

INSTRUCTION MANUAL

SSP-10  
DIATHERMY MACHINE

Designed and Built by:

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360517-1-T = RL-2534

INTRODUCTION

The "SSP-10" (Solid State Power--10 watts) is a custom-designed diathermy machine designed for RF/DC hair removal. Operating is initiated by the footswitch, and control is either manual (foot) or automatic (timers). The unit is designed to provide 0-45 volts DC into "any" impedance (with built in current limiting for a maximum of about 6 ma). Automatic squelch is provided by a timer with a range of 0.1 second to 100 seconds. The RF section provides 0-60 volts peak into  $200-300\Omega$  (optimum) or about 9 watts. If current (RF ma) is to be properly monitored, the special four-wire probe (with current sampling resistor) must be used.

Caution Notes

The RF output will sustain "short term" short circuits (a few seconds), but long term shorts at high power will destroy the current sampling resistor and may produce lower lifetime of the power output transistor.

The DC section will sustain long term shorts.

Mixing of RF and DC signals is accomplished by first inserting a DC block capacitor at the tip of the RF probe and connecting the DC output to the tip of the DC block capacitor (see Fig. 1).

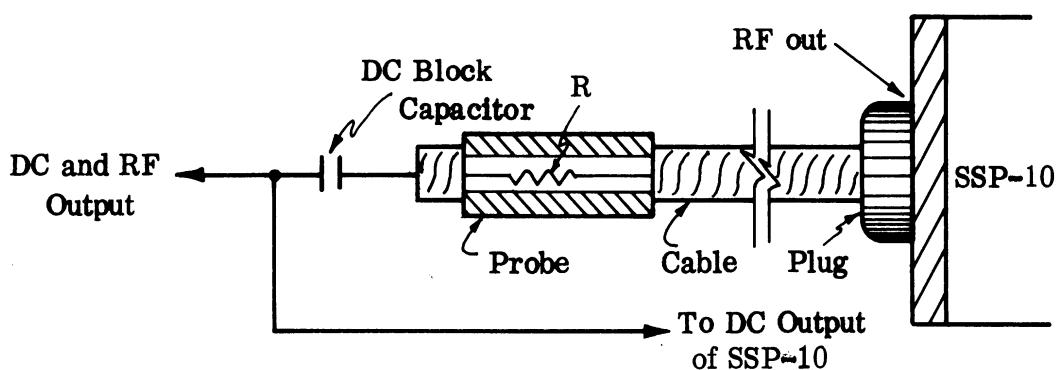


Figure 1

OPERATION

## DC Output

Connect the proper plug and cable to the "DC Output" jack (amphenol plug No. 31-202; Mil. No. UG-88CU). Turn power switch ("line") to on.

If the RF section is not to be used turn "RF trigger" to off. On the DC section, flip the "DC trigger" to "auto" or "man". "Man" will turn the DC on as long as the footswitch is closed. "Auto" will turn the DC on as long as the footswitch is closed, or the self timer runs out, whichever occurs first. Timing is accomplished by setting the "multiplier" to the proper range. ( $x 1$  provides 0.1 to 1.0 second;  $x 10$  is 1.0 to 10.0 seconds;  $x 100$  gives 10.0 to 100.0 seconds.) The "time" dial sets the time within the range as set by the multiplier. So time,  $T$ , is given by:

$$T = (t) \times M ,$$

where       $t$  is time dial setting (continuous from 0.1 to 1.0),

$M$  is multiplier dial setting ( $x 1$ ;  $x 10$ ; or  $x 100$ ).

The DC voltage is varied by the "level" control and is continuous from 0-45 volts. The output impedance is low (100 ohms), until current limiting results (attempt to draw more than 5 ma).

## RF Output

The timing is obtained in the exact manner as outlined in the "DC Output" section above.

RF level control is accomplished by the "RF level" control and can be varied from zero to maximum. The actual RF current (calibrated in RMS) is shown on the "RF MA" meter (display is up to 500 ma RMS). About 300 ma is the maximum available current under any normal (or shorted) loading. The RF current is sampled at the probe tip by a 12 ohm resistor. Shielded

or unshielded loads may be used on the probe. The required plug for connection is an amphenol 126-217. If shielded leads are used the shield to the main RF output buss ("B" terminal on plug) may be grounded (to pin "D") or left floating (pin "C").

#### RF Probe Modification

##### Unshielded Leads

If unshielded leads are used, connect the probe as in the schematic "RF Ammeter and Output Circuit"; pins "D" and "C" may be left open (see Fig. 2).

##### 2 Coax Lines

If 2 miniature coax lines are used, see the schematic and use pin "D" or pin "C" for shield of main ("B") (see Fig. 3).

##### 1 Coax - 1 Balanced Shielded

Connect the coax shield to pin "D". Connect main center conductor to "B". On the balanced line, ground the shield to "D", and connect lines to "A" and "E". This should give maximum reduction of undesirable radiation but some reduction in power output if a high impedance coax ( $300\Omega$ ) is not available (Fig. 4).

#### CIRCUIT DESCRIPTION

The footswitch bounce action is removed by  $R_1$  and  $C_1$  in the Schmidt trigger circuit. The S.T. output drives the timer triggering circuits which properly connect and charge  $C_8$  in the timer circuits.

The timers consist of a squelchable clock (IC 1), a staircase timer (IC 2), and a comparator (Latch (IC 3)). Upon resetting the latch (IC 3), IC 1 produces a square wave of frequency determined by  $C_9$ ,  $10$ , or  $11$ . C 3, 4 and D 2, 3 and IC 2 produce a negative staircase at the "-" input

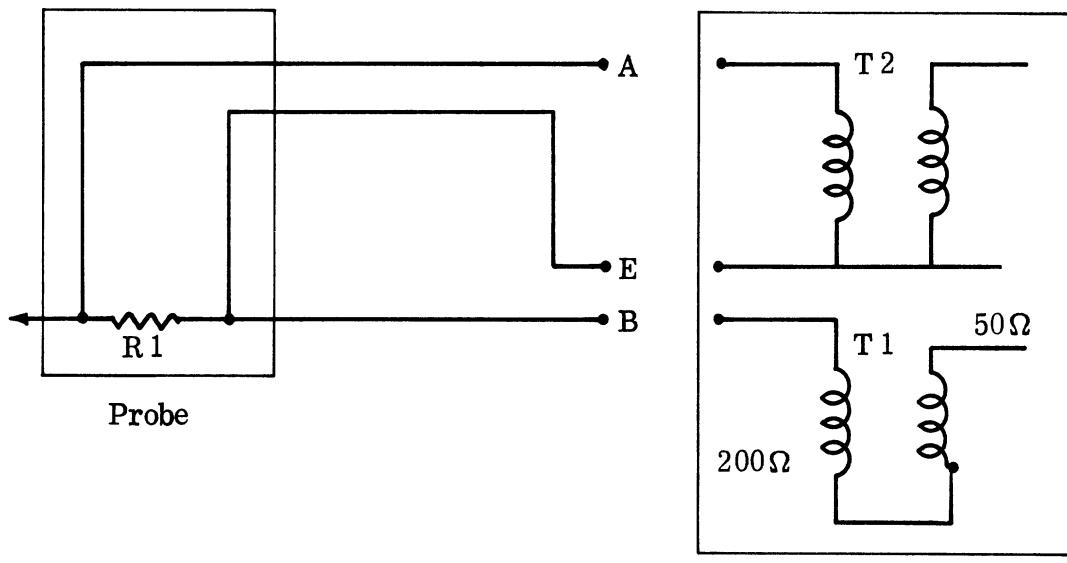


Figure 2: Unshielded Leads

RF Output Circuit

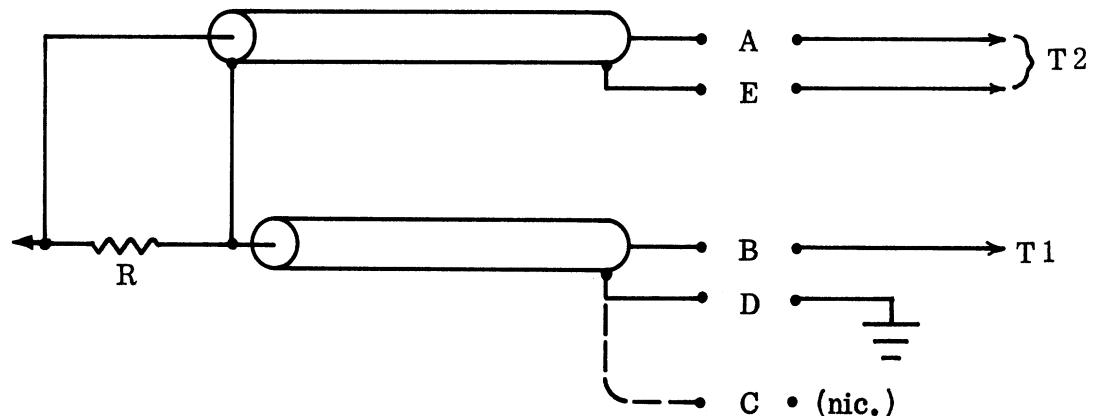


Figure 3: Two Coax Lines

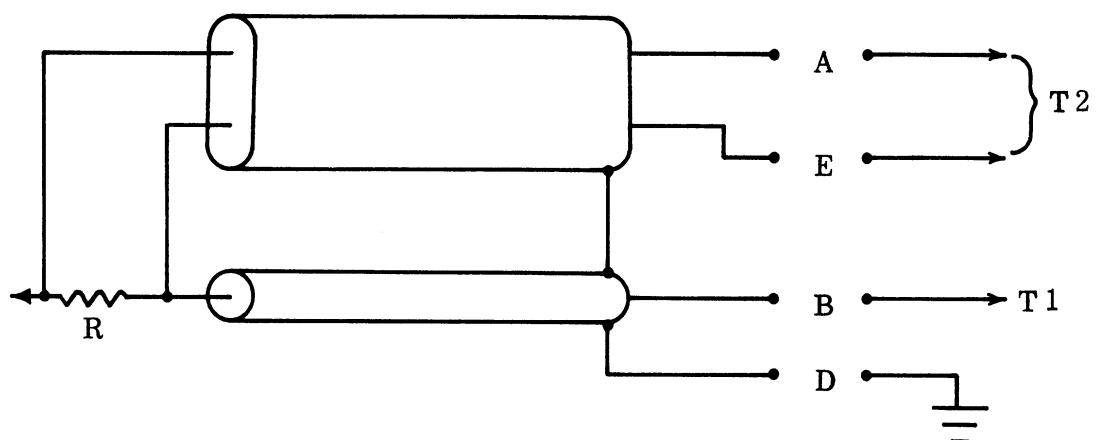


Figure 4: One Coax -- One Balanced Line

of the comparator (IC 3). The "time" dial establishes the voltage and time that IC 3 will flip states. If the voltage at pin 6 (IC 2) is more negative than V<sub>B</sub> at R 14, IC 3 output will go to a positive (high) state. D 5 allows latching action and Q<sub>1</sub> discharges timing integrator C 4.

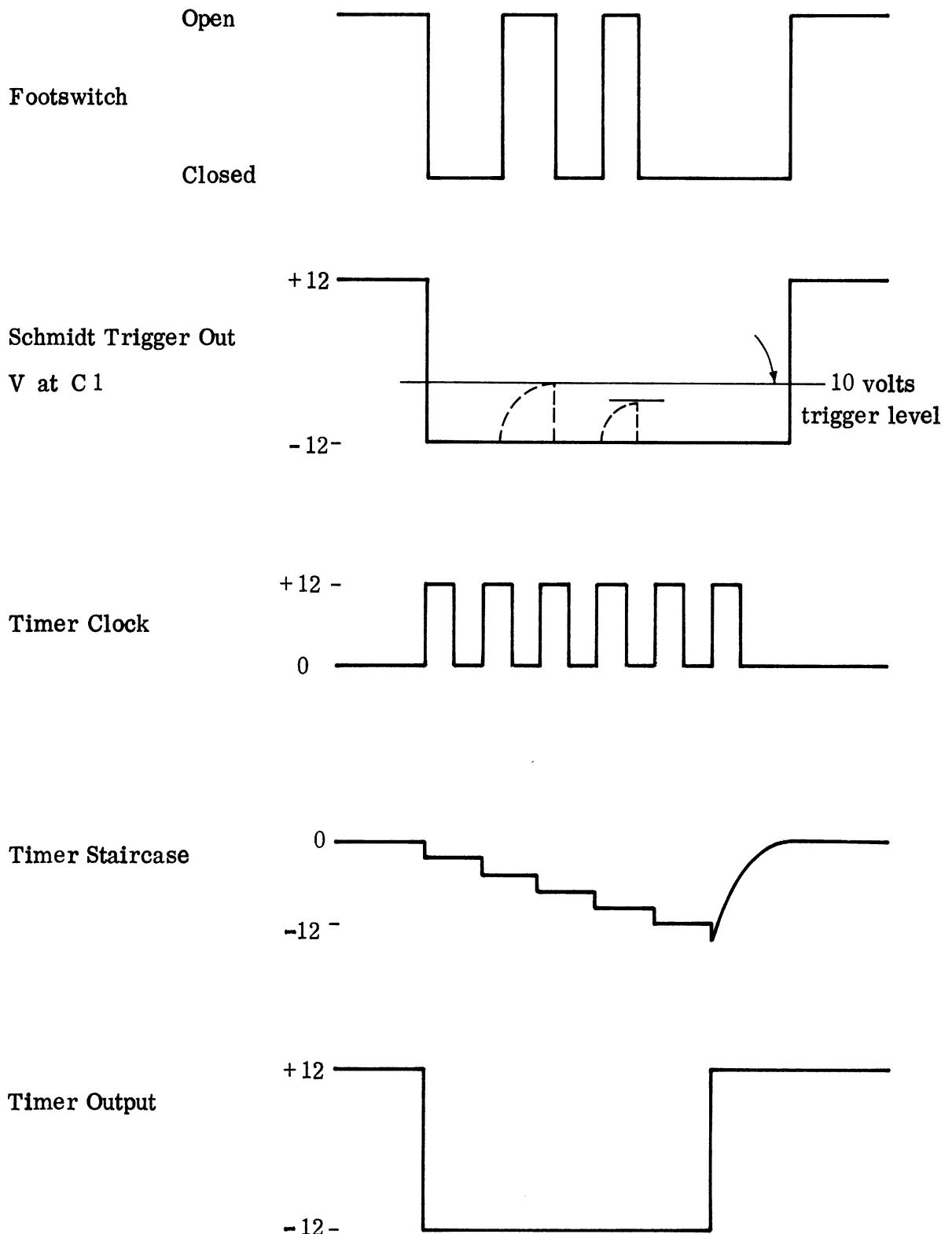
Squelch voltages are derived from the timer or Schmidt-trigger outputs (auto or manual, respectively) and turn off the DC output and RF oscillator when the outputs are high.

The DC power supply voltage is varied by R 24. Current limiting is performed by R 5 and Q 2. Q 3 squelches the DC output.

RF level adjusts the quiescent current in Q 3 and Q 2 in the oscillator, thus providing variable gain and output voltage.

The RF Power Amplifier is courtesy of RCA (RCA Solid State Power Circuits SP-52). It is designed to drive 50-75 ohm loads from 2-30 MHz. Impedance translation to 200 ohms is obtained by T 1 in the RF output circuit. RF output current reading is detected by R 1 and associated transformers and rectifier circuits. The chokes L 1 and L 2 remove RF voltage from the RF MA meter which floats on the RF output.

## TIMER WAVEFORMS



## SSP-10 CALIBRATION

### I. RF Current Meter Calibration

Load =  $330\Omega$  5 percent carbon resistor

RMS Milliamps <sup>(1)</sup>	SSP-10 RF MA Meter
108	100
99	90
89	80
79	70
70	60
59	50
50	40
38	30
28	20
17	10

### II. RF Current Meter Calibration

Load =  $150\Omega$  ( $47 \times 3$  10 percent)

RMS Milliamps <sup>(1)</sup>	SSP-10 RF MA Meter
154	160 (*)
145	150
135	140
126	130
118	120
109	110
100	100
53	50

\* Maximum undistorted current into 150 ohms.

### III. Peak-to-Peak Voltage vs. Level Control

Load = 330Ω

Volts Peak-to-Peak <sup>(2)</sup>	RF Level
102	10
100	9
99	8
95	7
88	6
82	5
68	4
42	3
20	2
4.2	1
0.7	0

- Notes:
- (1) Measured using Hewlett-Packard 8405A vector voltmeter with Tektronics P6022 current probe with termination (in 1 ma/mv position).
  - (2) Measured with Tektronics Model 545B oscilloscope (type "CA" plug in unit) with x 10 attenuator probe (P 6006 - 10 MEG 7pF).

Engineer: C. B. Loftis, Jr.

## SCHMIDT TRIGGER AND TIMER SETUP

$C_1$	.1 $\mu f$ mylar	$Q_{1, 2, 5, 8, 9, 12}$	2N3904
$C_2$	.01 $\mu f$ mylar	$Q_{3, 4, 6, 7, 10, 11}$	2N3906
$C_{3, 4}$	.1 $\mu f$ mylar	$Q_{13, 14}$	
		$D_{1-5}$	1N914
		Lamp Leecraft 14 v 80 ma	

$R_1$	470 K	$R_7$	680
$R_2$	330 K	$R_{8, 12, 15, 19}$	33 K
$R_3$	4.7 K	$R_{9, 16}$	47 K
$R_4$	270 K	$R_{10, 13, 17, 20}$	12 K
$R_5$	82 K	$R_{11, 14, 18, 21}$	18 K
$R_6$	3.9 K	$R_{22, 23}$	6.8 K

## POWER SUPPLY BOARD

$R_{1, 2, 5}$	100 r	$R_{10, 15, 21}$	10 K 1/4 $\omega$ (Turn Put (DC))
$R_3$	1.5 K	$R_{16}$	1 r 5 $\omega$ $\omega\omega$
$R_4$	1 K	$R_{18}$	2.7 K
$R_6$	22 K	$R_{19}$	680
$R_{7, 12}$	10 K	$R_{20}$	27 K
$R_{8, 13, 16}$	5.6 r 1/4 $\omega$	$R_{22}$	.33 2 $\omega$ $\omega\omega$
$R_{9, 14}$	27 K	$R_{23}$	470 2 $\omega$
		$R_{24}$	10 K 2 $\omega$ Put

		Mallery	
$D_{1, 2, 5, 6, 7, 8, 9-12, 13-16}$	M2.5A Power Diode	$C_{4, 6}$	2000 $\mu f$ 25 v Electrolytic
$D_3$	IN914 Switching Diode	$C_{5, 7, 9}$	100 pf Ceramic
$D_4$	47 v 1 $\omega$ Zener	$C_8$	4500 $\mu f$ 50 v Electrolytic
$D_{17}$	12 v 1 $\omega$ Zener	IC 1	3 RCA CA3085
$C_{1, 2}$	150 $\mu f$ 50 v Electrolytic	$Q_{1, 2, 3}$	2N3053
$C_{3, 9}$	.1 $\mu f$ 100 v Mylar	$Q_{4, 5, 8}$	40409 RCA $\omega$ Heat Sink
		$Q_6$	HEP 248 Mot
		$Q_7$	2N3904 Mot

## RF POWER AMPLIFIER

$Q_1$	2N3553	TO-5	$C_1, 2, 19, 8, 12$	.01 $\mu f$	Ceramic
$Q_2$	2N3375	TO-60 (heat sink)	$13, 3$		
$Q_3$	2N5070	TO-60 (heat sink)	$C_{16, 15, 14, 5, 10}$	.1 $\mu f$	Mylar
$R_1$	120		$C_{4, 6}$	.001 $\mu f$	Ceramic
$R_2$	150		$C_{17, 18, 9}$	5 $\mu f$	150 volts
$R_3$	1.2 K		$C_{11}$	.0033 $\mu f$	Ceramic
$R_{4, 6}$	15		$D_{1-3}$		IN3193
$R_5$	2.7				
$R_7$	1.5 K		$L_1$	25 $\mu h$	
$R_8$	10		$L_2$	25 $\mu h$	
$R_{9, 10}$	15		$L_3$	50 $\mu h$	
$R_{11}$	120		$T_1, T_2$	20 turns bifilar wound No. 32 on	
$R_{12}$	820 $1\omega$			$Q_3$	CF105
$R_{13}$	180 $2\omega$		$L_4$	82 $\mu h$	
$R_{14}$	10 $1\omega$				
$R_{15}$	7.5 $5\omega$				
$R_{16}$	5.1 $1\omega$				
$R_{17}$	5.1 $  1\omega$				
$R_{18}$	27				

## 13.56 MHz OSCILLATOR

$R_1$	10 K	1 turn	14 $\omega$ Pot (P.C.)				
$R_2$	10 K		$R_8$	120	$R_{14}$	100	$Q_1$
$R_3$	6.8 K		$R_9$	2.2 K	$R_{15}$	6.8 K	$Q_2$
$R_4$	330		$R_{10}$	33 K	$R_{16}$	10 K 2 $\omega$ Put	$Q_3$
$R_5$	100		$R_{11}$	22 K	$R_{17}$	1 K	$Q_4$
$R_6$	1 K		$R_{12}$	22 K			$Q_5$
$R_7$	4.7 K		$R_{13}$	120 $R_{13}$	$D_1$	IN914	$Q_6$
							2N3393

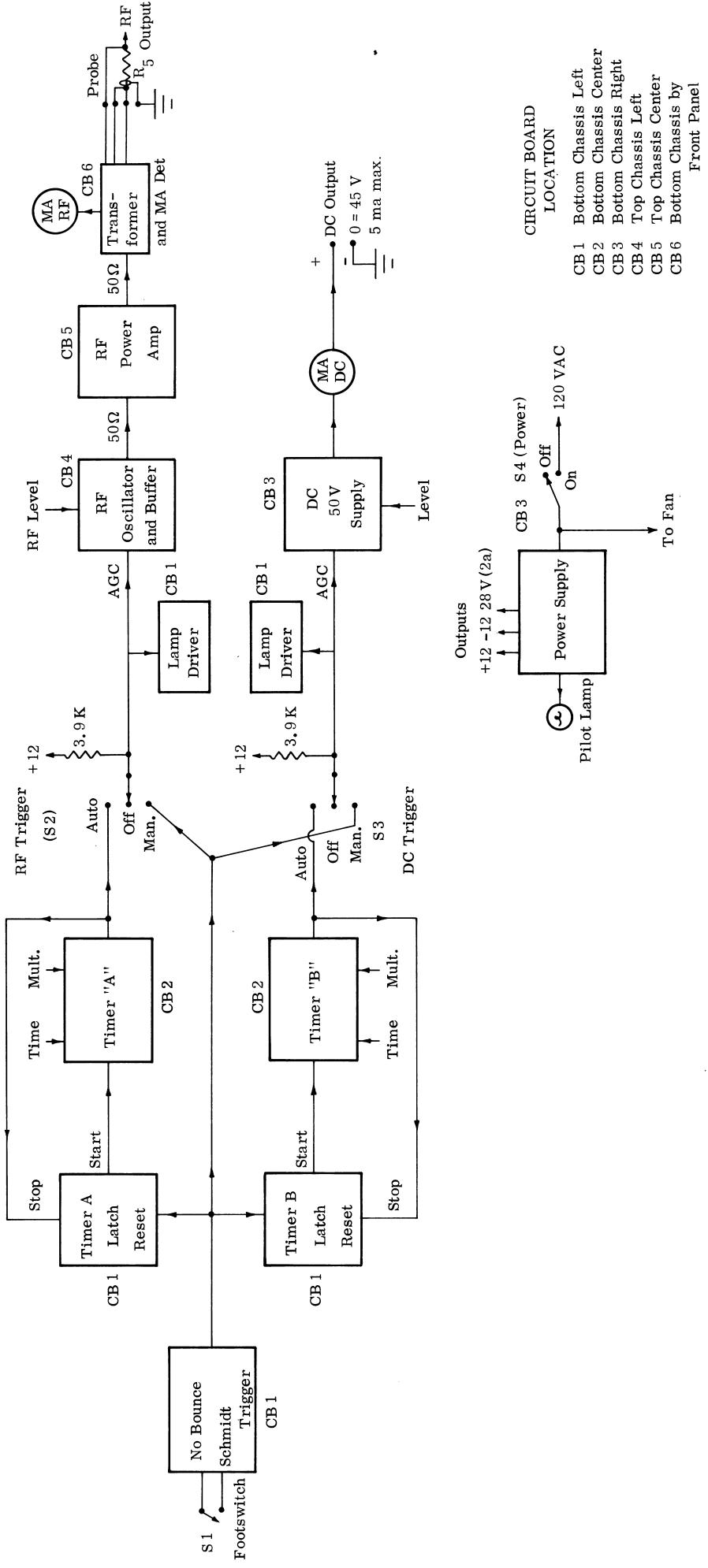
## RF AMMETER AND OUTPUT CIRCUIT

$R_3$	220 $2\omega$ Carbon
$R_1$	12 Current Sampling Resistor
$R_2$	5 K $1/5\omega$ 10 Turn Pot (RF MA Calibrate)
$C_1$	.01 $\mu f$ Ceramic
$C_2$	.1 $\mu f$ Mylar
$C_{3, 4}$	.01 $\mu f$ Ceramic
$L_{1, 2}$	82 $\mu h$ RFC
$T_1$	Approx. 20T Bifilar Wound No. 28 on Q-3 CF105 Core
$T_2$	"Vary-L" 5012.5 A Wide Band Transformer
$T_3$	"Vary-L" 50450 A Wide Band Transformer

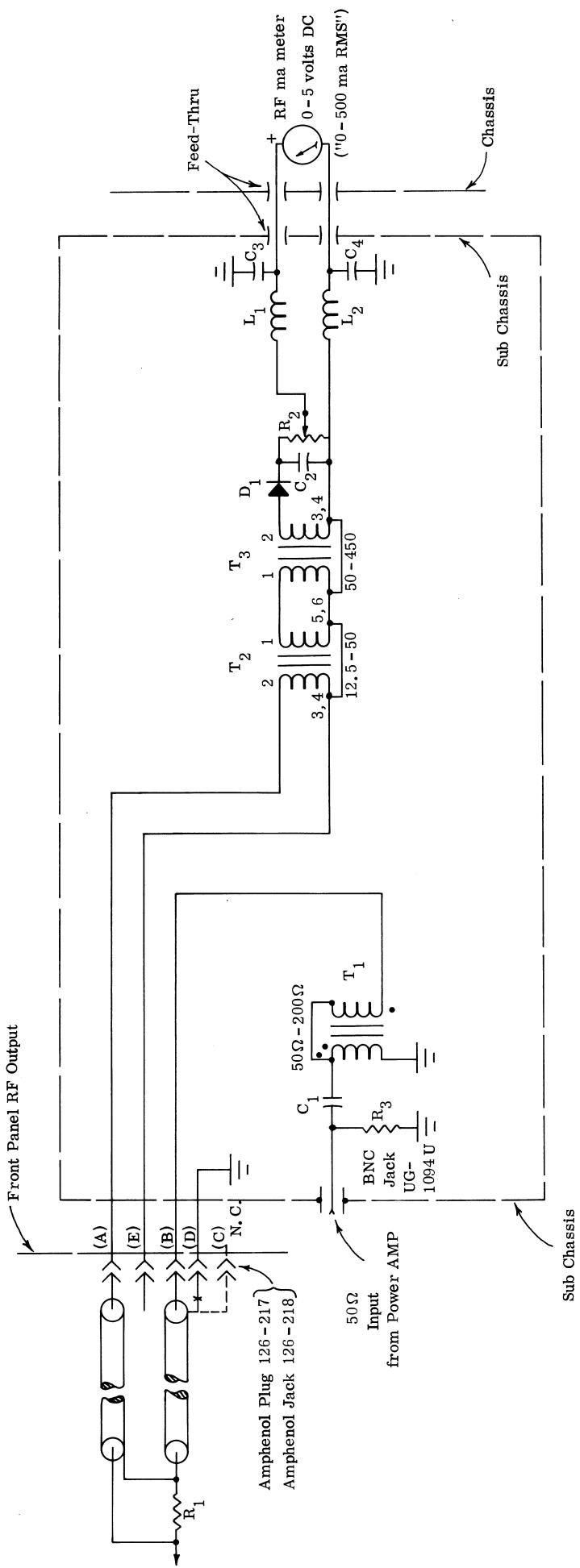
## UNIVERSAL TIMER CIRCUIT (A or B)

$R_{1, 2}$	15 K	$C_{1, 2}$	.001 $\mu f$ Ceramic
$R_3$	18 K	$C_3$	.022 $\mu f$ Mylar
$R_4$	8.2 K	$C_4$	50 $\mu f$ 25 v Electrolytic
$R_5$	10 K 1 turn pot (time calib.)	$C_5$	47 pf CER
$R_{6, 8}$	22 K	$C_6$	.01 $\mu f$ CER
$R_{7, 11}$	1 K	$C_7$	470 PF CER
$R_{9, 10}$	10 K	$C_8$	.1 $\mu f$ Mylar
$R_{12}$	8.2 K	$C_9$	.022 $\mu f$ Mylar (Timing)
$R_{13}$	2.2 K	$C_{10}$	.22 $\mu f$ Mylar (Timing)
$R_{14}$	10 K $2\omega$ 1 turn pot (time)	$C_{11}$	2.0 $\mu f$ Mylar (Timing)
$R_{15}$	1 K	$D_{1-5}$	IN914
		IC 1	RCA CA3047
		IC 2,	3 $\mu a$ 741 or CA3741 CT

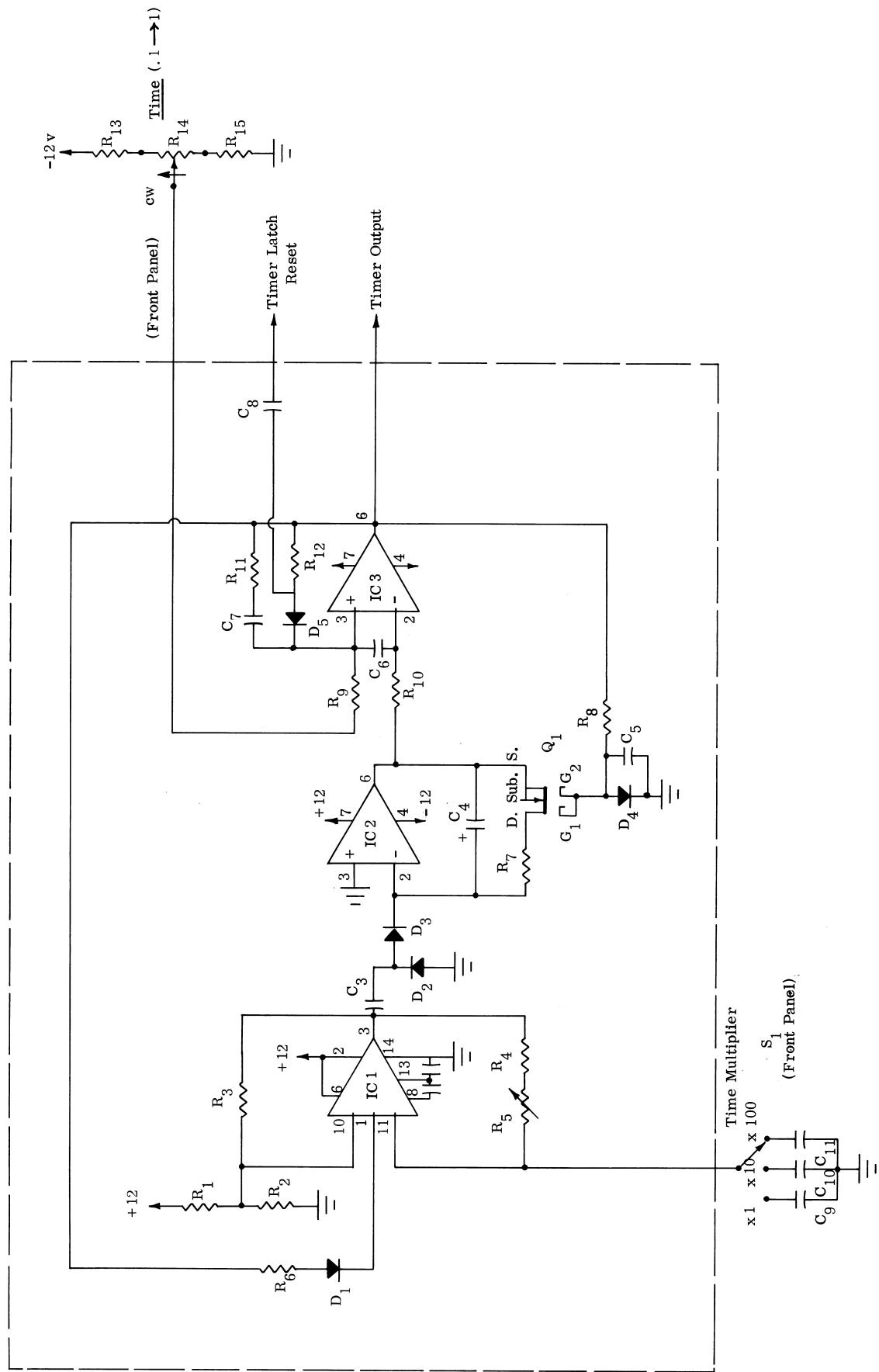
SSP-10 BLOCK DIAGRAM  
(Solid State Power -- 10 Watts)

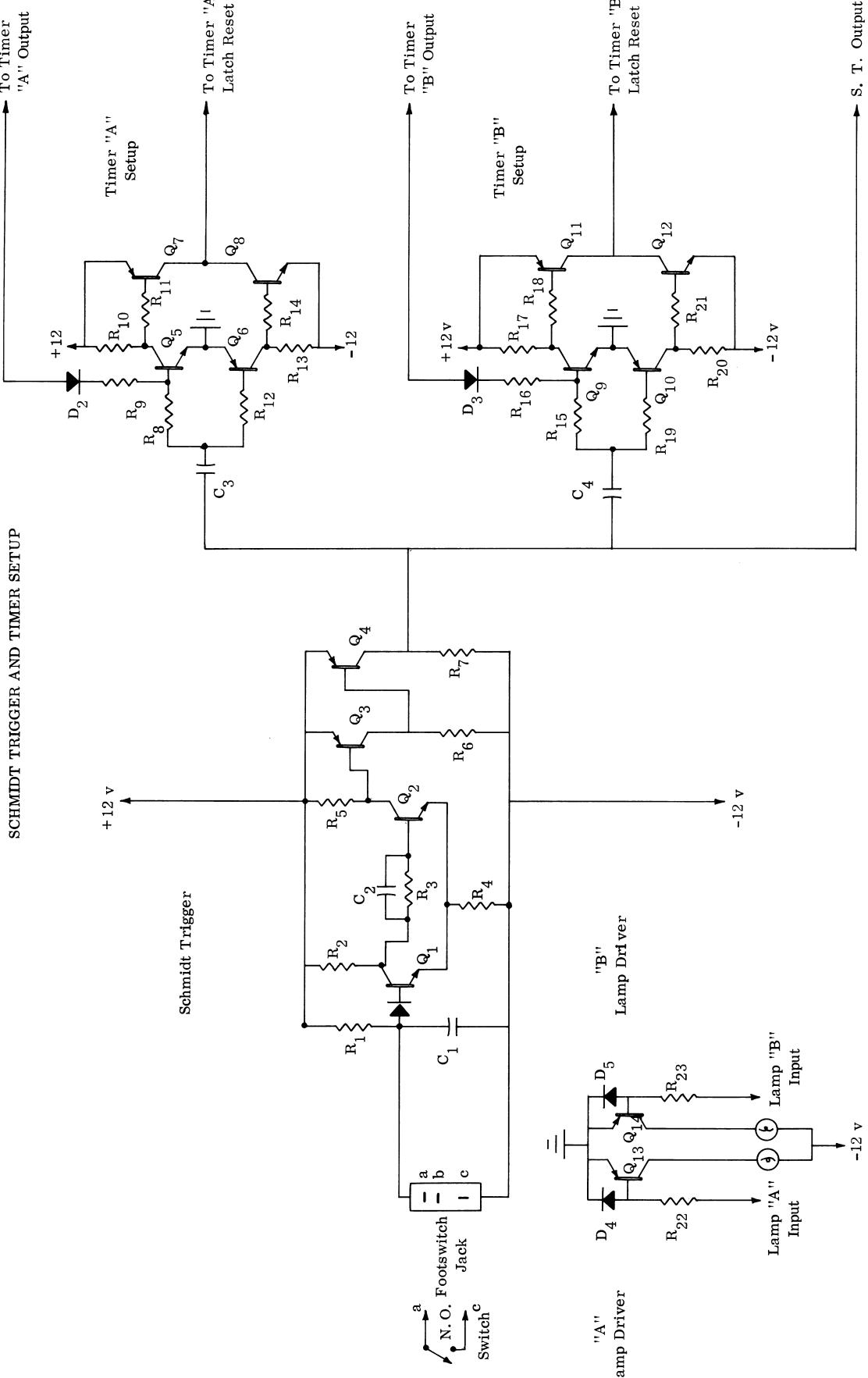


RF AMMETER AND OUTPUT CIRCUIT

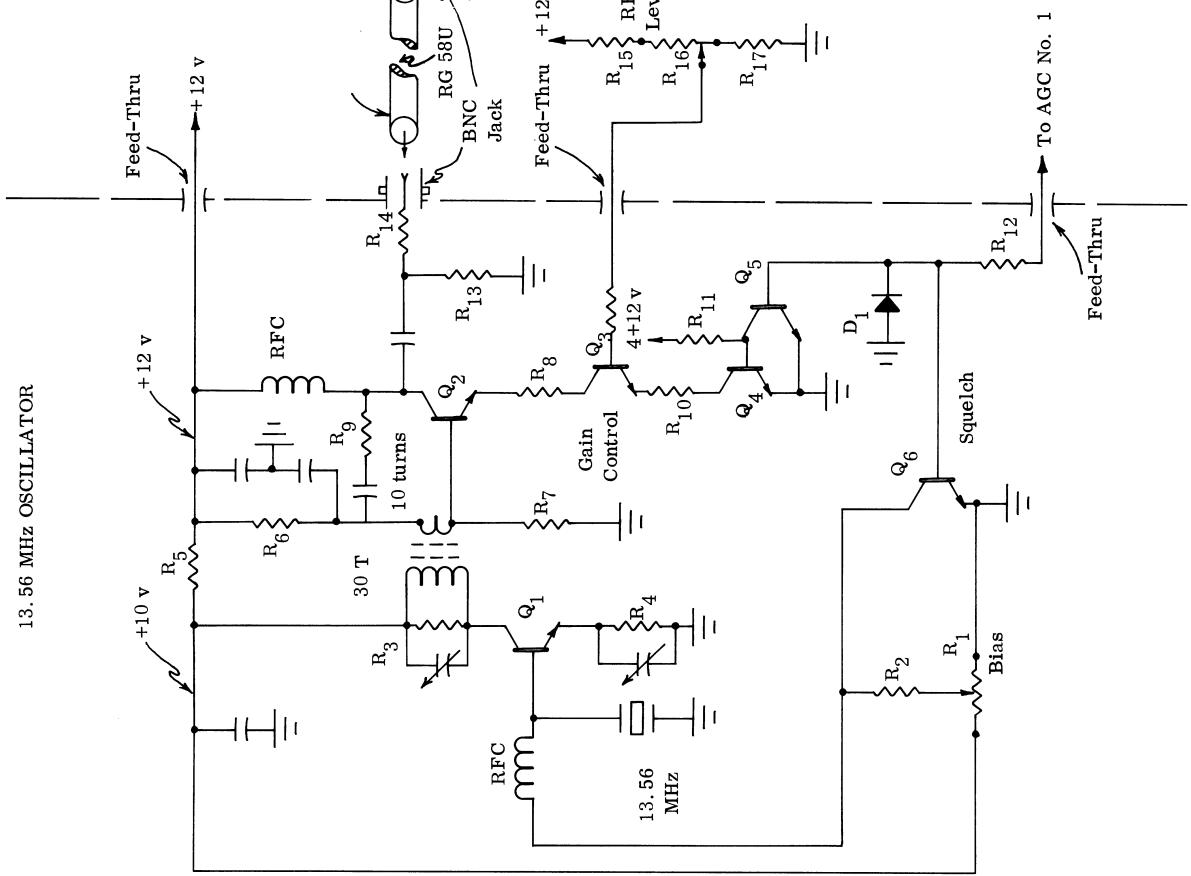


UNIVERSAL TIMER CIRCUIT  
(A or B)





### 13.56 MHz OSCILLATOR



### RF POWER AMPLIFIER

