 3). Thus the penetration of the incident field increases for larger d and R_0 as expected, which is also evidenced by the increased oscillations attributable to the termination. Even though this penetration increases the contribution from the termination, the asymptotic values decrease with increasing d and R_0 , implying that the leading edge

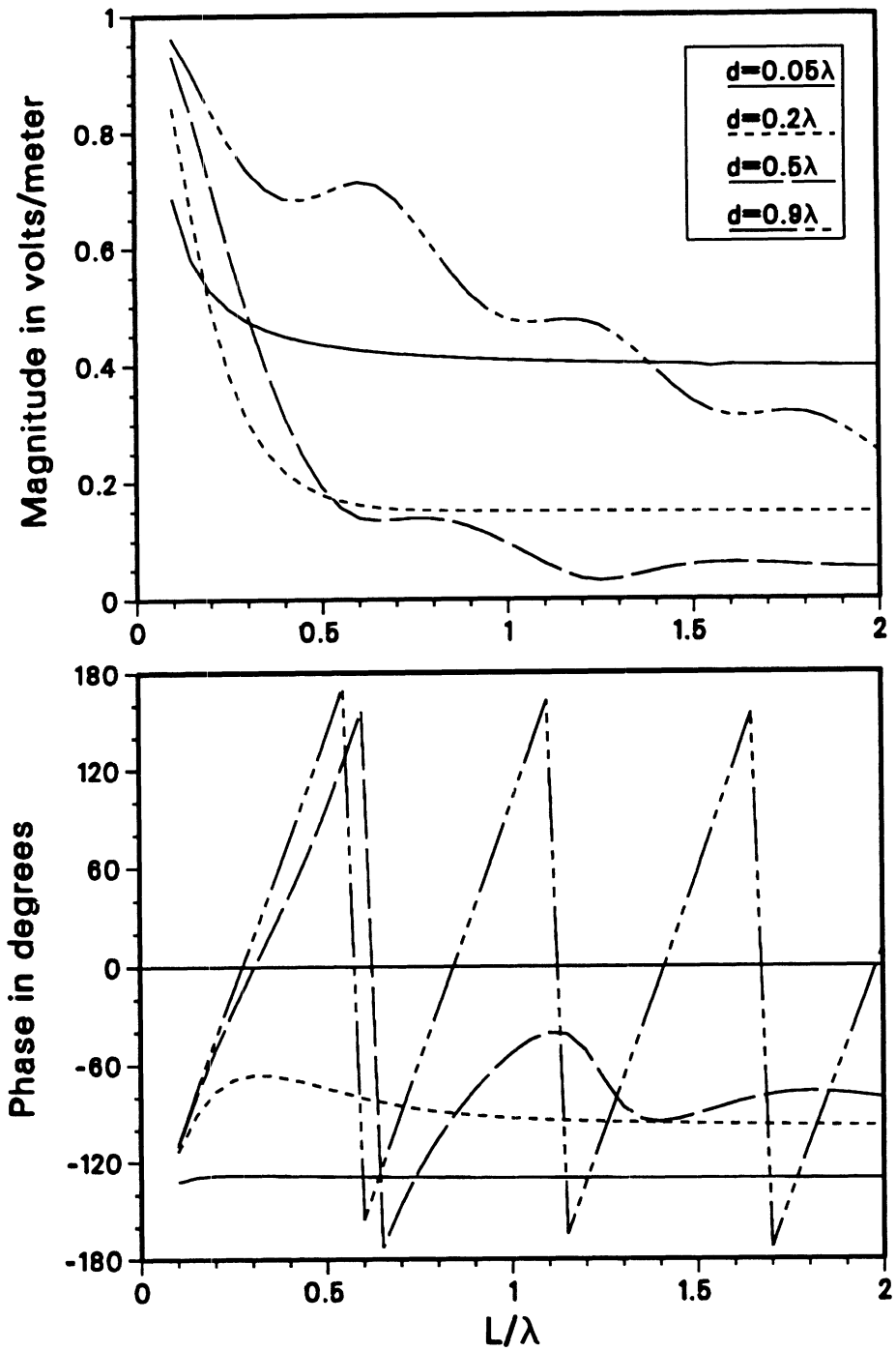


Figure 2: Reflected field E_{z0}^r of the infinite array for several d with $\phi_0 = 0$ and $R_0 = 500\Omega$ (quadratic taper).

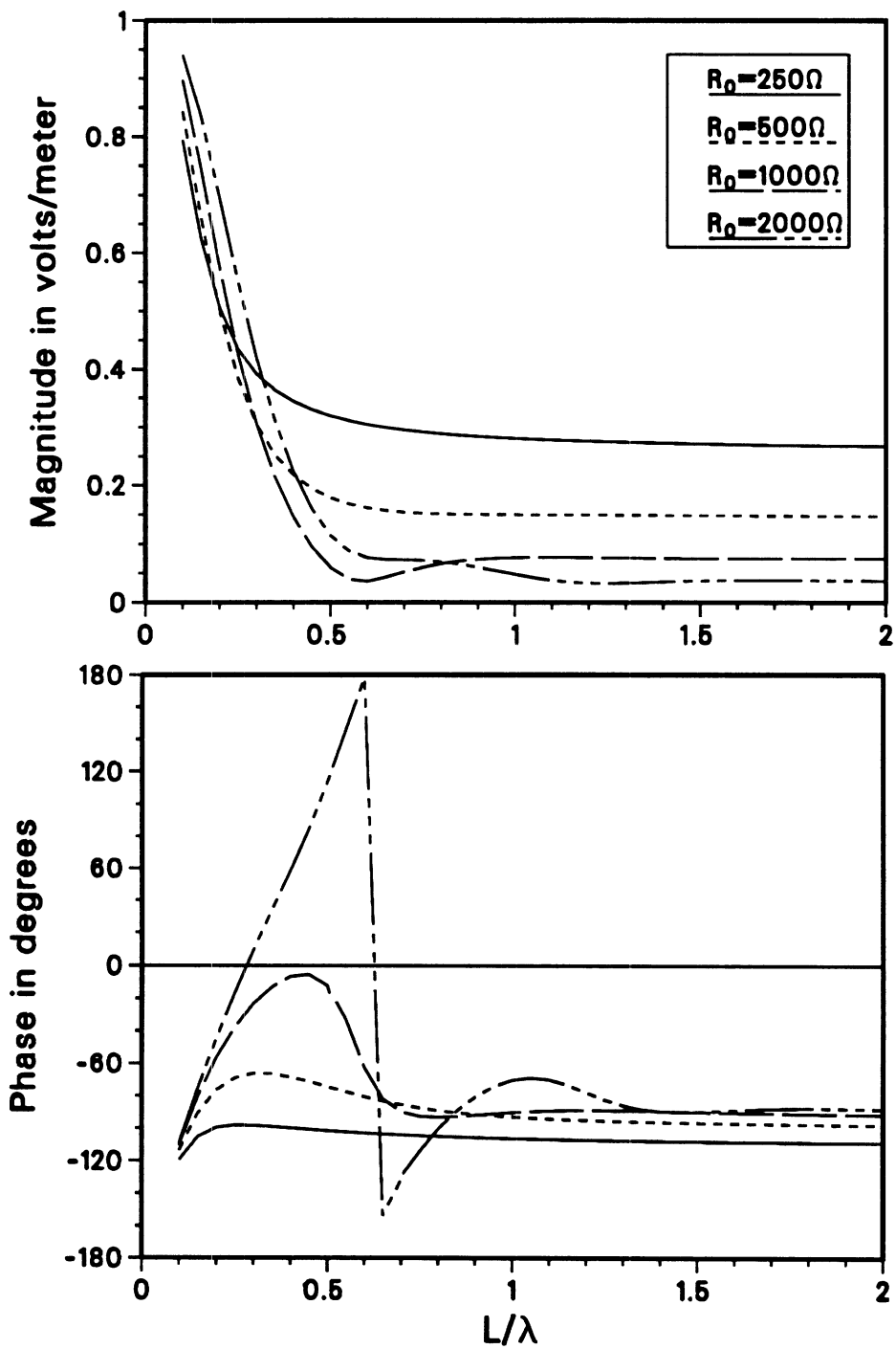


Figure 3: Reflected field E_{z0}^r of the infinite array for several R_0 with $\phi_0 = 0$ and $d = 0.2\lambda$ (quadratic taper).

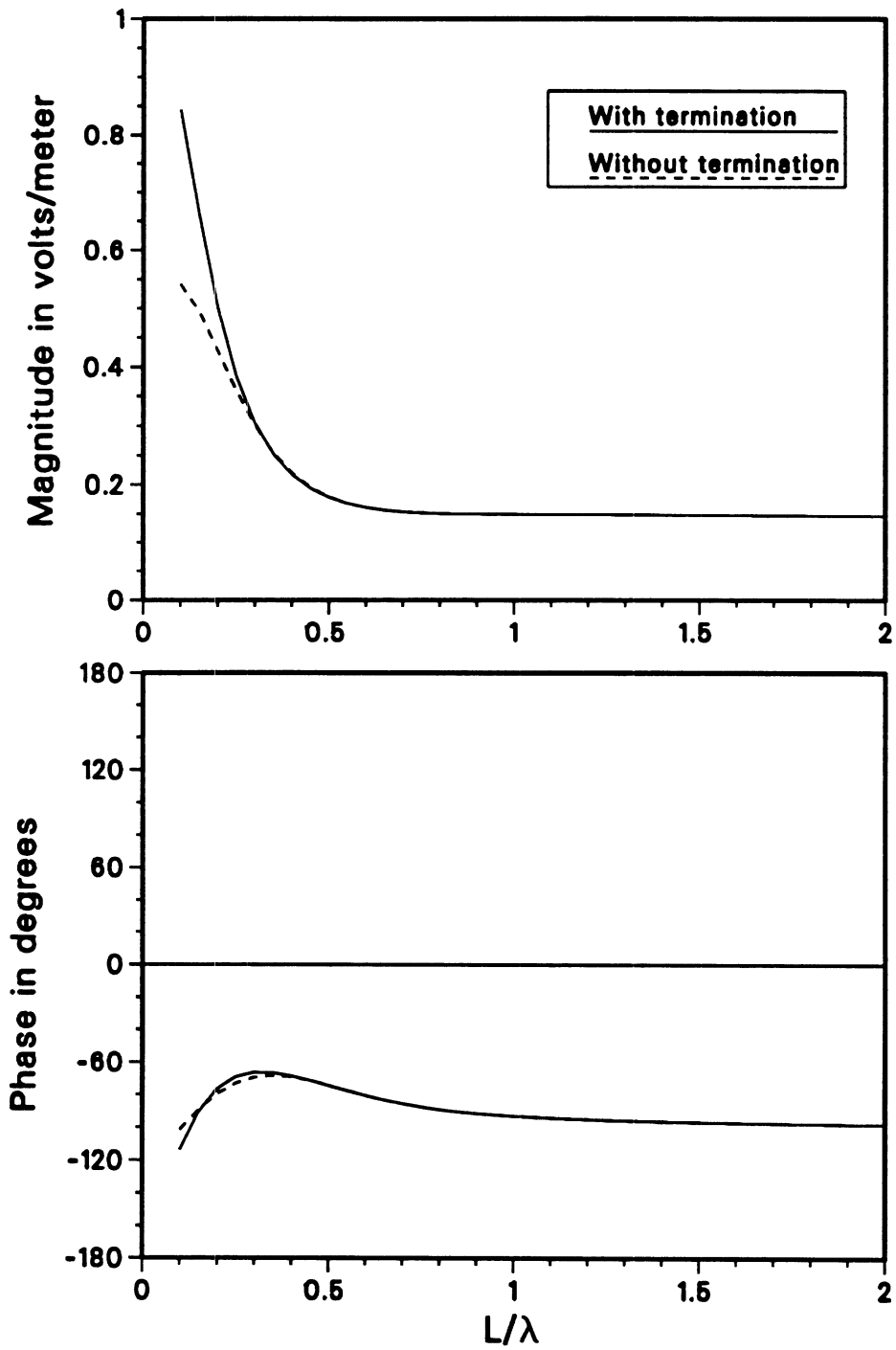


Figure 4: Comparison of reflected fields $E_{z_0}^r$ of the infinite array with and without the perfectly conducting termination at $x = -L$ for $\phi_0 = 0$, $d = 0.2\lambda$, and $R_0 = 500\Omega$ (quadratic taper).

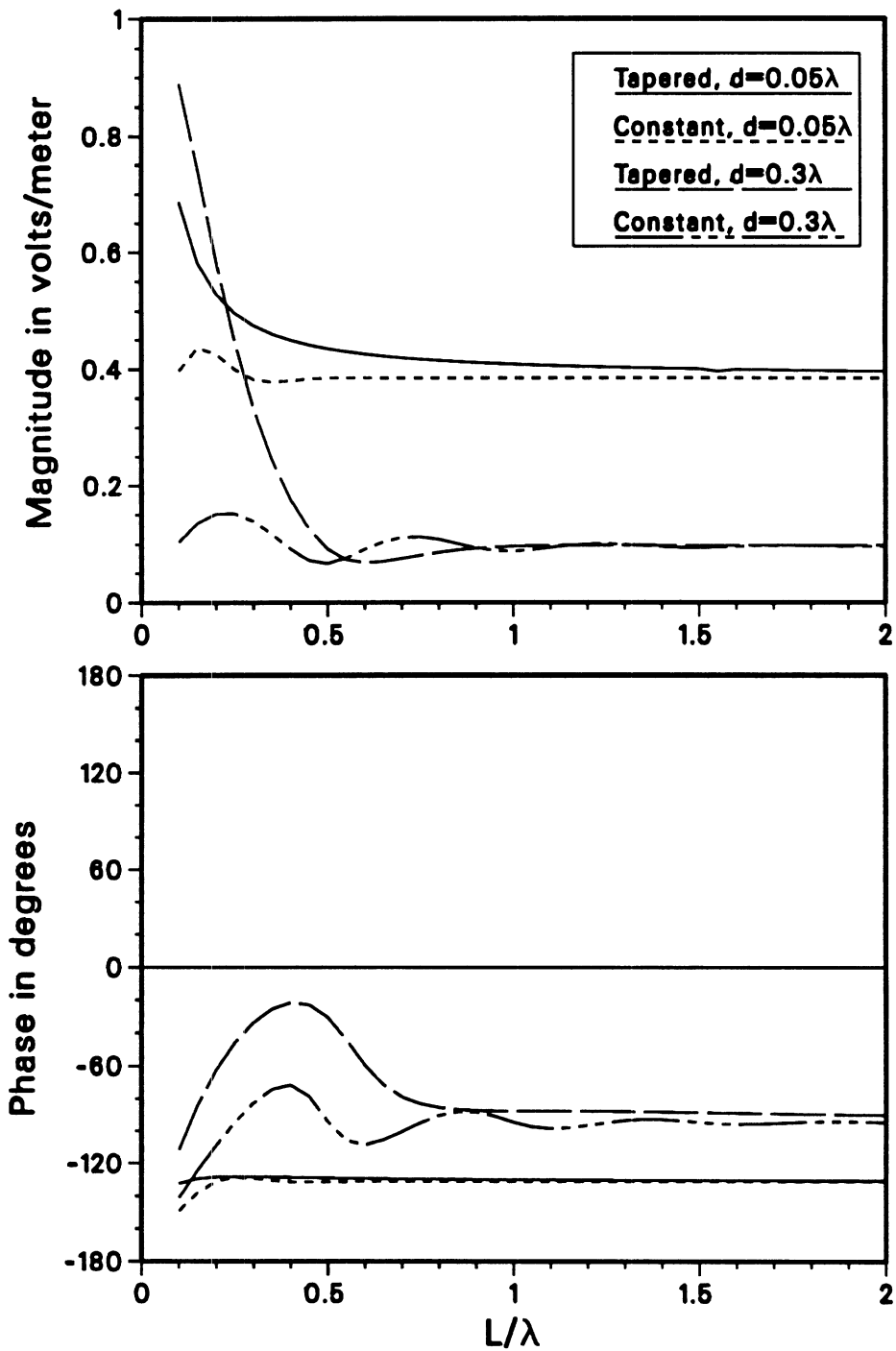


Figure 5: Comparison of reflected fields E_{z0}^r of the infinite array with quadratic taper and constant resistivity profiles for $\phi_0 = 0$ and $R_0 = 500\Omega$.

contribution is reduced. This behavior is expected for increasing R_0 , but requires further analysis as a function of d .

As d approaches zero, the interaction between the adjacent sheets in the infinite array increases, and the effective resistivity decreases since the sheets appear in parallel with one another. Furthermore, in the limit the infinite array becomes a lossy material slab of thickness L , which can be characterized by some reflection coefficient, assuming the field transmitted at the leading surface is completely absorbed. Modeling the slab with a resistive sheet of resistivity $R_0/2$ in the $x = 0$ plane, we find that

$$E_z^r = -\frac{1}{1 + \frac{R_0}{Z} \cos \phi_0} e^{ik(x \cos \phi_0 - y \sin \phi_0)},$$

which gives comparable results to the asymptotic values of (15) for $d = 0.05\lambda$. The choice of a resistive sheet model as opposed to a lossy dielectric layer one is consistent with the nature of the transmitted field propagating in the $\phi_0 - \pi$ direction and gives the proper dependence on R_0 . For $d \simeq 0.2\lambda$, it was observed from the data that for normal incidence and $R_0 \gtrsim 500\Omega$ the asymptotic value was comparable to the edge on backscatter of a resistive half plane of constant resistivity R_0 normalized by the backscatter of a perfectly conducting half plane, giving

$$E_z^r = -\frac{iZ}{4R_0} e^{-\frac{z}{\pi R_0}}, \quad \phi_0 = 0.$$

The accuracy of this model in magnitude and phase implies that any interaction between adjacent resistive sheets is negligible. As d is increased from 0.2λ , the asymptotic values of E_z^r continue to decrease, suggesting that the absorption of the incident field is increased, though this phenomenon is not yet understood.

5.2 H -polarization

The reflected field H_{z0}^r was generated by (16), and, as expected, the magnitude is a constant as a function of L for $\phi_0 = 0$, as shown in Fig. 6 with $d = 0.2\lambda$, $R_0 = 500\Omega$. The reflection is simply the reflected field from the perfectly conducting termination, which is e^{i2kL} . With the electric vector perpendicular to the resistive sheets, the sheets have no effect on the incident field, and this is so as a function of d , L , and R_0 . For $\phi_0 \simeq 0$ the result still remains virtually unchanged, and as ϕ_0 is increased the effect of the resistive sheets increases, which is evidenced by the results in Fig. 6 for $\phi_0 = \pi/4$.

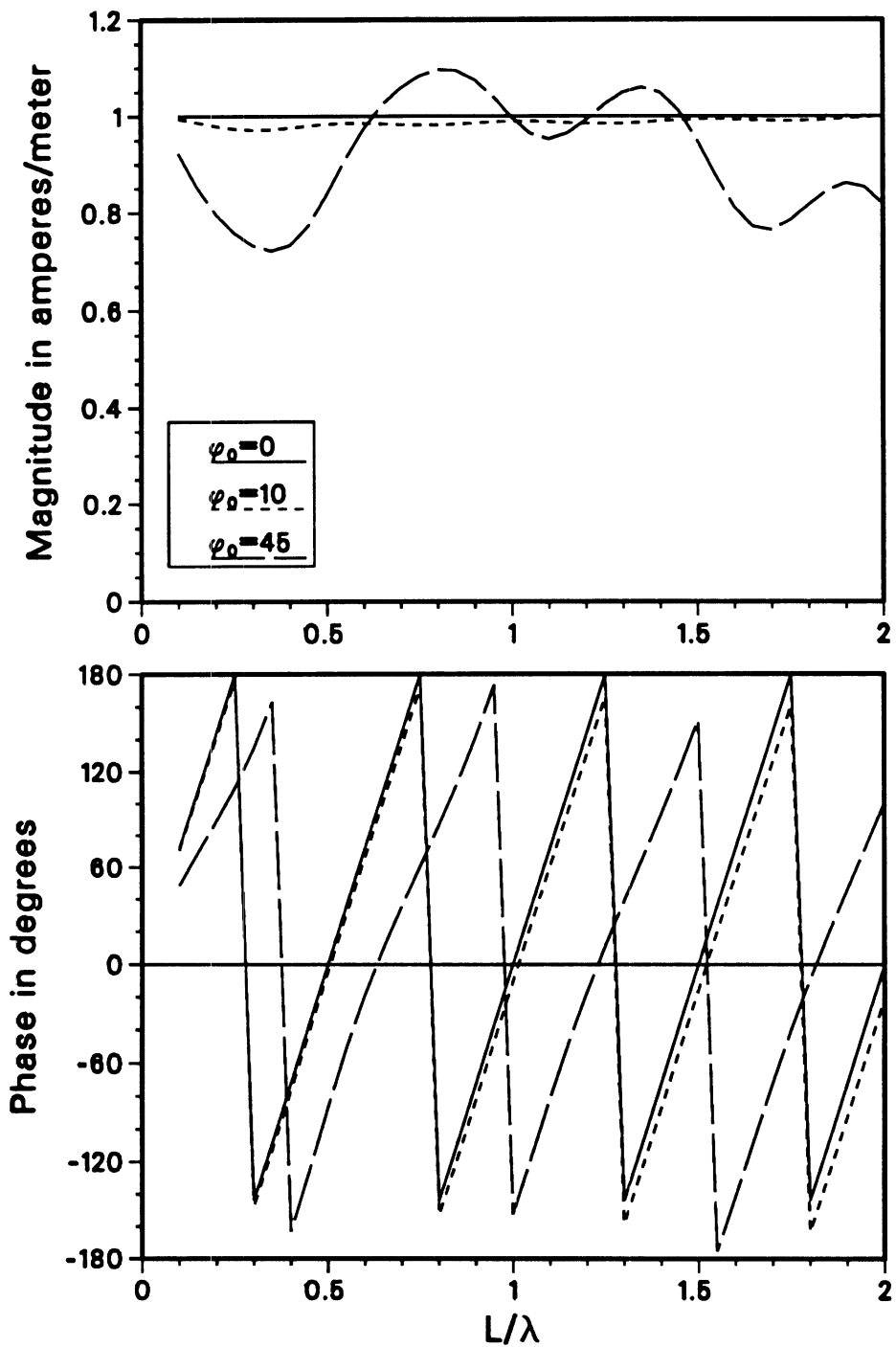


Figure 6: Reflected field $H_{z_0}^r$ of the infinite array for several ϕ_0 with $d = 0.2\lambda$ and $R_0 = 500\Omega$ (quadratic taper).

References

- [1] Kovacs, A. and R. M. Morey, "Radar Anisotropy of Sea Ice Due to Preferred Azimuthal Orientation of the Horizontal c Axes of Ice Crystals," *J. Geophys. Res.*, Vol. 83, No. C12, pp. 6037-6046, 1978.
- [2] Sarabandi, K., "Scattering from Variable Resistive and Impedance Sheets," *J. Electromagn. Waves Appl.*, Vol. 4, No. 9, pp. 865-891, 1990.

Appendix

The following is the fortran code INFARR listed with its subroutines:

```
C
C      INFARR.FTN                      John R. Natzke
C                                      9/17/92
C
C      This program computes the scattering from an infinite array
C      of parallel resistive sheets of length L and separation d
C      under plane wave illumination with the  $\exp(-i\omega t)$  time
C      convention.  The sheets are terminated with a perfectly
C      conducting plane.
C
C      The user is prompted for all of the input parameters, and
C      the output is the specular reflected field written to the
C      files smagdt (magnitude) and sphadt (phase).
C
C
C      parameter(ip=2500,ic=4)
C      integer EorH,indx(ic),xory
C      real k,L,ky
C      real RC(ic),R(ip),xp(ip),yp(ip),dxC(ic),dyC(ic),drC(ic)
C      complex ci,ctemp,carg,Erz,Hrz,Lzn,Lxn,Lyn,dGxn,dGyn
C      complex Z(ip,ip),In(ip),Vm,kx,kxp,Km
C      integer ipvt(ip),iretrn
C      complex wk(ip)
C      data ci,pi/(0.,1.),3.141593/
C      open(3,file='smagdt')
C      open(4,file='sphadt')
c...Declaring constant values
10      k=2*pi
C      Zo=sqrt(4.e-07*pi/8.854e-12)
C      noIter=1
C      drg=pi/180
c...Prompting user for input parameters
print *, '1) E- or 2) H-pol?'
read *,EorH
print *, '1) With or 2) without pec termination?'
read *,ibac
numC=1
if(ibac .eq. 1) numC=2
indx(1)=1
15      do 20 i=1,numC
C          if(i .eq. 1)then
C              print*, 'For the resistive strips --'
C              xory=1
C              print *, ' Enter L:'
C              read *,L
C              W=L
C              xm=-W
C              ym=0
C              print *, ' 1) Constant or 2) Tapered Resistivity?'
C              read *,ich
C              ich1=ich
C          else
C              print *, 'For the back short --'
C              ich=1
C              xory=2
C              xm=-L
C              ym=0
C              print *, ' Enter d:'
C              read *,d
C              W=d
C          endif
c      print *, ' Number of elements',int(10*(1-W)+100*W)
c      read *,
C      if(W .lt. 1)then
```

```

        N=int(10*(1-W)+100*W)
    else
        N=int(100*W)
    endif
    drC(i)=W/N
    if(xory .eq. 1)then
        dxC(i)=W/N
        dyC(i)=0
    else
        dxC(i)=0
        dyC(i)=W/N
    endif
    if(ich .eq. 1)then
        if(i .eq. 1)then
            print*, ' Surface resistivity (ohms):'
            read*,RC(i)
        else
            RC(i)=0
        endif
    else
        RC(i)=0
        rLod=W
        print*, ' Maximum resistivity Ro (ohms):'
        read*,Ro
        print *, ' Order of taper (1 = linear):'
        read *,itap
    endif
    dxstar=dxC(i)/2
    dystar=dyC(i)/2
    do 18 j=indx(i),indx(i)+N-1
c   Element position arrays
        xp(j)=xm+dxstar+(j-indx(i))*dxC(i)
        yp(j)=ym+dystar+(j-indx(i))*dyC(i)
c   Resistive Loading
        if(ich .eq. 1)then
            R(j)=RC(i)
        else
            if(xory .eq. 1)then
                diff=xp(j)-xm
            else
                diff=yp(j)-ym
            endif
            R(j)=Ro*(diff/rLod)**itap+RC(i)
        endif
18    continue
        indx(i+1)=indx(i)+N
20    continue
    if(numC .eq. 1)then
        print *, ' Enter d:'
        read *,d
    endif
    M=indx(numC)+N-1
    print*, ' Total number of current elements M =',M
    print*, ' Pmax:'
    read*,Pmax
c   print *, 'Number of angular iterations:'
c   read *,noIter
c   if(noIter .gt. 1)then
c       star=1*drg
c       fini=pi-star
c       step=(fini-star)/(noIter-1)
c       phi=star
c   endif
    print*, ' Incident angle:'
    read*,phio
    phio=phio*drg
    do 30 jp=-8*Pmax,8*Pmax,1
        chk=jp/d-sin(phio)
        if(chk .eq. 1 .or. chk .eq. -1)then
            print*, ' SINGULARITY for p =',jp,': Change d or phio.'
            GOTO 15
        endif
30    continue

```

```

30    continue
      print *, 'Number of length iterations:'
      read *, noIter
      if(noIter .gt. 1)then
        print*, 'Final length'
        read*, fini
        wstar=L
        wstep=(fini-wstar)/(noIter-1)
      endif
      DO 700 iter=1,noIter
        print *, ' '
        print *, ' M = ',M,' L = ',L,' d = ',d,' phio = ',phio/drg
        print *, ' Generating impedance and source matrices . . . '
c*****
c*****      Impedance, Source,      *****
c*****      and Current Matrices      *****
c*****
      A=k*Zo/d
      inr=1
      do 250 j=1,M
        xm=xp(j)
        ym=yp(j)
c  Impedance matrix elements
        inc=1
        do 244 i=1,M
          if((inr.eq.inc.and.j.eq.indx(inr))
& .or.(inr.eq.inc.and.inc.eq.2.and.i.eq.indx(inc))
& .or.(inr.ne.inc))then
            xn=xp(i)
            yn=yp(i)
            dx=dxC(inc)
            dy=dyC(inc)
            dr=drC(inc)
            Z(j,i)=0
            lim=1
            if(i .eq. j) lim=8
            if(EorH .eq. 1)then
c  E-pol
              do 230 jp=-lim*Pmax,lim*Pmax
                ky=2*pi*jp/d-k*sin(phio)
                carg=k**2-ky**2
                kx=csqrt(carg)
                if(inc .eq. 1)then
                  Lzn=cexp(ci*ky*(ym-yn))/kx**2
                  if(i .eq. j)then
                    Lzn=Lzn*ci*(1-cexp(ci*kx*dx/2))
                  else
                    Lzn=Lzn*cexp(ci*kx*abs(xm-xn))
& *csin(kx*dx/2)
                endif
                else
                  Lzn=cexp(ci*(kx*abs(xm-xn)+ky*(ym-yn)))
& /kx*dy/2*sinc(ky*dy/2)
                endif
                Z(j,i)=Z(j,i)+A*Lzn
              enddo
            endif
          enddo
        enddo
      enddo
230    continue
      else
c  H-pol
        do 240 jp=-lim*Pmax,lim*Pmax
          ky=2*pi*jp/d-k*sin(phio)
          carg=k**2-ky**2
          kx=csqrt(carg)
          kxp=ci*k*(abs(1.*jp)/d-sgn(1.*jp)*sin(phio))
          if(inr .eq. 1)then
c  Exs--
            if(inc .eq. 1)then
              Lxn=cexp(ci*ky*ym)/kx**2
              if(i .eq. j)then
                Lxn=Lxn*ci*(1-cexp(ci*kx*dx/2))
              else
                Lxn=Lxn*cexp(ci*kx*abs(xm-xn))
            endif
          endif
        enddo
      enddo

```



```

&          *csin(kx*dx/2)
endif
dGxn=0
xi=xn-dx/2
do 232 iend=1,2
  dGxn=ci/2*sgn(xm-xi)
  *cexp(ci*ky*ym)*(cexp(ci*kx*abs(xm-xi))
& -cexp(ci*kxp*abs(xm-xi)))-dGxn
  if(jp .eq. 0)then
    dGxn=ci/2*sgn(xm-xi)
    *cexp(-ci*k*(ym-yn)*sin(phio))
    *(exp(k*abs(xm-xi)*sin(phio))
& / (1-cexp(-k/d*(abs(xm-xi)-ci*(ym-yn))))
& +cexp(-k*(abs(xm-xi)*(sin(phio)+1./d)
& +ci*(ym-yn)/d))/(1-cexp(-k/d*(abs(xm-xi)
& +ci*(ym-yn)))))+dGxn
  endif
  xi=xn+dx/2
232 continue
  Z(j,i)=Z(j,i)+A*(Lxn-dGxn/k**2)
else
  dGxn=0
  yi=yn-dy/2
  do 234 iend=1,2
    dGxn=ci/2*sgn(xm-xn)
    *cexp(ci*ky*(ym-yi))*(cexp(ci*kx*abs(xm-xn))
& -cexp(ci*kxp*abs(xm-xn)))-dGxn
    if(jp .eq. 0)then
      dGxn=ci/2*sgn(xm-xn)
      *cexp(-ci*k*(ym-yi)*sin(phio))
      *(exp(k*abs(xm-xn)*sin(phio))
& / (1-cexp(-k/d*(abs(xm-xn)-ci*(ym-yi))))
& +cexp(-k*(abs(xm-xn)*(sin(phio)+1./d)
& +ci*(ym-yi)/d))/(1-cexp(-k/d*(abs(xm-xn)
& +ci*(ym-yi)))))+dGxn
    endif
    yi=yn+dy/2
234 continue
    Z(j,i)=Z(j,i)-A*dGxn/k**2
  endif
  else
c    Eys--
    if(inc .eq. 1)then
      dGyn=0
      xi=xn-dx/2
      do 236 iend=1,2
        if(jp .ne. 0)then
          dGyn=ci/2*cexp(ci*ky*(ym-yn))
          *(ky/kx*cexp(ci*kx*abs(xm-xi))+ci
& *sgn(1.*jp)*cexp(ci*kxp*abs(xm-xi)))-dGyn
        else
          dGyn=ci/2*(-tan(phio)*cexp(ci*k
& *(abs(xm-xi)*cos(phio)-(ym-yn)*sin(phio))
& -ci*cexp(-k*(abs(xm-xi)/d+ci*(ym-yn)
& *sin(phio)))*(cexp(k*(abs(xm-xi)*sin(phio)
& +ci*(ym-yn)/d))/(1-cexp(-k/d*(abs(xm-xi)
& -ci*(ym-yn)))))-cexp(-k*(abs(xm-xi)
& *sin(phio)+ci*(ym-yn)/d))/(1-cexp(-k/d
& *(abs(xm-xi)+ci*(ym-yn)))))-dGyn
        endif
        xi=xn+dx/2
236 continue
        Z(j,i)=Z(j,i)-A*dGyn/k**2
      else
        Lyn=cexp(ci*(kx*abs(xm-xn)+ky*(ym-yn)))
& /kx*dy/2*sinc(ky*dy/2)
        dGyn=0
        yi=yn-dy/2
        do 238 iend=1,2
          if(jp .ne. 0)then
            dGyn=ci/2*cexp(ci*ky*(ym-yi))
            *(ky/kx*cexp(ci*kx*abs(xm-xn))+ci

```

```

&          *sgn(1.*jp)*cexp(ci*kxp*abs(xm-xn))-dGyn
      else
        dGyn=ci/2*(-tan(phio)*cexp(ci*k
&          *(abs(xm-xn)*cos(phio)-(ym-yi)*sin(phio)))
&          -ci*cexp(-k*(abs(xm-xn)/d+ci*(ym-yi)
&          *sin(phio)))*(cexp(k*(abs(xm-xn)*sin(phio)
&          +ci*(ym-yi)/d))/(1-cexp(-k/d*(abs(xm-xn)
&          -ci*(ym-yi)))))-cexp(-k*(abs(xm-xn)
&          *sin(phio)+ci*(ym-yi)/d))/(1-cexp(-k/d
&          *(abs(xm-xn)+ci*(ym-yi)))))-dGyn
      endif
      yi=yn+dy/2
238      continue
      Z(j,i)=Z(j,i)+A*(Lyn-dGyn/k**2)
      endif
    endif
240    continue
  endif
  else
    if(inr .eq. 1)then
      if(i .lt. j)then
        Z(j,i)=Z(i,j)
      else
        Z(j,i)=Z(j-1,i-1)
      endif
    else
      Z(j,i)=Z(j-1,i-1)
    endif
  endif
  if(i .eq. indx(inc+1)-1) inc=inc+1
244  continue
c...Incident Field (Source) matrix elements
  if(EorH .eq. 1)then
    Vm=cexp(-ci*k*(xm*cos(phio)+ym*sin(phio)))
  else
    if(inr .eq. 1)then
      Vm=Zo*sin(phio)
&      *cexp(-ci*k*(xm*cos(phio)+ym*sin(phio)))
    else
      Vm=Zo*cos(phio)
&      *cexp(-ci*k*(xm*cos(phio)+ym*sin(phio)))
    endif
  endif
  In(j)=Vm
  if(j .eq. indx(inr+1)-1) inr=inr+1
250  continue
c...Resistivity profile
  do 260 j=1,M
    Z(j,j)=Z(j,j)+R(j)
260  continue
c...Calling subroutines to calculate the current matrix
  print *, ' Solving [Znm][In] = [Vm] . . .'
  call CGECO(Z,ip,M,ipvt,cond,wk)
  print *, ' The condition number is ',cond
  call CGESL(Z,ip,M,ipvt,In,0)
c*****
c*****      Diffracted Field      *****
c*****
c...Direction of diffracted field(s)
  jp=0
  dowhile(jp .lt. d*(1+sin(phio)))
    print*, ' Reflected field exists for p =',jp,
&          ', phip =',asin(jp/d-sin(phio))/drg,'.'
    jp=jp+1
  enddo
  jp=-1
  dowhile(jp .gt. -d*(1-sin(phio)))
    print*, ' Reflected field exists for p =',jp,
&          ', phip =',asin(jp/d-sin(phio))/drg,'.'
    jp=jp-1
  enddo
c...Specular reflected field

```

```

print*, ' For p = 0, phip = ', -phio/drg, ', '
jp=0
Erz=0
Hrz=0
ky=k*(jp/d-sin(phio))
carg=k**2-ky**2
kx=csqrt(carg)
do 610 inc=1,numC
  Lzn=0
  do 600 i=indx(inc),indx(inc+1)-1
    xn=xp(i)
    yn=yp(i)
c      print *, i, xn, yn, cabs(In(i))
    Lzn=In(i)*cexp(-ci*(kx*xn+ky*yn))+Lz
600  continue
    dr=drC(inc)
    if(EorH .eq. 1)then
      if(inc .eq. 1)then
        Erz=-k*Zo/kx/d*csin(kx*dr/2)/kx*Lzn
      else
        Erz=-k*Zo/kx/d*dr/2*sinc(ky*dr/2)*Lzn+Erz
      endif
    else
      if(inc .eq. 1)then
        Hrz=1./d*ky/kx**2*csin(kx*dr/2)*Lzn
      else
        Hrz=-1./d*dr/2*sinc(ky*dr/2)*Lzn+Hrz
      endif
    endif
610  continue
    if(EorH .eq. 1)then
      print *, '   Erz: ', cabs(Erz), pha(Erz)
      write(3,*) L,cabs(Erz)
      write(4,*) L,pha(Erz)
    else
      write(3,*) L,cabs(Hrz)
      write(4,*) L,pha(Hrz)
      print *, '   Hrz: ', cabs(Hrz), pha(Hrz)
    endif

if(noIter .gt. 1)then
do 620 i=1,numC
  if(i .eq. 1)then
    ich=ich1
    xory=1
    L=wstar+iter*wstep
    W=L
    xm=-W
    ym=0
  else
    ich=1
    xory=2
    xm=-L
    ym=0
    W=d
  endif
  if(W .lt. 1.)then
    exN=10*(1-W)
  else
    exN=0
  endif
  N=int(exN+100*W)
  drC(i)=W/N
  if(xory .eq. 1)then
    dxC(i)=W/N
    dyC(i)=0
  else
    dxC(i)=0
    dyC(i)=W/N
  endif
  dxstar=dxC(i)/2
  dystar=dyC(i)/2

```

```

do 618 j=indx(i),indx(i)+N-1
  xp(j)=xm+dxstar+(j-indx(i))*dxC(i)
  yp(j)=ym+dystar+(j-indx(i))*dyC(i)
  if(ich .eq. 1)then
    R(j)=RC(i)
  else
    if(xory .eq. 1)then
      diff=xp(j)-xm
    else
      diff=yp(j)-ym
    endif
    R(j)=Ro*(diff/W)**itap+RC(i)
  endif
618   continue
      indx(i+1)=indx(i)+N
620   continue
      M=indx(numC)+N-1
      endif

c      phio=star+(iter)*step
700   continue

      print *, ' Again (1=yes) ? '
      read *,ians
      if(ians .eq. 1) GOTO 10
      print*, ' '
      print*, 'The output files are SMAGDT and SPHADT.'
800   stop
      END
c*****
      FUNCTION SGN(R)
c*****
      if(R .ne. 0)then
        sgn=R/abs(R)
      else
        sgn=1.
      endif
      return
      end
c*****
      FUNCTION PHA(C)
c*****
      complex c
      if(Real(c) .ne. 0)then
        pha=180./3.141593*atan2(aImag(c),Real(c))
      elseif(aImag(c) .ne. 0)then
        pha=aImag(c)/abs(aImag(c))*90
      else
        pha=90
      endif
      return
      end
c*****
c
c      The following subroutines are standard LINPACK routines
c      to perform L-U decomposition and back substitution on a
c      single precision complex matrix. See CC-Memo 407 sec 2.1
c      for documentation on these routines.
c
c*****
c
c      SUBROUTINE CGECO(A, LDA, N, IPVT, RCOND, Z)
c
c*****
c
c      NAASA 2.1.042 CGECO FTN-A 05-02-78 THE UNIV OF MICH COMP CTR
c
c      INTEGER LDA, N, IPVT(1)
c      COMPLEX A(LDA, 1), Z(1)
c      REAL RCOND
c
c
c      CGECO FACTORS A COMPLEX MATRIX BY GAUSSIAN ELIMINATION

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```

C      AND ESTIMATES THE CONDITION OF THE MATRIX.
C
C      IF RCOND IS NOT NEEDED, CGEFA IS SLIGHTLY FASTER.
C      TO SOLVE  $A * X = B$  , FOLLOW CGECO BY CGESL.
C      TO COMPUTE  $INVERSE(A) * C$  , FOLLOW CGECO BY CGESL.
C      TO COMPUTE  $DETERMINANT(A)$  , FOLLOW CGECO BY CGEDI.
C      TO COMPUTE  $INVERSE(A)$  , FOLLOW CGECO BY CGEDI.
C
C      ON ENTRY
C
C      A      COMPLEX(LDA, N)
C             THE MATRIX TO BE FACTORED.
C
C      LDA    INTEGER
C             THE LEADING DIMENSION OF THE ARRAY A .
C
C      N      INTEGER
C             THE ORDER OF THE MATRIX A .
C
C      ON RETURN
C
C      A      AN UPPER TRIANGULAR MATRIX AND THE MULTIPLIERS
C             WHICH WERE USED TO OBTAIN IT.
C             THE FACTORIZATION CAN BE WRITTEN  $A = L * U$  WHERE
C             L IS A PRODUCT OF PERMUTATION AND UNIT LOWER
C             TRIANGULAR MATRICES AND U IS UPPER TRIANGULAR.
C
C      IPVT   INTEGER(N)
C             AN INTEGER VECTOR OF PIVOT INDICES.
C
C      RCOND  REAL
C             AN ESTIMATE OF THE RECIPROCAL CONDITION OF A .
C             FOR THE SYSTEM  $A * X = B$  , RELATIVE PERTURBATIONS
C             IN A AND B OF SIZE EPSILON MAY CAUSE
C             RELATIVE PERTURBATIONS IN X OF SIZE  $EPSILON / RCOND$  .
C             IF RCOND IS SO SMALL THAT THE LOGICAL EXPRESSION
C              $1.0 + RCOND .EQ. 1.0$ 
C             IS TRUE, THEN A MAY BE SINGULAR TO WORKING
C             PRECISION. IN PARTICULAR, RCOND IS ZERO IF
C             EXACT SINGULARITY IS DETECTED OR THE ESTIMATE
C             UNDERFLOWS.
C
C      Z      COMPLEX(N)
C             A WORK VECTOR WHOSE CONTENTS ARE USUALLY UNIMPORTANT.
C             IF A IS CLOSE TO A SINGULAR MATRIX, THEN Z IS
C             AN APPROXIMATE NULL VECTOR IN THE SENSE THAT
C              $NORM(A * Z) = RCOND * NORM(A) * NORM(Z)$  .
C
C      LINPACK. THIS VERSION DATED 07/14/77 .
C      CLEVE MOLER, UNIVERSITY OF NEW MEXICO, ARGONNE NATIONAL LABS.
C
C      SUBROUTINES AND FUNCTIONS
C
C      LINPACK CGEFA
C      BLAS CAXPY, CDOTC, CSSCAL, SCASUM
C      FORTRAN ABS, AIMAG, AMAX1, CMLPX, CONJG, REAL
C
C      INTERNAL VARIABLES
C
C      COMPLEX CDOTC, EK, T, WK, WKM
C      REAL ANORM, S, SCASUM, SM, YNORM
C      INTEGER INFO, J, K, KB, KP1, L
C
C      COMPLEX ZDUM, ZDUM1, ZDUM2, CSIGN1
C      REAL CABS1
C      CABS1(ZDUM) = ABS(REAL(ZDUM)) + ABS(AIMAG(ZDUM))
C      CSIGN1(ZDUM1, ZDUM2) = CABS1(ZDUM1) * (ZDUM2 / CABS1(ZDUM2))
C
C      Compute 1-NORM of A
C
C      ANORM = 0.0E0
C      DO 10 J = 1, N

```

```

ANORM = AMAX1(ANORM, SCASUM(N, A(1, J), 1))
10 CONTINUE
C
CCC Factor
C
CALL CGEFA(A, LDA, N, IPVT, INFO)
C
RCOND = 1/(NORM(A)*(ESTIMATE OF NORM(INVERSE(A)))) .
ESTIMATE = NORM(Z)/NORM(Y) WHERE A*Z = Y AND CTRANS(A)*Y = E .
CTRANS(A) IS THE CONJUGATE TRANSPOSE OF A .
THE COMPONENTS OF E ARE CHOSEN TO CAUSE MAXIMUM LOCAL
GROWTH IN THE ELEMENTS OF W WHERE CTRANS(U)*W = E .
THE VECTORS ARE FREQUENTLY RESCALED TO AVOID OVERFLOW.
C
SOLVE CTRANS(U)*W = E
C
EK = CMPLX(1.0E0, 0.0E0)
DO 20 J = 1, N
  Z(J) = CMPLX(0.0E0, 0.0E0)
20 CONTINUE
DO 100 K = 1, N
  IF (CABS1(Z(K)) .NE. 0.0E0) EK = CSIGN1(EK, -Z(K))
  IF (CABS1(EK-Z(K)) .LE. CABS1(A(K, K))) GO TO 30
  S = CABS1(A(K, K))/CABS1(EK-Z(K))
  CALL CSSCAL(N, S, Z, 1)
  EK = CMPLX(S, 0.0E0)*EK
30 CONTINUE
  WK = EK - Z(K)
  WKM = -EK - Z(K)
  S = CABS1(WK)
  SM = CABS1(WKM)
  IF (CABS1(A(K, K)) .EQ. 0.0E0) GO TO 40
  WK = WK/CONJG(A(K, K))
  WKM = WKM/CONJG(A(K, K))
  GO TO 50
40 CONTINUE
  WK = CMPLX(1.0E0, 0.0E0)
  WKM = CMPLX(1.0E0, 0.0E0)
50 CONTINUE
  KP1 = K + 1
  IF (KP1 .GT. N) GO TO 90
  DO 60 J = KP1, N
    SM = SM + CABS1(Z(J)+WKM*CONJG(A(K, J)))
    Z(J) = Z(J) + WK*CONJG(A(K, J))
    S = S + CABS1(Z(J))
60 CONTINUE
  IF (S .GE. SM) GO TO 80
  T = WKM - WK
  WK = WKM
  DO 70 J = KP1, N
    Z(J) = Z(J) + T*CONJG(A(K, J))
70 CONTINUE
80 CONTINUE
90 CONTINUE
  Z(K) = WK
100 CONTINUE
  S = 1.0E0/SCASUM(N, Z, 1)
  CALL CSSCAL(N, S, Z, 1)
C
CCC Solve CTRANS(L)*Y = V
C
DO 120 KB = 1, N
  K = N + 1 - KB
  IF (K .LT. N) Z(K) = Z(K) + CDOTC(N-K, A(K+1, K), 1, Z(K+1), 1)
  IF (CABS1(Z(K)) .LE. 1.0E0) GO TO 110
  S = 1.0E0/CABS1(Z(K))
  CALL CSSCAL(N, S, Z, 1)
110 CONTINUE
  L = IPVT(K)
  T = Z(L)
  Z(L) = Z(K)
  Z(K) = T

```

```

120 CONTINUE
   S = 1.0E0/SCASUM(N,Z,1)
   CALL CSSCAL(N,S,Z,1)
C
   YNORM = 1.0E0
C
CCC  Solve L*V = Y
C
   DO 140 K = 1, N
     L = IPVT(K)
     T = Z(L)
     Z(L) = Z(K)
     Z(K) = T
     IF (K .LT. N) CALL CAXPY(N-K,T,A(K+1,K),1,Z(K+1),1)
     IF (CABS1(Z(K)) .LE. 1.0E0) GO TO 130
     S = 1.0E0/CABS1(Z(K))
     CALL CSSCAL(N,S,Z,1)
     YNORM = S*YNORM
130 CONTINUE
140 CONTINUE
   S = 1.0E0/SCASUM(N,Z,1)
   CALL CSSCAL(N,S,Z,1)
   YNORM = S*YNORM
C
CCC  Solve U*Z = V
C
   DO 160 KB = 1, N
     K = N + 1 - KB
     IF (CABS1(Z(K)) .LE. CABS1(A(K,K))) GO TO 150
     S = CABS1(A(K,K))/CABS1(Z(K))
     CALL CSSCAL(N,S,Z,1)
     YNORM = S*YNORM
150 CONTINUE
     IF (CABS1(A(K,K)) .NE. 0.0E0) Z(K) = Z(K)/A(K,K)
     IF (CABS1(A(K,K)) .EQ. 0.0E0) Z(K) = CMPLX(1.0E0,0.0E0)
     T = -Z(K)
     CALL CAXPY(K-1,T,A(1,K),1,Z(1),1)
160 CONTINUE
C   MAKE ZNORM = 1.0
   S = 1.0E0/SCASUM(N,Z,1)
   CALL CSSCAL(N,S,Z,1)
   YNORM = S*YNORM
C
   IF (ANORM .NE. 0.0E0) RCOND = YNORM/ANORM
   IF (ANORM .EQ. 0.0E0) RCOND = 0.0E0
   RETURN
   END
C*****C
C
   SUBROUTINE CGEFA(A,LDA,N,IPVT,INFO)
C
C*****C
C
C NAASA 2.1.043 CGEFA   FTN-A 05-02-78   THE UNIV OF MICH COMP CTR
C
   INTEGER LDA,N,IPVT(1),INFO
   COMPLEX A(LDA,1)
C
C   CGEFA FACTORS A COMPLEX MATRIX BY GAUSSIAN ELIMINATION.
C
C   CGEFA IS USUALLY CALLED BY CGECO, BUT IT CAN BE CALLED
C   DIRECTLY WITH A SAVING IN TIME IF RCOND IS NOT NEEDED.
C   (TIME FOR CGECO) = (1 + 9/N)*(TIME FOR CGEFA) .
C
C   ON ENTRY
C
C     A      COMPLEX(LDA, N)
C           THE MATRIX TO BE FACTORED.
C
C     LDA    INTEGER
C           THE LEADING DIMENSION OF THE ARRAY A .
C

```

```

C      N      INTEGER
C      THE ORDER OF THE MATRIX A .
C
C ON RETURN
C
C      A      AN UPPER TRIANGULAR MATRIX AND THE MULTIPLIERS
C      WHICH WERE USED TO OBTAIN IT.
C      THE FACTORIZATION CAN BE WRITTEN A = L*U WHERE
C      L IS A PRODUCT OF PERMUTATION AND UNIT LOWER
C      TRIANGULAR MATRICES AND U IS UPPER TRIANGULAR.
C
C      IPVT   INTEGER(N)
C      AN INTEGER VECTOR OF PIVOT INDICES.
C
C      INFO   INTEGER
C      = 0   NORMAL VALUE.
C      = K   IF U(K,K) .EQ. 0.0 . THIS IS NOT AN ERROR
C      CONDITION FOR THIS SUBROUTINE, BUT IT DOES
C      INDICATE THAT CGESL OR CGEDI WILL DIVIDE BY ZERO
C      IF CALLED. USE RCOND IN CGECO FOR A RELIABLE
C      INDICATION OF SINGULARITY.
C
C LINPACK. THIS VERSION DATED 07/14/77 .
C CLEVE MOLER, UNIVERSITY OF NEW MEXICO, ARGONNE NATIONAL LABS.
C
C SUBROUTINES AND FUNCTIONS
C
C BLAS CAXPY,CSCAL,ICAMAX
C FORTRAN ABS,AIMAG,CMLX,REAL
C
C INTERNAL VARIABLES
C
C COMPLEX T
C INTEGER ICAMAX,J,K,KP1,L,NM1
C
C COMPLEX ZDUM
C REAL CABS1
C CABS1(ZDUM) = ABS(REAL(ZDUM)) + ABS(AIMAG(ZDUM))
C
C Gaussian elimination with partial pivoting
C
C      INFO = 0
C      NM1 = N - 1
C      IF (NM1 .LT. 1) GO TO 70
C      DO 60 K = 1, NM1
C          KP1 = K + 1
C
C      FIND L = PIVOT INDEX
C
C      L = ICAMAX(N-K+1,A(K,K),1) + K - 1
C      IPVT(K) = L
C
C Zero pivot implies this column already triangularized
C
C      IF (CABS1(A(L,K)) .EQ. 0.0E0) GO TO 40
C
C      Interchange if necessary
C
C      IF (L .EQ. K) GO TO 10
C          T = A(L,K)
C          A(L,K) = A(K,K)
C          A(K,K) = T
C
C 10      CONTINUE
C
C Compute multipliers
C
C      T = -CMLX(1.0E0,0.0E0)/A(K,K)
C      CALL CSCAL(N-K,T,A(K+1,K),1)
C
C Row elimination with column indexing
C
C      DO 30 J = KP1, N

```



```

                T = A(L,J)
                IF (L .EQ. K) GO TO 20
                A(L,J) = A(K,J)
                A(K,J) = T
20             CONTINUE
                CALL CAXPY(N-K,T,A(K+1,K),1,A(K+1,J),1)
30             CONTINUE
                GO TO 50
40             CONTINUE
                INFO = K
50             CONTINUE
60             CONTINUE
70             CONTINUE
                IPVT(N) = N
                IF (CABS1(A(N,N)) .EQ. 0.0E0) INFO = N
                RETURN
                END
C
C*****C
C
C             SUBROUTINE CGESL(A,LDA,N,IPVT,B,JOB)
C
C*****C
C
C NAASA 2.1.044 CGESL   FTN-A 05-02-78   THE UNIV OF MICH COMP CTR
C
C             INTEGER LDA,N,IPVT(1),JOB
C             COMPLEX A(LDA,1),B(1)
C
C             CGESL SOLVES THE COMPLEX SYSTEM
C             A * X = B  OR  CTRANS(A) * X = B
C             USING THE FACTORS COMPUTED BY CGECO OR CGEFA.
C
C             ON ENTRY
C
C             A             COMPLEX(LDA, N)
C                           THE OUTPUT FROM CGECO OR CGEFA.
C
C             LDA           INTEGER
C                           THE LEADING DIMENSION OF THE ARRAY  A .
C
C             N             INTEGER
C                           THE ORDER OF THE MATRIX  A .
C
C             IPVT          INTEGER(N)
C                           THE PIVOT VECTOR FROM CGECO OR CGEFA.
C
C             B             COMPLEX(N)
C                           THE RIGHT HAND SIDE VECTOR.
C
C             JOB           INTEGER
C                           = 0             TO SOLVE  A*X = B ,
C                           = NONZERO      TO SOLVE  CTRANS(A)*X = B  WHERE
C                                           CTRANS(A) IS THE CONJUGATE TRANSPOSE.
C
C             ON RETURN
C
C             B             THE SOLUTION VECTOR  X .
C
C             ERROR CONDITION
C
C             A DIVISION BY ZERO WILL OCCUR IF THE INPUT FACTOR CONTAINS A
C             ZERO ON THE DIAGONAL.  TECHNICALLY THIS INDICATES SINGULARITY
C             BUT IT IS OFTEN CAUSED BY IMPROPER ARGUMENTS OR IMPROPER
C             SETTING OF LDA .  IT WILL NOT OCCUR IF THE SUBROUTINES ARE
C             CALLED CORRECTLY AND IF CGECO HAS SET RCOND .GT. 0.0
C             OR CGEFA HAS SET INFO .EQ. 0 .
C
C             TO COMPUTE INVERSE(A) * C  WHERE  C  IS A MATRIX
C             WITH  P  COLUMNS
C             CALL CGECO(A,LDA,N,IPVT,RCOND,Z)
C             IF (RCOND IS TOO SMALL) GO TO ...

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C          DO 10 J = 1, P
C              CALL CGESL(A,LDA,N,IPVT,C(1,J),0)
C          10 CONTINUE
C
C          LINPACK. THIS VERSION DATED 07/14/77 .
C          CLEVE MOLER, UNIVERSITY OF NEW MEXICO, ARGONNE NATIONAL LABS.
C
C          SUBROUTINES AND FUNCTIONS
C
C          BLAS CAXPY,CDOTC
C          FORTRAN CONJG
C
C          INTERNAL VARIABLES
C
C          COMPLEX CDOTC,T
C          INTEGER K,KB,L,NM1
C
C          NM1 = N - 1
C          IF (JOB .NE. 0) GO TO 50
C
C          JOB = 0 , SOLVE A * X = B
C          FIRST SOLVE L*Y = B
C
C          IF (NM1 .LT. 1) GO TO 30
C          DO 20 K = 1, NM1
C              L = IPVT(K)
C              T = B(L)
C              IF (L .EQ. K) GO TO 10
C                  B(L) = B(K)
C                  B(K) = T
C          10 CONTINUE
C              CALL CAXPY(N-K,T,A(K+1,K),1,B(K+1),1)
C          20 CONTINUE
C          30 CONTINUE
C
C          NOW SOLVE U*X = Y
C
C          DO 40 KB = 1, N
C              K = N + 1 - KB
C              B(K) = B(K)/A(K,K)
C              T = -B(K)
C              CALL CAXPY(K-1,T,A(1,K),1,B(1),1)
C          40 CONTINUE
C          GO TO 100
C          50 CONTINUE
C
C          JOB = NONZERO, SOLVE CTRANS(A) * X = B
C          FIRST SOLVE CTRANS(U)*Y = B
C
C          DO 60 K = 1, N
C              T = CDOTC(K-1,A(1,K),1,B(1),1)
C              B(K) = (B(K) - T)/CONJG(A(K,K))
C          60 CONTINUE
C
C          NOW SOLVE CTRANS(L)*X = Y
C
C          IF (NM1 .LT. 1) GO TO 90
C          DO 80 KB = 1, NM1
C              K = N - KB
C              B(K) = B(K) + CDOTC(N-K,A(K+1,K),1,B(K+1),1)
C              L = IPVT(K)
C              IF (L .EQ. K) GO TO 70
C                  T = B(L)
C                  B(L) = B(K)
C                  B(K) = T
C          70 CONTINUE
C          80 CONTINUE
C          90 CONTINUE
C          100 CONTINUE
C          RETURN
C          END

```

C

C

```

C*****C
C
C      SUBROUTINE CAXPY(N,CA,CX,INCX,CY,INCY)
C
C*****C
C
C NAASA  1.1.014 CAXPY      FTN-A 05-02-78      THE UNIV OF MICH COMP CTR
C
C      CONSTANT TIMES A VECTOR PLUS A VECTOR.
C      JACK DONGARRA, LINPACK, 6/17/77.
C
C      COMPLEX CX(1),CY(1),CA
C      INTEGER I,INCX,INCY,IX,IY,N
C
C      IF(N.LE.0)RETURN
C      IF (ABS(REAL(CA)) + ABS(AIMAG(CA)) .EQ. 0.0 ) RETURN
C      IF(INCX.EQ.1.AND.INCY.EQ.1)GOTO 20
C
C      Code for unequal increments or equal increments
CCC      Not equal to 1
C
C      IX = 1
C      IY = 1
C      IF(INCX.LT.0)IX = (-N+1)*INCX + 1
C      IF(INCY.LT.0)IY = (-N+1)*INCY + 1
C      DO 10 I = 1,N
C          CY(IY) = CY(IY) + CA*CX(IX)
C          IX = IX + INCX
C          IY = IY + INCY
C 10 CONTINUE
C      RETURN
C
C      Code for both increments equal to 1
CCC
C
C 20 DO 30 I = 1,N
C      CY(I) = CY(I) + CA*CX(I)
C 30 CONTINUE
C      RETURN
C      END
C
C*****C
C      COMPLEX FUNCTION CDOTC(N,CX,INCX,CY,INCY)
C*****C
C
C NAASA  1.1.012 CDOTC      FTN-A 05-02-78      THE UNIV OF MICH COMP CTR
C
C      FORMS THE DOT PRODUCT OF TWO VECTORS, CONJUGATING THE FIRST
C      VECTOR.
C      JACK DONGARRA, LINPACK, 6/17/77.
C
C      COMPLEX CX(1),CY(1),CTEMP
C      INTEGER I,INCX,INCY,IX,IY,N
C
C      CTEMP = (0.0,0.0)
C      CDOTC = (0.0,0.0)
C      IF(N.LE.0)RETURN
C      IF(INCX.EQ.1.AND.INCY.EQ.1)GOTO 20
C
C      Code for unequal increments or equal increments
CCC      Not equal to 1
C
C      IX = 1
C      IY = 1
C      IF(INCX.LT.0)IX = (-N+1)*INCX + 1
C      IF(INCY.LT.0)IY = (-N+1)*INCY + 1
C      DO 10 I = 1,N
C          CTEMP = CTEMP + CONJG(CX(IX))*CY(IY)
C          IX = IX + INCX
C          IY = IY + INCY
C 10 CONTINUE
C      CDOTC = CTEMP
C      RETURN

```

```

C
CCC      Code for both increments equal to 1
C
  20 DO 30 I = 1,N
      CTEMP = CTEMP + CONJG(CX(I))*CY(I)
  30 CONTINUE
      CDOTC = CTEMP
      RETURN
      END

C
C*****C
C
      SUBROUTINE  CSCAL(N,CA,CX,INCX)
C
C*****C
C
C NAASA  1.1.019  CSCAL      FTN-A 05-02-78      THE UNIV OF MICH COMP CTR
C
C      SCALES A VECTOR BY A CONSTANT.
C      JACK DONGARRA, LINPACK,  6/17/77.
C
C      COMPLEX CA,CX(1)
C      INTEGER I,INCX,N,NINCX
C
C      IF(N.LE.0)RETURN
C      IF(INCX.EQ.1)GOTO 20
C
CCC      Code for increment not equal to 1
C
      NINCX = N*INCX
      DO 10 I = 1,NINCX,INCX
          CX(I) = CA*CX(I)
  10 CONTINUE
      RETURN
C
CCC      Code for increment equal to 1
C
  20 DO 30 I = 1,N
      CX(I) = CA*CX(I)
  30 CONTINUE
      RETURN
      END

C
C*****C
C
      SUBROUTINE  CSSCAL(N,SA,CX,INCX)
C
C*****C
C
C NAASA  1.1.018  CSSCAL      FTN-A 05-02-78      THE UNIV OF MICH COMP CTR
C
C      SCALES A COMPLEX VECTOR BY A REAL CONSTANT.
C      JACK DONGARRA, LINPACK,  6/17/77.
C
C      COMPLEX CX(1)
C      REAL SA
C      INTEGER I,INCX,N,NINCX
C
C      IF(N.LE.0)RETURN
C      IF(INCX.EQ.1)GOTO 20
C
CCC      Code for increment not equal to 1
C
      NINCX = N*INCX
      DO 10 I = 1,NINCX,INCX
          CX(I) = CMLPX(SA*REAL(CX(I)),SA*AIMAG(CX(I)))
  10 CONTINUE
      RETURN
C
CCC      Code for increment equal to 1
C
  20 DO 30 I = 1,N

```

```

        CX(I) = CMPLX(SA*REAL(CX(I)),SA*AIMAG(CX(I)))
30 CONTINUE
    RETURN
    END
C
C*****C
    INTEGER FUNCTION ICAMAX(N,CX,INCX)
C*****C
C
C NAASA 1.1.021 ICAMAX   FTN-A 05-02-78   THE UNIV OF MICH COMP CTR
C
C     FINDS THE INDEX OF ELEMENT HAVING MAX. ABSOLUTE VALUE.
C     JACK DONGARRA, LINPACK, 6/17/77.
C
C     COMPLEX CX(1)
C     REAL SMAX
C     INTEGER I, INCX, IX, N
C     COMPLEX ZDUM
C     REAL CABS1
C     CABS1(ZDUM) = ABS(REAL(ZDUM)) + ABS(AIMAG(ZDUM))
C
C     ICAMAX = 1
C     IF(N.LE.1)RETURN
C     IF(INCX.EQ.1)GOTO 20
C
CCC     Code for increment not equal to 1
C
C     IX = 1
C     SMAX = CABS1(CX(1))
C     IX = IX + INCX
C     DO 10 I = 2,N
C         IF(CABS1(CX(IX)).LE.SMAX) GO TO 5
C         ICAMAX = I
C         SMAX = CABS1(CX(IX))
C     5     IX = IX + INCX
C     10 CONTINUE
C     RETURN
C
CCC     Code for increment equal to 1
C
C     20 SMAX = CABS1(CX(1))
C     DO 30 I = 2,N
C         IF(CABS1(CX(I)).LE.SMAX) GO TO 30
C         ICAMAX = I
C         SMAX = CABS1(CX(I))
C     30 CONTINUE
C     RETURN
C     END
C
C*****C
    REAL FUNCTION SCASUM(N,CX,INCX)
C*****C
C
C NAASA 1.1.010 SCASUM   FTN-A 05-02-78   THE UNIV OF MICH COMP CTR
C
C     TAKES THE SUM OF THE ABSOLUTE VALUES OF A COMPLEX VECTOR AND
C     RETURNS A SINGLE PRECISION RESULT.
C     JACK DONGARRA, LINPACK, 6/17/77.
C
C     COMPLEX CX(1)
C     REAL STEMP
C     INTEGER I, INCX, N, NINCX
C
C     SCASUM = 0.0E0
C     STEMP = 0.0E0
C     IF(N.LE.0)RETURN
C     IF(INCX.EQ.1)GOTO 20
C
CCC     Code for increment not equal to 1
C
C     NINCX = N*INCX
C     DO 10 I = 1,NINCX,INCX

```

```
        STEMP = STEMP + ABS(REAL(CX(I))) + ABS(AIMAG(CX(I)))
10 CONTINUE
    SCASUM = STEMP
    RETURN
C
CCC      Code for increment equal to 1
C
20 DO 30 I = 1,N
        STEMP = STEMP + ABS(REAL(CX(I))) + ABS(AIMAG(CX(I)))
30 CONTINUE
    SCASUM = STEMP
    RETURN
    END
```