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METHODS OF IMPROVING
TRANSIENT STABILITY OF A SYNCHRONOUS GENERATOR

Chaman N. Kashkari

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TO SHEILA AND MEERA
# TABLE OF CONTENTS

<table>
<thead>
<tr>
<th>Chapter</th>
<th>Title</th>
<th>Page</th>
</tr>
</thead>
<tbody>
<tr>
<td>List of Symbols</td>
<td></td>
<td>v</td>
</tr>
<tr>
<td>List of Figures</td>
<td></td>
<td>viii</td>
</tr>
<tr>
<td>Abstract</td>
<td></td>
<td>xi</td>
</tr>
<tr>
<td>Chapter I</td>
<td>Scope of Investigation</td>
<td>1</td>
</tr>
<tr>
<td>Chapter II</td>
<td>Mathematical Representation of the System</td>
<td>6</td>
</tr>
<tr>
<td>Chapter III</td>
<td>Excitation Systems - Phase Plane Analysis of Derivative Signals</td>
<td>20</td>
</tr>
<tr>
<td>Chapter IV</td>
<td>Excitation Systems - Digital Computer Simulation</td>
<td>45</td>
</tr>
<tr>
<td>Chapter V</td>
<td>Excitation Systems - Digital Computer Results</td>
<td>49</td>
</tr>
<tr>
<td>Chapter VI</td>
<td>Normalized Stability Margin</td>
<td>58</td>
</tr>
<tr>
<td>Chapter VII</td>
<td>Phase Shifting Transformers - Basic Analysis</td>
<td>64</td>
</tr>
<tr>
<td>Chapter VIII</td>
<td>Application of Phase Shifting Transformers</td>
<td>83</td>
</tr>
<tr>
<td>Chapter IX</td>
<td>Switched Series Capacitors - Approximate Analysis</td>
<td>87</td>
</tr>
<tr>
<td>Chapter X</td>
<td>Switched Series Capacitors - Experimental Study</td>
<td>106</td>
</tr>
<tr>
<td>Chapter XI</td>
<td>Switched Series Capacitors - Fault Study and Relaying</td>
<td>123</td>
</tr>
<tr>
<td>Chapter XII</td>
<td>Conclusions</td>
<td>140</td>
</tr>
<tr>
<td>References</td>
<td></td>
<td>142</td>
</tr>
<tr>
<td>Appendix I</td>
<td>Mathematical Representation of a Synchronous Generator</td>
<td>144</td>
</tr>
<tr>
<td>Appendix II</td>
<td>Computer Programs</td>
<td>158</td>
</tr>
</tbody>
</table>
**List of Symbols**

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>$A_1 - A_6$</td>
<td>Gain in Excitation System Auxiliary Signal Channels</td>
</tr>
<tr>
<td>$E_d$</td>
<td>Direct-Axis Component of Excitation Voltage</td>
</tr>
<tr>
<td>$E'_d$</td>
<td>Direct-Axis Component of Transient Internal Voltage</td>
</tr>
<tr>
<td>$E_{ex}$</td>
<td>Exciter Voltage Referred to Armature Circuit</td>
</tr>
<tr>
<td>$E_i$</td>
<td>Voltage Behind Transient Reactance</td>
</tr>
<tr>
<td>$E_q$</td>
<td>Voltage Behind Synchronous Impedance</td>
</tr>
<tr>
<td>$E'_q$</td>
<td>Voltage Proportional to Field Flux Linkage</td>
</tr>
<tr>
<td>$E_{qd}$</td>
<td>Voltage Behind Quadrature- Axis Synchronous Reactance</td>
</tr>
<tr>
<td>$E_{qo}$</td>
<td>Steady State Value of $E_q$</td>
</tr>
<tr>
<td>$E_x$</td>
<td>Exciter Field Voltage</td>
</tr>
<tr>
<td>$E_1$</td>
<td>Voltage Behind Transient Reactance</td>
</tr>
<tr>
<td>$E_2$</td>
<td>Infinite Bus Voltage</td>
</tr>
<tr>
<td>$G$</td>
<td>Machine Rating</td>
</tr>
<tr>
<td>$H$</td>
<td>Inertia Constant</td>
</tr>
<tr>
<td>$i_d$</td>
<td>Direct-Axis Component of Armature Current</td>
</tr>
<tr>
<td>$i_f$</td>
<td>Current in Machine Field Winding</td>
</tr>
<tr>
<td>$i_q$</td>
<td>Quadrature Axis Component of Armature Current</td>
</tr>
<tr>
<td>$K_1, K_2$</td>
<td>Design Parameters in Auxiliary Signals Channels</td>
</tr>
<tr>
<td>$L_f$</td>
<td>Inductance of Field Winding</td>
</tr>
<tr>
<td>$M$</td>
<td>Inertia Constant of the Machine</td>
</tr>
<tr>
<td>$N$</td>
<td>Slope of the Phase Trajectory</td>
</tr>
</tbody>
</table>
### List of Symbols (Continued)

<table>
<thead>
<tr>
<th>Symbol</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>$P_1$, $P_2$, $P_i$, $P$</td>
<td>Input Power</td>
</tr>
<tr>
<td>$P_{m1}$, $P_{m2}$, $P_{m3}$</td>
<td>Amplitudes of Prefault, During Fault and Post Fault Power Angle Curves</td>
</tr>
<tr>
<td>$P_u$, $P_o$</td>
<td>Output Power</td>
</tr>
<tr>
<td>$r_a$</td>
<td>Armature Resistance</td>
</tr>
<tr>
<td>$r_f$</td>
<td>Resistance of Field Winding</td>
</tr>
<tr>
<td>$S_d$</td>
<td>Direct Axis Saturation Factor</td>
</tr>
<tr>
<td>$S_q$</td>
<td>Quadrature Axis Saturation Factor</td>
</tr>
<tr>
<td>$T'd_o$</td>
<td>Open Circuit Transient Time Constant</td>
</tr>
<tr>
<td>$T_e$</td>
<td>Exciter Field Constant</td>
</tr>
<tr>
<td>$T'q_o$</td>
<td>Quadrature Axis Transient Open-Circuit Time Constant</td>
</tr>
<tr>
<td>$v_b$</td>
<td>Busbar Voltage</td>
</tr>
<tr>
<td>$v_d$</td>
<td>Direct Axis Component of Terminal Voltage</td>
</tr>
<tr>
<td>$v_{dm}$</td>
<td>Direct Axis Component of Machine Terminal Voltage</td>
</tr>
<tr>
<td>$v_{dp}$</td>
<td>Direct Axis Component of Potier Voltage</td>
</tr>
<tr>
<td>$v_m$</td>
<td>Machine Terminal Voltage</td>
</tr>
<tr>
<td>$v_{qm}$</td>
<td>Quadrature Axis Component of Machine Terminal</td>
</tr>
<tr>
<td>$v_{qp}$</td>
<td>Quadrature Axis Component of Potier Voltage</td>
</tr>
<tr>
<td>$v_q$</td>
<td>Quadrature Axis Component of Terminal Voltage</td>
</tr>
<tr>
<td>$v_r$</td>
<td>Reference Voltage</td>
</tr>
<tr>
<td>$\omega$</td>
<td>Frequency in Radians per Second</td>
</tr>
<tr>
<td>Symbol</td>
<td>Description</td>
</tr>
<tr>
<td>--------</td>
<td>-------------</td>
</tr>
<tr>
<td>$x_d$</td>
<td>Direct Axis Synchronous Reactance</td>
</tr>
<tr>
<td>$x'_d$</td>
<td>Direct Axis Transient Reactance</td>
</tr>
<tr>
<td>$x_e$</td>
<td>External Reactance</td>
</tr>
<tr>
<td>$x_p$</td>
<td>Potier Reactance</td>
</tr>
<tr>
<td>$x_q$</td>
<td>Quadrature Axis Synchronous Reactance</td>
</tr>
<tr>
<td>$x'_q$</td>
<td>Quadrature Axis Transient Reactance</td>
</tr>
<tr>
<td>$\delta$</td>
<td>Torque Angle</td>
</tr>
<tr>
<td>$\delta_o$</td>
<td>Steady-State Torque Angle</td>
</tr>
<tr>
<td>$\phi$</td>
<td>Magnetic Flux</td>
</tr>
<tr>
<td>$\mu_v$</td>
<td>Gain in the Voltage Channel</td>
</tr>
<tr>
<td>Figure</td>
<td>Description</td>
</tr>
<tr>
<td>--------</td>
<td>-----------------------------------------------------------------------------</td>
</tr>
<tr>
<td>2.1</td>
<td>Elementary Power System</td>
</tr>
<tr>
<td>2.2</td>
<td>Phasor Diagram of the System</td>
</tr>
<tr>
<td>2.3</td>
<td>Phasor Diagram of the System with Saturation</td>
</tr>
<tr>
<td>2.4</td>
<td>An Exciter for a Synchronous Generator</td>
</tr>
<tr>
<td>3.1</td>
<td>Torque Angle vs. Rate of Change of Torque Angle Curve</td>
</tr>
<tr>
<td>3.2</td>
<td>Effect of Excitation Control on Operating Point</td>
</tr>
<tr>
<td>3.3</td>
<td>Effect of Excitation Control on Torque Angle/Rate of Change of Torque Angle Curve</td>
</tr>
<tr>
<td>3.4</td>
<td>Effect of Excitation Control on Operating Point</td>
</tr>
<tr>
<td>3.5</td>
<td>Construction of a Phase Trajectory using the Method of Isoclines</td>
</tr>
<tr>
<td>3.6a</td>
<td>Effect of $K_2$ on Isoclines $k_2 = .005 \times 180/\pi$</td>
</tr>
<tr>
<td>3.6b</td>
<td>Effect of $K_2$ on Isoclines $k_2 = .01 \times 180/\pi$</td>
</tr>
<tr>
<td>3.6c</td>
<td>Effect of $K_2$ on Isoclines $k_2 = .015 \times 180/\pi$</td>
</tr>
<tr>
<td>5.1</td>
<td>Swing Curves</td>
</tr>
<tr>
<td>5.2</td>
<td>Sending End Terminal Voltage</td>
</tr>
<tr>
<td>5.3</td>
<td>Machine Field Voltage</td>
</tr>
<tr>
<td>5.4</td>
<td>Swing Curve for input 1.4</td>
</tr>
<tr>
<td>5.5</td>
<td>Exciter Field Voltage</td>
</tr>
<tr>
<td>6.1</td>
<td>Determination of Stability Margin</td>
</tr>
<tr>
<td>6.2</td>
<td>Pre-fault, During-fault and Post-fault Power Angle Curves</td>
</tr>
<tr>
<td>Figure</td>
<td>Description</td>
</tr>
<tr>
<td>--------</td>
<td>------------------------------------------------------------------------------</td>
</tr>
<tr>
<td>7.1</td>
<td>Application of P.S. Transformers in a Typical System</td>
</tr>
<tr>
<td>7.2</td>
<td>Effect of Phase Shift on Power Angle Curve</td>
</tr>
<tr>
<td>7.3</td>
<td>Phase-Plane Analysis of Phase Shifting Transformers</td>
</tr>
<tr>
<td>7.4</td>
<td>Determination of the Effect of Phase-Shift on Stability Margin</td>
</tr>
<tr>
<td>7.5</td>
<td>$\theta$ for Dead-Beat Control</td>
</tr>
<tr>
<td>7.6</td>
<td>Determination of Minimum $\theta$</td>
</tr>
<tr>
<td>8.1</td>
<td>A Typical System Using a P.S. Transformer</td>
</tr>
<tr>
<td>8.2</td>
<td>Effect of Phase Shift Transformer on Swing Curve</td>
</tr>
<tr>
<td>9.1</td>
<td>A Typical System Using Switched Series Capacitors</td>
</tr>
<tr>
<td>9.2</td>
<td>Power Angle Curves for the Above System</td>
</tr>
<tr>
<td>9.3</td>
<td>Effects of Switched Series Capacitors on Power Angle Curves</td>
</tr>
<tr>
<td>9.4</td>
<td>Series Compensation for Maximum Transient Stability Limit</td>
</tr>
<tr>
<td>9.5</td>
<td>Determination of Minimum Capacitive Reactance</td>
</tr>
<tr>
<td>9.6</td>
<td>Single Machine System</td>
</tr>
<tr>
<td>9.7</td>
<td>Positive Sequence Networks</td>
</tr>
<tr>
<td>9.8</td>
<td>Effect of Switched Series Capacitor on Swing Curves</td>
</tr>
<tr>
<td>10.1</td>
<td>Schematic Diagram of the Laboratory System</td>
</tr>
<tr>
<td>10.2</td>
<td>Experimental Set Up</td>
</tr>
<tr>
<td>10.3</td>
<td>Experimental Set Up</td>
</tr>
<tr>
<td>Figure</td>
<td>Description</td>
</tr>
<tr>
<td>--------</td>
<td>-------------------------------------------------------</td>
</tr>
<tr>
<td>10.4</td>
<td>Experimental Set Up</td>
</tr>
<tr>
<td>10.5</td>
<td>Experimental Set Up</td>
</tr>
<tr>
<td>10.6</td>
<td>Principle of Torque Angle Measurement</td>
</tr>
<tr>
<td>10.7</td>
<td>Schematic Diagram of the System</td>
</tr>
<tr>
<td>10.8</td>
<td>Phasor Diagram</td>
</tr>
<tr>
<td>10.9</td>
<td>Comparison of Theoretical and Experimental Curves</td>
</tr>
<tr>
<td>10.10</td>
<td>Experimental Recordings</td>
</tr>
<tr>
<td>11.1</td>
<td>A Generalized Circuit</td>
</tr>
<tr>
<td>11.2</td>
<td>A Typical By-pass Circuit</td>
</tr>
<tr>
<td>11.3</td>
<td>Swing Curve with Capacitor Insertion</td>
</tr>
<tr>
<td>11.4(a)</td>
<td>Circuit for Distance Relay Study</td>
</tr>
<tr>
<td>11.4(b)</td>
<td>A Simplified Circuit</td>
</tr>
<tr>
<td>11.5</td>
<td>A Simple Two Machine System</td>
</tr>
<tr>
<td>11.6</td>
<td>Impedance Seen by Relay during Swing</td>
</tr>
<tr>
<td>A-1.1</td>
<td>Two Pole Synchronous Machine</td>
</tr>
<tr>
<td>A-1.2</td>
<td>Y-Connected Synchronous Machine Windings</td>
</tr>
<tr>
<td>A-1.3</td>
<td>Phasor Diagram</td>
</tr>
</tbody>
</table>
ABSTRACT

This work is an investigation of three techniques for enhancing transient stability of a synchronous machine connected to an infinite bus through parallel transmission lines. These techniques are as follows:

1. Application of Auxiliary Signals in Excitation Systems
   This research includes an analytical study of the effects of auxiliary signals used in Excitation Systems. These signals are proportional to derivatives of torque angle. A very interesting conclusion of the study is that if the signals are of correct magnitude, then the oscillations will be strongly damped. The mathematical techniques used in this study can be applied while designing an excitation system. These techniques are based on Phase Plane Analysis.

2. Use of Phase-Shifting Transformers
   Another interesting method for controlling transients utilizes phase-shift windings in step-up transformers. This research is concerned with an analytical study of the effect of phase-shifting transformers on the transient stability of a Synchronous Generator. Since the power is carried by the phase-shift winding during transients only, it can be designed for short time rating with resultant economy.

3. Switched Series Capacitors
   An experimental and an analytical investigation has been made on switched series capacitors. The analytical study includes some application problems associated with the capacitors. The results indicate that it is technically feasible to use switched series capacitors in long transmission lines.
Chapter I

SCOPE OF INVESTIGATION

Introduction

Since the early days of alternating current power generation and transmission, engineers have been faced with the problem of power system stability. During the past few years this problem has assumed new importance due to system interconnections, which have created complex networks having load areas far away from the generating plants. The Northeast blackout of November 9, 1965, created a new awareness of this problem in the minds of the public, the government and the electric utilities. The effects of the blackout were so profound that the utilities cannot afford to have another such occurrence. Engineers must therefore search for new techniques to improve the reliability of electric power supply.

Definitions

According to the American Standards Association\(^1\), stability with reference to power systems is defined as follows:

"Stability when applied to a system of two or more synchronous machines through an electrical network, is the condition in which the difference of angular positions of the machines either remains constant while not subject to a disturbance or becomes constant following an aperiodic disturbance.

"Transitory Stability is a condition which exists in a power system if, after an aperiodic disturbance, the system regains its steady-state stability.

"Steady State Stability is a condition which exists in a power system if it operates with stability when not subjected to an aperiodic disturbance."

\(^1\) American Standards Association
It will be observed that depending upon the nature of the disturbance, stability can be classified in the following categories:

1. Steady-state-stability under slow or gradual load change.
2. Transient-stability—when the system is subjected to sudden and large magnitude disturbance.

By means of an automatic voltage regulator, it is possible to create stability of a generator under conditions for which the generator is inherently unstable. Regulator induced generator stability is referred to as "Dynamic Stability".

Methods of Improving Stability

During system planning, a stability study is required in order to ensure that it will be able to ride through the dynamic oscillations following any system disturbance and maintain stability. The power capability of long, interregional tie lines is usually limited by the transient stability limits. Such economic factors as the high cost of long lines and the additional revenue obtainable from the delivery of additional power provides a strong incentive to explore all economically and technically feasible means of raising the stability limits.

The methods of increasing the stability limits may be classified as follows:

1. Switching
2. Excitation systems
3. Compensation
4. Energy control.
1. **Switching** includes rapid clearing of faults, increasing the number of intermediate switching stations in the case of two or more parallel lines and the use of reclosing breakers. These methods have been traditionally used by power system engineers.

2. **Excitation systems.** Fast excitation systems have considerable effect on stability, particularly dynamic stability. There is however considerable interest in the development of faster excitation systems which would improve both dynamic and transient stability.

3. **Compensation.** By compensation is meant the introduction of series capacitors or shunt reactors.

4. **Energy control.** This includes braking resistors and rapid reduction of input power by fast valve action, on the prime mover.

**Scope of Investigation**

The aim of this research is to develop new techniques for improving transient stability of a synchronous generator, connected to an infinite bus by means of a double circuit transmission line. The following research was carried out:

(1) **Application of Auxiliary Signals**

The main function of an excitation system is to regulate the machine terminal voltage. A voltage signal proportional to the difference between the magnitudes of reference voltage and the
terminal voltage is applied to the field circuit of a control exciter, thereby correcting the terminal voltage. The possibility of using signals other than the terminal voltage as input signals to the regulator has been mentioned recently. These investigators have reported that the transient performance of the synchronous generator is improved if additional signals are applied to the excitation system.

This research is concerned with an analytical study of the fundamental effects of auxiliary signals, proportional to derivative of torque angle $\delta$, on the transient performance of a synchronous generator. The study is based on phase plane techniques of nonlinear control systems. Previous investigations of excitation system problems involved either linearizing of the differential equations or use of empirical methods. By the approach followed in this report, expressions for appropriate signals are developed in a straightforward manner. The gains in various feedback channels can be determined from the knowledge of system parameters. This research lays foundations for future research on excitation system design.

(2) Use of Phase-Shifting Transformers

An interesting method is developed which may have considerable application in long distance transmission of power. This involves the use of auxiliary windings in the sending-end
step-up transformer. This winding provides phase shift. During a transient, power flows through this winding, instead of the normal winding, with resultant damping of the oscillations and enhancement of the stability limit. Since the auxiliary winding carries power during the transient only, it can be short-time rated with resultant economy.

(3) Use of Switched-Series Capacitors

Kimbark$^6$ and Smith$^{12}$ have suggested inserting capacitors in the lines to compensate for the increase in line reactance due to switching out of a line section. The author has extended this idea by an experimental and theoretical study of a typical installation using switched series capacitors. Some application problems, inherent in use of switched series capacitors have been taken into consideration, thus establishing the feasibility of using this technique in modern installations.
Chapter II

MATHEMATICAL REPRESENTATION OF THE SYSTEM

Introduction

This study is based on the classical work of Park\textsuperscript{7} who developed a mathematical model for an ideal synchronous machine. Details of the assumptions on which the model is based and the derivation of the equations are given in Appendix I. A simplifying approximation in the model is that the magnetic saturation has been neglected. In this chapter the machine equations have been suitably modified to account for saturation. This chapter also gives a description of the system under consideration and develops mathematical models for the components so that the system can be represented by a set of differential equations. The equations derived here have been used in Chapters III to V, for digital simulation of the system, to investigate the effects of feeding auxiliary signals to the excitation system.

Description of the System

The elementary power system which has been considered for the purpose of analysis is shown in figure 2.1. It consists of a salient-pole synchronous generator, connected to an infinite bus by a double circuit transmission line. This is a classical configuration for such a study.
Machine Equations and Saturation

The equations of an ideal synchronous machine have been developed in Appendix I. It will now be necessary to modify these equations so that saturation can be taken into account. Many methods, all approximate, have been given in literature to include saturation. The method used here is that outlined by Olive and is based upon a correction of the flux linkages within the machine. It is approximate in that leakage flux has been assumed known for stator circuits only and has been assumed zero for rotor circuits. The saturation factor \( S_d \) or \( S_q \) at any specified voltage for the direct or quadrature axis, is defined as

\[
\frac{\text{Field current to give that voltage on the magnetization curve}}{\text{Field current to give that voltage on the airgap line}} - 1
\]

The direct axis saturation factor may be obtained directly from the machine open circuit curve corresponding to the value of the quadrature axis component of the Potier voltage or at

\[
v_{qp} = v_q + r_a i_q + i_d x_p
\]

where

\[
S_d = f(v_{qp})
\]

\( v_{qp} \) is the quadrature axis component of Potier voltage, \( v_q \) is the quadrature axis component of terminal voltage, \( r_a \) is the armature resistance, \( i_q \) is the quadrature axis component of current, \( i_d \) is the
direct axis component of current, \( x_p \) is the Potier reactance and \( f \) is a symbol for a certain function.

The value of the quadrature axis saturation factor may also be obtained from the same open-circuit curve by multiplying the value found at the quadrature axis component of Potier voltage by the ratio \( x_q/x_d' \), where \( x_q \) and \( x_d \) are quadrature axis and direct axis synchronous reactances of the machine.

\[
v_{dp} = v_d + r_a i_d - x_p i_q
\]

where \( v_{dp} \) is the direct axis component of Potier voltage and \( v_d \) is the direct axis component of terminal voltage.

\[
S_q = \frac{x_q}{x_d} f(v_{dp})
\]

The equations of a synchronous machine as developed in Appendix I are

\[
v_q = -r_a i_q - x_d i_d + E_q \tag{2.1}
\]

\[
v_d = -r_a i_d + x_q i_q + E_d \tag{2.2}
\]

\[
E'_q = -(x_d - x_d') i_d + E_q \tag{2.3}
\]

\[
E'_d = (x_q - x_q') i_q + E_d \tag{2.4}
\]

\[
\frac{d}{dt} E'_q = \frac{1}{T'_{do}} [E_{ex} - E_q] \tag{2.5}
\]

\[
\frac{d}{dt} E'_d = -\frac{E_d}{T'_{qo}} \tag{2.6}
\]
In the above equations, $E'_q$ and $E'_d$ are the quadrature and direct axis components of excitation voltage. $E'_q'$ and $E'_d'$ are the quadrature and direct-axis components of transient internal voltage. $x'_q$ and $x'_d$ are the quadrature and direct axis transient reactances. $T'_{qo}$ and $T'_{do}$ are the quadrature and direct axis open circuit time constants. $E_{ex}$ is the exciter voltage referred to the armature circuit.

Since the machine under consideration is a salient pole synchronous machine ($x'_q = x'_q$), $E'_d = E_d = 0$ and if the resistance terms are neglected, the equations reduce to

\begin{equation}
    v_q = E'_q - i_d x'_d
\end{equation}

\begin{equation}
    v_d = x'_q i_q
\end{equation}

\begin{equation}
    E'_q = E'_q - (x'_d - x'_d') i_d
\end{equation}

\begin{equation}
    \frac{d E'_q}{dt} = \frac{[E_{ex} - E'_q]}{T'_{do}}
\end{equation}

If saturation is included, the equations modify to the following

\begin{equation}
    v_q = E'_q - \left[ -\frac{x'_d - x'_p}{1 + S_d} + x'_p \right] i_d
\end{equation}

\begin{equation}
    v_d = \left[ -\frac{x'_q - x'_p}{1 + S_d} + x'_p \right] i_q
\end{equation}

\begin{equation}
    E'_q = E'_q - \frac{(x'_d - x'_d')}{(1 + S_d)} i_d
\end{equation}
Fig. 2.2 Phasor Diagram of the System
\[
\frac{d E'_q}{dt} = \frac{1}{T_{do}} [E_{ex} - (1+S_d)E_q]
\]  

(2.14)

A phasor diagram showing the above situation is shown in figure 2.3.

Power Output Relationships (Neglecting Saturation)

The power output from a salient pole machine during a transient is given by

\[
P_u = \frac{v_b E'_q}{x'_d + x_e} \sin \delta + v_b^2 \frac{(x'_d - x_q)}{2(x_q + x_e)(x'_d + x_e)} \sin2\delta
\]  

(2.15)

where \(v_b\) is the busbar voltage, \(\delta\) is the angle between the infinite bus and the quadrature axis of the rotor and \(x_e\) is the external reactance.

Since the rate of change of \(E'_q\) involves \(E_q\), it would be convenient to relate \(E_q\) and \(E'_q\) by certain known quantities as follows. Refer to figure 2.2

\[
E_{qd} - i_d(x_q + x_e) = v_b \cos \delta
\]  

(2.16)

where \(E_{qd}\) is the voltage behind the quadrature-axis synchronous reactance. Therefore

\[
i_d = \frac{E_{qd} - v_b \cos \delta}{(x_q + x_e)}
\]  

(2.17)

But

\[
E'_q = E_{qd} - i_d(x_q - x'_d)
\]  

(2.18)

Substituting equation (2.17) in equation (2.18),
Fig. 2.3 Phasor Diagram of the System with Saturation
\[ E'_q = E_{qd} - \frac{E_{qd} \cdot v_b \cos \delta}{x_q + x_e} (x_q - x'_d) \]
\[ = E_{qd} \frac{(x'_d + x'_e)}{x_q + x_e} + v_b \cos \delta \frac{(x_q - x'_d)}{(x_q + x_e)} \]  
(2.19)

Again from the figure,

\[ E_q = E_{qd} + i_d (x_d - x_q) \]  
(2.20)

and

\[ E_{qd} - E'_q = i_d (x_q - x'_d) \]  
(2.18)

From the above equation,

\[ i_d = \frac{E_{qd} - E'_q}{x_q - x'_d} \]

Substituting this expression for \( i_d \) in eq. (2.20), the result is

\[ E_q = E_{qd} \frac{(x_d - x'_d)}{(x_q - x'_d)} - E'_q \frac{(x_d - x_q)}{(x_q - x'_d)} \]  
(2.21)

From (2.19) and (2.21),

\[ E_q = E'_q \frac{(x'_d + x'_e)}{(x'_d + x_e)} - v_b \cos \delta \frac{(x_q - x'_d)}{(x'_d + x_e)} \]  
(2.22)

During a stability study, it will be necessary to use the equation involving rate of change of \( E'_q \), namely

\[ \frac{d}{dt} E'_q = \frac{1}{T'_do} \left[ E_{ex} - E'_q \right] \]  
(2.5)
Since \( E_q \) is known from eq. (2.22), it will be necessary to know \( E_{ex} \), the machine field voltage, which is also the exciter armature voltage. It is therefore, necessary to develop a mathematical model for the exciter as will be shown in the next section.

Exciter Representation

Figure 2.4 represents an exciter with a separately excited field winding. \( E_x \) is voltage applied to the field and \( E_{ex} \) is its armature voltage. Usually the machine voltage is regulated by controlling the exciter field voltage \( E_x \).

If \( r_f \), \( L_f \) are the resistance and the inductance of the exciter field winding and \( i_f \) is the field current, then the following relationships hold.

\[
\begin{align*}
    r_f i_f + L_f \frac{di_f}{dt} &= E_x \\
    E_{ex} &= k \phi = k'i_f
\end{align*}
\]

(2.23) (2.24)

where \( \phi \) is the magnetization flux and \( k \) and \( k' \) are constants. The flux \( \phi \) is proportional to \( i_f \), as saturation in the exciter field is neglected. This is an acceptable approximation, since the ceiling voltage of modern exciters is many times the normal voltage.

From eqs. (2.23) and (2.24)

\[
i_f = \frac{E_x}{r_f + p L_f}, \quad p \text{ is operator } \frac{d}{dt}
\]
Fig. 2.4 An Exciter for a Synchronous Generator
\[ E_{ex} = \frac{k'E_x}{r_f + pL_f} = \frac{k'E_x}{r_f(1 + pT_e)}, \quad T_e \text{ is time constant of the field winding} \]

If \( E_{ex} \) and \( E_x \) are in per unit system, then

\[ E_{ex} = \frac{E_x}{1 + pT_e} \quad (2.25) \]

or

\[ \frac{dE_{ex}}{dt} = \frac{1}{T_e} (E_x - E_{ex}) \quad (2.26) \]

\( E_x \), the exciter field voltage will consist of a steady state component and a transient component proportional to the difference between terminal voltage and reference voltage, or

\[ E_x = E_{xo} - \mu_v(v_m - v_r) \quad (2.27) \]

where \( \mu_v \) is gain in voltage channel. From eq. (2.26) and eq. (2.5) the steady state component of \( E_x \) is equal to \( E_{qo} \), the steady-state value of \( E_q \). In any excitation control with auxiliary signals eq. (2.27) will be

\[ E_x = E_{qo} - \mu_v(v_m - v_r) + f[\text{auxiliary signals}] \quad (2.28) \]

where \( f \) is a certain function.

**Terminal Voltage**

In eqs. (2.27) and (2.28), \( v_m \) is the machine terminal voltage and it is necessary to find an expression for \( v_m \) in terms of the known quantities. Referring to figure 2.2,

\[ v_{dm} = v_b \sin \delta \frac{x_q}{x_q + x_e} \quad (2.29) \]
\[ v_{qm} = v_b \cos \delta + i_d x_e \]  \hspace{1cm} (2.30)

But

\[ i_d = \frac{E_q - v_b \cos \delta}{x_e + x_d} \]  \hspace{1cm} (2.31)

Therefore

\[ v_{qm} = v_b \cos \delta \frac{x_d}{x_e + x_d} + E_q \frac{x_e}{x_e + x_d} \]  \hspace{1cm} (2.32)

\[ v_m^2 = v_{dm}^2 + v_{qm}^2 \]  \hspace{1cm} (2.33)

**Machine Equations with Saturation**

Referring to figure 2.3, the following equations hold

\[ E_q = E'_q + \frac{(x_d - x'_d)}{1 + S_d} i_d \]

\[ v_b \cos \delta = E'_q - i_d (x'_d + x_e) \]

or

\[ i_d = \frac{E'_q - v_b \cos \delta}{(x'_d + x_e)} \]

\[ E_q = E'_q + \frac{(x_d - x'_d)}{(1 + S_d)} \left[ \frac{E'_q - v_b \cos \delta}{x'_d + x_e} \right] \]

\[ E_q = E'_q \left[ 1 + \frac{x_d - x'_d}{(x'_d + x_e)(1 + S_d)} \right] \]

\[ - \frac{v_b \cos \delta (x_d - x'_d)}{(x'_d + x_e) (1 + S_d)} \]  \hspace{1cm} (2.34)

\[ \frac{v_{dm}}{v_b \sin \delta} = \frac{x_p}{1 + S_q} + \frac{x_q - x_p}{x_e + x_p + x_q - x_p} \frac{1}{1 + S_q} \]
\[ v_{dm} = \frac{x_p S_q + x_q}{(x_e + x_p) S_q + x_q + x_e} v_b \sin \delta \]  

(2.35)

Similarly

\[ v_{qm} = v_b \cos \delta + i_d x_e \]

But

\[ i_d = \frac{E_q - v_b \cos \delta}{x_e + x_p + x_d - x_p} \frac{1}{1 + S_d} \]

Therefore

\[ v_{qm} = v_b \cos \delta + \frac{E_q - v_b \cos \delta}{x_e + x_p + x_d - x_p} x_e \frac{1}{1 + S_d} \]

\[ = \frac{E_q x_e}{x_e + x_p + x_d - x_p} + v_b \cos \delta \left[ 1 - \frac{x_e}{x_e + x_p + x_d - x_p} \right] \frac{1}{1 + S_d} \]

\[ v_m^2 = v_{dm}^2 + v_{qm}^2 \]  

(2.36)

(2.33)

Similarly

\[ \frac{d}{dt} E'_q = \frac{1}{T'_d} \left[ E_{ex} - (1 + S_d) E_q \right] \]  

(2.37)

The equations derived in this chapter represent the mathematical model for digital simulation of the synchronous generator.
Chapter III

EXCITATION SYSTEMS

(PHASE PLANE ANALYSIS OF DERIVATIVE SIGNALS)

Introduction

The main function of an excitation system is to regulate the machine terminal voltage. A voltage signal proportional to the terminal voltage is compared with a reference voltage and the error is amplified and fed to the exciter field. Besides regulating the terminal voltage, an excitation system will have a beneficial effect on the stability of a synchronous machine. It will improve dynamic stability to an appreciable degree and transient stability to some extent by rapidly increasing the airgap flux following a fault which produces a depression of machine terminal voltage.

During the last few years many researchers have worked on the problem of improving transient stability by making improvements in the excitation systems\textsuperscript{2−5}. They have attempted to feed auxiliary signals, besides the voltage deviation, to the excitation system. Notable among them are Langer et al\textsuperscript{3} who have used linearized equations on an analog computer, feeding first and second derivatives of load angle to the regulator. Dineley et al\textsuperscript{4} have used a digital computer to find the effect of varying gains in load angle channels.

This research is concerned with an analysis of nonlinear equations of a synchronous machine by the phase plane method and lays the
foundations for a systematic study and design of amplification factors in derivative channels.

The Phase Plane

The phase plane approach is a very convenient method of analyzing some non-linear systems. Consider the equation

$$A'_n \frac{d^n x}{dt^n} + A'_{n-1} \frac{d^{n-1} x}{dt^{n-1}} + \ldots + A'_1 x + A'_0 x = 0$$

(3.1)

If \(x, \dot{x}, \ddot{x}, \ldots, \frac{d^{n-1} x}{dt^{n-1}}\) are treated as co-ordinates of a certain phase space, then this equation can be represented by a single curve in the space. This curve is called a phase trajectory. In second order systems, it is fairly easy to obtain a phase trajectory.

In the above equation \(A'_n\) is the coefficient of the \(n\)th derivative.

If the differential equation is of second order, the equation may be

$$\ddot{x} + A\dot{x} + Bx = F.$$

A phase plane applied to this equation will have coordinates \(x\) and \(\dot{x}\).

The Swing Equation

The equation governing the oscillations of the salient-pole synchronous machine rotor is

$$M \frac{d^2 \delta}{dt^2} = P_i - \left[ \frac{E'_q v_b}{X} \sin \delta - k \sin 2\delta \right]$$
where

\[ M = \text{inertia constant} \quad E'q = \text{voltage proportional to field flux linkages} \]
\[ \delta = \text{torque angle} \quad v_b = \text{infinite bus voltage} \]
\[ t = \text{time} \quad X = x'_d + x_e \quad \text{where} \]
\[ P_i = \text{input power} \quad x'_d = \text{transient reactance} \]
\[ x_e = \text{external reactance} \]

\[ k = \frac{v_b^2}{x_q} \frac{x_q - x'_d}{2(x'_d + x_e)(x_q + x_e)} = \text{a constant, if} \ x'_d \ \text{is assumed to be constant.} \]

Let \[ \omega = \frac{d\delta}{dt} \]

\[ \frac{d\omega}{dt} = \frac{1}{M} \left[ P_i - \frac{E'q v_b}{X} \sin \delta + k \sin 2\delta \right] \]

or

\[ \frac{d\delta}{d\omega} = \frac{\omega M}{P_i - \frac{E'q v_b}{X} \sin \delta + k \sin 2\delta} \quad (3.3) \]

If \( E'q, v_b, X \) and \( M \) are constants, then the variables in eq. (3.3) can be separated to yield

\[ M\omega d\omega = \left[ P_i - \frac{E'q v_b}{X} \sin \delta + k \sin 2\delta \right] d\delta \]

Integrating both sides

\[ \frac{M \omega^2}{2} = P_i (\delta - \delta_0) + \frac{E'q v_b}{X} (\cos \delta - \cos \delta_0) \]
\[ - \frac{k}{2} [\cos 2\delta - \cos 2\delta_0] \]
where initial values of \( \delta \) and \( \omega \) are \( \delta_0 \) and zero respectively.

Therefore

\[
\omega^2 = \frac{2}{M} \left[ P_1 (\delta - \delta_0) + \frac{E'_q v_b}{X} (\cos \delta - \cos \delta_0) - \frac{k}{2} (\cos 2\delta - \cos 2\delta_0) \right]
\]

(3.4)

A phase plane plot was drawn for a synchronous machine system having the following constants and initial conditions.

\[ x_d = 1.25 \quad x'_d = .28 \quad v_b = 1.0 \quad x_q = .70 \quad x_e = .5 \]

\( H \), the inertia constant = 3.0 \( G \), the machine rating = 1.0

\[ M, \text{ the inertia constant} = \frac{GH}{\pi f} = \frac{1.0 \times 3.0}{180 \times 60} = .000278 \]

Input power = 1.0 \( \omega \), the initial steady state torque angle = 50.2°

With reference to figure 2.2, the value of \( E'_q \) is calculated as follows:

\[
E'_q = E_{qd} - (x_q - x'_d) i_d
\]

\[
\bar{E}_{qd} = \bar{v}_b + j(x_e + x_q) \bar{i}
\]

\[ = 1.00 + j(0.5 + 0.7) 1.00 = 1.00 + j 1.2 = 1.57 \]

\[ i_d = i \sin \delta_0 = .765 \]

Therefore

\[ E'_q = 1.57 - (.7 - .28) .765 = 1.245 \]

The phase plane plot is shown in figure 3.1 when the input power was suddenly increased from 1.0 to 1.3. From the plot it
Fig. 3.1 Torque angle/Rate of change of torque angle curve
is observed that the speed of the machine increases rapidly at the beginning of the transient and it reaches a maximum at an angle of about 70°, when it begins to decrease. The oscillations are from 50.2° to 87.0° and since damping has been neglected, these are continuous.

**Auxiliary Signals**

During the preceding analysis, it was assumed that the voltage behind transient reactance, \( E'_q \) was constant and figure 3.1 was drawn on the same basis. In this section, the influence of varying this voltage will be taken into account as follows:

**Signal Proportional to Torque Angle**

Assume that the voltage \( E'_q \) was regulated according to the following relationship

\[
E'_q = E'_{q0} + K_1(\delta - \delta_0) \quad (3.5)
\]

where \( \delta_0 \) is the steady state value of \( \delta \) before the transient occurred, \( E'_{q0} \) is the initial value of \( E'_q \) and \( K_1 \) is a constant, the value of which will be determined. \( \delta \) and \( \delta_0 \) are in radians.

Substituting (3.5) in (3.3) gives

\[
\frac{d\delta}{d\omega} = \frac{\omega M}{P_i - \frac{v_b}{X} [E'_{q0} + K_1(\delta - \delta_0)] \sin \delta + k \sin 2\delta} \quad (3.6)
\]

The effect of this signal is depicted in figure 3.2 which shows that the amplitude of the power angle curves increases from instant to instant due to action of the auxiliary signal. The operating point
Fig. 3.2 Effect of Excitation Control on operating point
moves from a to d through b and c . . It is observed that the amplitude of the oscillation is decreased from \( \delta_1 \) to \( \delta'_1 \).

Again separating the variables in eq. (3.6)

\[
M_\omega \delta \omega = [ P_1 - \frac{v_b}{X} ] [ E' q_0 + K_1 (\delta - \delta'_o) ] \sin \delta + k \sin 2\delta \] d\delta
\]

Integrating this equation, the result is

\[
\frac{\omega^2}{2} = P_1 (\delta - \delta'_o) + \frac{E' q_0 v_b}{X} \left( \cos \delta - \cos \delta'_o \right)
\]

\[
+ \frac{K_1 v_b}{X} \left[ (\delta - \delta'_o) \cos \delta + (\sin \delta'_o - \sin \delta) \right] - \frac{k}{2} (\cos 2\delta - \cos 2\delta'_o)
\]

Comparing equations (3.4) and (3.7), the extra term in (3.7) is

\[
\frac{K_1 v_b}{X} \left[ (\delta - \delta'_o) \cos \delta + (\sin \delta'_o - \sin \delta) \right]
\]

For positive value of \( K_1 \), this expression will have a negative value, hence this type of control will decrease the rotor speed and it will also decrease the angular swing. A phase plane plot for this condition is shown in figure 3.3 for a value of \( K_1 = 0.01 \times \frac{180}{\pi} \).

From the above analysis it is clear that if a signal proportional to the rotor angle is utilized, such that the voltage behind the transient reactance is changed, according to eq. (3.5), a substantial improvement in the reduction of transient oscillations will result. The improvement will depend on the value of \( K_1 \) which will have to be carefully chosen.
Fig. 3.3 Effect of Excitation Control on torque angle/rate of change of torque angle curve
Signal Proportional to Rate of Change of Torque Angle

Assume now that a signal proportional to the rate of change of torque angle was used in such a way that

\[ E'_q = E'_q_0 + K_2 \frac{d\delta}{dt} \]

in which \( K_2 \) is a design parameter. \( \frac{d\delta}{dt} \) is in radians/sec.

A great advantage of this type of control is that is is possible to design the parameter \( K_2 \) such that, for maximum value of transmitted power, the transient is controlled in a dead-beat non-oscillatory fashion. This can be readily observed from figure 3.4. At point a', the value of \( \frac{d\delta}{dt} = 0 \) but very soon \( \frac{d\delta}{dt} \) becomes positive, and the amplitude of the power angle curve continues to increase rapidly and the operating point a' moves along through b', c', d' and reaches e' which is the new operating point. It should be kept in mind that the maximum value of \( \frac{d\delta}{dt} \) is at d' and while above the \( P_i \) line, it steadily becomes less. Hence if the value of \( K_2 \) is chosen properly, the movement from a' to e' can be non-oscillatory.

Consider eq. (3.3) again and substitute the new value for \( E'_q \)

\[
\frac{d\delta}{d\omega} = \frac{\omega M}{P_i - \frac{V_b}{X} [E'_q + K_2 \frac{d\delta}{dt}] \sin \delta + k \sin 2\delta}
\]
or

\[
[k \sin 2\delta + P_i - \frac{V_b}{X} [E'_q + K_2 \frac{d\delta}{dt}] \sin \delta] \frac{d\delta}{d\omega} = M \omega \frac{d\omega}{d\delta}
\]

since

\[
\frac{d\delta}{dt} = \omega
\]
Fig. 3.4 Effect of Excitation Control on operating point
\[ [k \sin 2\delta + P_i - \frac{v_b}{X} [E'q_0 + K_2\omega] \sin \delta] \, d\delta = M \omega d\omega \]

or

\[
\frac{d\omega}{d\delta} = \frac{P_i - \frac{v_b}{X} [E'q_0 + K_2\omega] \sin \delta + k \sin 2\delta}{M \omega} = [A - B [C + K_2\omega] \sin \delta + k' \sin 2\delta] \frac{1}{\omega} \tag{3.8}
\]

where \(A, B, C\) and \(k'\) are constants.

The variables of this equation cannot be separated and hence another method will be used to draw the phase trajectory.

**The Method of Isoclines**

The method of isoclines can be illustrated by considering the following equation

\[
\frac{dy}{dx} = \frac{Q(x,y)}{P(x,y)} \tag{3.9}
\]

There are three variables \(x, y\) and \(\frac{dy}{dx}\) in this equation. \(\frac{dy}{dx}\) is the slope of the curve \((x,y)\) in the \(x,y\) plane. Assume that the slope is constant and equal to \(N_1\), then

\[
\frac{Q(x,y)}{P(x,y)} = N_1 \tag{3.10}
\]

which is a function of \(x\) and \(y\) only. This curve is called an "isocline".

If a number of isoclines are drawn, the phase trajectory can be constructed from the given initial conditions by drawing straight line segments directed according to the slope of the isoclines.
The method of drawing isoclines is illustrated in figure 3.5.

The initial operating point is A. The phase trajectory should cross isocline \( N_1 \) at a slope \( N_1 \). Therefore one dotted line is drawn at a slope \( N_1 \). But the trajectory should cross isocline \( N_2 \) at a slope \( N_2 \). Accordingly another dotted line is drawn at a slope \( N_2 \). Point B is therefore chosen as the mid point of the segment intercepted by the dotted lines. Similar procedure is followed in locating points C and D.

It may be pointed out that this is an approximate procedure to draw the trajectory but the error is very small. A great advantage of this method is that a number of phase trajectories may be drawn with little more effort than is required for a single trajectory.

Consider eq. (3.8) again,

\[
\frac{d\omega}{d\delta} = \left[ A - B (C + k' \omega) \sin \delta + k' \sin 2\delta \right] / \omega
\]

Let

\[
\frac{d\omega}{d\delta} = N_1 = \text{a constant}
\]
Fig. 3.5 Construction of a Phase Trajectory using the method of Isoclines.
Phase Plane Analysis

\[ K_2 = 0.005 \times \frac{180}{\pi} \]

Isoclines

Phase Trajectory

Effect of \( K_2 \) on Isoclines

Fig. 3.6-4
Effect of $K_2$ on isolines

\[ K_2 = 0.01 \times \frac{180}{\pi} \]

fig. 5.6-\(x\).
Phase Plane Analysis of Auxiliary Signals applied to Excitation System.

\[ K_2 = 0.015 \times \frac{180}{\pi} \]

Isoclines

Phase Trajectory

Fig. 3 a-c
If $N_1$ is substituted for $\frac{d\omega}{d\delta}$ in eq. (3.8), then the following equation is obtained:

$$\omega N_1 = A - B(C + K_2 \omega) \sin \delta + k' \sin 2\delta$$

$$\omega(N_1 + BK_2 \sin \delta) = A - BC \sin \delta + k' \sin 2\delta$$

or

$$\omega = \frac{A - BC \sin \delta + k' \sin 2\delta}{N_1 + BK_2 \sin \delta}$$ (3.11)

Equation (3.11) is the equation of the isocline. It would be possible to draw a set of isoclines for different values of $N$ and to construct a phase trajectory from them. If $K_2$ is varied its effect can be conveniently analyzed from the shape of the phase trajectories.

Figure 3.6a through 3.6c show isoclines and trajectories drawn for the parameters under consideration. From these figures the best value of $K_2$ is found to be $0.01 \times \frac{180}{\pi}$.

**Exciter Input**

In the previous sections, expressions were derived for the voltage behind the transient reactance to produce maximum damping. In a salient pole machine this voltage is equal to $E'q$ which is proportional to field flux linkages. In a practical system, $E'q$ is controlled by means of signals applied to the regulator. It is, therefore, necessary to derive expressions for voltage input to the regulator for the conditions of maximum damping.
As developed in Chapter II, the relationships between \( E'_q \) and the voltage applied to the exciter is as follows:

**For Excitation Winding (Neglecting Saturation)**

\[
\frac{d \ E'_q}{dt} = \frac{1}{T'_d} \ (E'_{ex} - E'_q)
\]  

(3.12)

where \( E'_{ex} \) = open-circuit excitation voltage in p.u.

\( E'_q \) = voltage behind synchronous reactance in p.u.

\( T'_d \) = field time constant in seconds.

\( E'_q \) = voltage proportional to field flux linkages.

**For Exciter**

\[
\frac{d \ E_{ex}}{dt} = \frac{1}{T_e} \ (E'_x - E'_{ex})
\]  

(3.13)

where \( E'_x \) = exciter field voltage in per unit.

\( T_e \) = exciter field time constant in seconds.

(a) Assume that a signal proportional to rotor angle deviation is applied to the excitation system such that

\[
E'_q = E'_{q0} + K_1 \ (\delta - \delta_0)
\]  

(3.4)

From eq. (3.12) \( E'_{ex} = E'_q + T'_d \frac{d}{dt} E'_q \)

\[
E'_q = E'_q + \frac{1}{d} \ (x'_d - x'_d)
\]
Therefore

\[ E_{ex} = E'_{qo} + i_d (x_d - x'_d) + T'_d [K_1 \frac{d\delta}{dt}] \]  \hspace{1cm} (3.14)

Substituting eq. (3.4) in eq. (3.14), the result is

\[ E_{ex} = E'_{qo} + K_1 (\delta - \delta_0) + i_d (x_d - x'_d) + T'_d K_1 \frac{d\delta}{dt} \]

\[ + T_e \left\{ K_1 \frac{d\delta}{dt} + (x_d - x'_d) \frac{di_d}{dt} \right\} \]

\[ + T'_d K_1 \frac{d^2\delta}{dt^2} \]  \hspace{1cm} (3.15)

From eq. (3.13)

\[ E_x = \text{exciter field voltage} = E_{ex} + T_e \frac{d}{dt} E_{ex} \]

\[ = E'_{qo} + K_1 (\delta - \delta_0) + i_d (x_d - x'_d) + T'_d K_1 \frac{d\delta}{dt} \]

\[ + \frac{d^2\delta}{dt^2} (T_e T'_d K_1) + \frac{d\delta}{dt} (T_e K_1 + T'_d K_1) \]

\[ + E'_{qo} + (\delta - \delta_0) K_1 + (x_d - x'_d)i_d \]

\[ + (x_d - x'_d) T_e \frac{di_d}{dt} \]

\[ = A_1 \frac{d^2\delta}{dt^2} + A_2 \frac{d\delta}{dt} + A_3 \frac{di_d}{dt} + E_q \]  \hspace{1cm} (3.16)

Therefore the voltage applied to the exciter field should have signals proportional to the second derivative of the torque angle, the first derivative of the torque angle and direct axis component of current.
(b) Assume that a signal is applied to the excitation system proportional to \( \frac{d\delta}{dt} \) such that

\[
E'_q = E'_qo + K_2 \frac{d\delta}{dt} \quad (3.17)
\]

\[
E_q = E'_q + i_d (x_d - x'_d) = E'_qo + K_2 \frac{d\delta}{dt} + i_d (x_d - x'_d)
\]

\[
E_{ex} = E'_q + i_d (x_d - x'_d) + T'do \frac{d}{dt} E'_q \quad (3.14)
\]

Substituting (3.17) in (3.14)

\[
E_{ex} = E'_qo + K_2 \frac{d\delta}{dt} + (x_d - x'_d)i_d + T'do [K_2 \frac{d^2\delta}{dt^2}]
\]

\[
E_x = E_{ex} + T_e \frac{d}{dt} E_{ex}
\]

\[
= E'_qo + K_2 \frac{d\delta}{dt} + (x_d - x'_d)i_d + T'do [K_2 \frac{d^2\delta}{dt^2}]
\]

\[
+ T_e [K_2 \frac{d^3\delta}{dt^3} + (x_d - x'_d) \frac{di_d}{dt} + T'do K_2 \frac{d^3\delta}{dt^3}]
\]

\[
= \frac{d^3\delta}{dt^3} (T_e T'do K_2) + \frac{d^2\delta}{dt^2} (T_e K_2 + T'do K_2)
\]

\[
+ (x_d - x'_d) T_e \frac{di_d}{dt} + E_q
\]

\[
= A_4 \frac{d^3\delta}{dt^3} + A_5 \frac{d^2\delta}{dt^2} + A_6 \frac{di_d}{dt} + E_q \quad (3.18)
\]

where A's are constants.

**Regulator Input (with saturation)**

(a) Assume that a signal is applied to the excitation system such that

\[
E'_q = E'_qo + K_1(\delta - \delta_o)
\]

\[
E_{ex} = (1 + S_d) E_q + T'do \frac{d}{dt} E'_q
\]
\[ E_q = E'_q + i d \left( \frac{x_d - x'_d}{1 + S_d} \right) \]

\[ E_{ex} = \left\{ E'_q + i d \left( \frac{x_d - x'_d}{1 + S_d} \right) \right\} (1 + S_d) + T' do \frac{d}{dt} E'_q \]

\[ = E'_q (1 + S_d) + i d (x_d - x'_d) + T' do \frac{d}{dt} E'_q \]

\[ = [E'_{q0} + K_1 (\delta - \delta_o)] (1 + S_d) + i d [x_d - x'_d] + T' do [K_1 \frac{d\delta}{dt}] \]

\[ E_x = E_{ex} + T e \frac{d}{dt} E_{ex} \]

\[ = [E'_{q0} + K_1 (\delta - \delta_o)] (1 + S_d) + i d (x_d - x'_d) \]

\[ + T' do K_1 \frac{d\delta}{dt} + T e \left\{ K_1 (1 + S_d) \frac{d\delta}{dt} \right\} + T e \left\{ (x_d - x'_d) \right\} \frac{d}{dt} i d \]

\[ + T' do T e K_1 \frac{d^2\delta}{dt^2} \]

\[ = A_1 \frac{d^2\delta}{dt^2} + A_2 \frac{d\delta}{dt} + A_3 \frac{di d}{dt} + (1 + S_d) E_q \quad (3.19) \]

(b) Assume that a signal proportional to \( \frac{d\delta}{dt} \) is applied to the excitation system such that

\[ E'_q = E'_{q0} + K_2 \frac{d\delta}{dt} \]

\[ E_q = E'_q + i d \left( \frac{x_d - x'_d}{1 + S_d} \right) \]

\[ E_{ex} = (1 + S_d) E_q + T' do \frac{d}{dt} E'_q \]

\[ E_{ex} = \left\{ E'_q + i d \left( \frac{x_d - x'_d}{1 + S_d} \right) \right\} (1 + S_d) + T' do \frac{d}{dt} E'_q \]

\[ = E'_q (1 + S_d) + (x_d - x'_d) i d + T' do \frac{d}{dt} E'_q \]
\[
= (1 + S_d) \left\{ E'_q \sec + K_2 \frac{d\delta}{dt} \right\} + (x_d' - x_d') i_d + T'_do \ K_2 \frac{d^2\delta}{dt^2}
\]

\[
E_x = E_{ex} + T_e \frac{d}{dt} E_{ex}
\]

\[
= (1 + S_d)(E'_q \sec + K_2 \frac{d\delta}{dt}) + (x_d' - x_d') i_d + T'_do \ K_2 \frac{d^2\delta}{dt^2}
\]

\[
+ T_e \left\{ (1 + S_d)K_2 \frac{d^2\delta}{dt^2} + (x_d' - x_d') \frac{di_d}{dt} + T'_do \ K_2 \frac{d^3\delta}{dt^3} \right\}
\]

\[
= \frac{d^3\delta}{dt^3} \ T_e \ T'_do \ K_2 + \frac{d^2\delta}{dt^2} \left\{ T_e (1 + S_d) K_2 + T'_do \ K_2 \right\}
\]

\[
+ \frac{d\delta}{dt} \left\{ (1 + S_d)K_2 \right\} + T_e (x_d' - x_d') \frac{di_d}{dt}
\]

\[
+ (x_d' - x_d') i_d
\]

\[
+ (1 + S_d) E'_qo
\]

\[
= \frac{d^3\delta}{dt^3} (T_e T'_do K_2) + \frac{d^2\delta}{dt^2} \left\{ T_e (1 + S_d) K_2 + T'_do \ K_2 \right\}
\]

\[
+ T_e (x_d' - x_d') \frac{di_d}{dt} + (1 + S_d) E_q
\]

\[
= A_4 \frac{d^3\delta}{dt^3} + A_5 \frac{d^2\delta}{dt^2} + A_6 \frac{di_d}{dt} + (1 + S_d) E_q \tag{3.20}
\]

Amplification Factors (without saturation)

Two modes of controlling transients have been developed. In one mode, signals proportional to \( \frac{d\delta}{dt} \), \( \frac{d^2\delta}{dt^2} \) and \( \frac{di_d}{dt} \) are needed and amplification factors in various channels are

\[
\frac{d^2\delta}{dt^2} \quad \text{amplification factor } A_1 = T_e T'_do K_1
\]

\[
\frac{d\delta}{dt} \quad \text{amplification factor } A_2 = T_e K_1 + T'_do K_1
\]
\[
\frac{di_d}{dt} \quad \text{amplification factor } A_3 = (x_d - x'_d) T_e. 
\]

In the second mode, which results in a dead-beat transient control, for maximum transmitted power the amplification factors are

\[
\frac{d^3 \delta}{dt^3} \quad \text{amplification factor } A_4 = T_e T'_d K_2 
\]

\[
\frac{d^2 \delta}{dt^2} \quad \text{amplification factor } A_5 = T_e K_2 + T'_d K_2 
\]

\[
\frac{di_d}{dt} \quad \text{amplification factor } A_6 = (x_d - x'_d) T_e. 
\]

(with saturation)

**First Mode**

\[
\frac{d^2 \delta}{dt^2} \quad \text{amplification factor } A_1 = T_e T'_d K_1 
\]

\[
\frac{d\delta}{dt} \quad \text{amplification factor } A_2 = T'_d K_1 + T_e K_1(1 + S_d) 
\]

\[
\frac{di_d}{dt} \quad \text{amplification factor } A_3 = T_e (x_d - x'_d) 
\]

**Second Mode**

\[
\frac{d^3 \delta}{dt^3} \quad \text{amplification factor } A_4 = T_e T'_d K_2 
\]

\[
\frac{d^2 \delta}{dt^2} \quad \text{amplification factor } A_5 = T_e (1 + S_d) K_2 + T'_d K_2 
\]

\[
\frac{di_d}{dt} \quad \text{amplification factor } A_6 = (x_d - x'_d) T_e. 
\]
Conclusion

An analytical method has been developed to determine the correct magnitudes of feedback signals, in excitation systems, to produce the desired control of swings. This information will now be applied, in Chapters IV and V, to a typical synchronous machine system.
Chapter IV

EXCITATION SYSTEMS (DIGITAL COMPUTER SIMULATION)

Introduction

The theory developed in Chapter III is applied to the system of figure 2.1. The equations describing the system were developed in Chapter II. The excitation system utilized auxiliary derivative signals as developed in Chapter III. The system was simulated on a digital computer.

For the investigation of the transient stability of the generator, several tests may be applied but all are concerned with the behavior of the system subsequent to a sudden disturbance. The latter may be a fault on a transmission line, the dropping of a line section, a sudden application of load or a change of input power.

The change in input power is the method used in this study. The generator is adjusted initially to operate at rated power output and terminal voltage and is then subjected to a sudden increase in the power input. Transient stability is said to exist, if the machine regains a state of equilibrium after such a disturbance.

Excitation Control

In Chapter III two modes for input voltage to the exciter field were developed which are as follows:
Mode I

\[ E_x = A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d \delta}{dt} + A_3 \frac{d i_d}{dt} + E_q \]

Mode II

\[ E_x = A_4 \frac{d^3 \delta}{dt^3} + A_5 \frac{d^2 \delta}{dt^2} + A_6 \frac{d i_d}{dt} + E_q \]

There are some practical considerations in applying the above theory to actual cases, as can be observed from the following:

a) There are upper and lower limits to the voltage that can be applied to the exciter field.

b) In actual cases it will be necessary to provide a signal proportional to the difference between the magnitudes of the terminal voltage and the desired terminal voltage.

c) Circuits have been designed to measure quickly and with reasonable accuracy the phase angle \( \delta \). Researchers have experimented with signals based on speed, from which it appears that it is possible to obtain with reasonable accuracy \( \frac{d \delta}{dt} \) and \( \frac{d^2 \delta}{dt^2} \) signals. It is not, however, certain as yet whether a noise free \( \frac{d^3 \delta}{dt^3} \) signal can be obtained in actual practice.

At this stage it was therefore decided to restrict the computer investigation to application of \( \frac{d \delta}{dt} \) and \( \frac{d^2 \delta}{dt^2} \) signals only, or Mode I.
d) The influence of saturation depends largely on the shape of the saturation curve. In Chapter II and III, machine equations were modified to take saturation into account. So that the true effects of derivative signals are not masked, it was decided to neglect saturation in the computer study. Other investigators have followed the same procedure\textsuperscript{3-5}. It can, however, be taken into account if desired, by using the modified equations of Chapters II and III.

System Equations

\[
P_u = \frac{v_b E'}{x_d + x_e} \sin \delta + \frac{v_b^2 (x'_d - x'_q)}{2(x_q + x_e)(x'_d + x_e)} \sin 2\delta \tag{2.15}
\]

\[
E'_q = \frac{E'_q (x_d + x_e)}{(x'_d + x_e)} - v_b \cos \delta \frac{(x_q - x'_d)}{(x'_d + x_e)} \tag{2.22}
\]

\[
\frac{dE_{\text{ex}}}{dt} = \frac{1}{T_e} (E_x - E_{\text{ex}}) \tag{2.26}
\]

The exciter field voltage is \(E_x\) and by equation 3.16

\[
E_x = A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d\delta}{dt} + A_3 \frac{di_d}{dt} + E_q \tag{3.16}
\]

In the steady-state condition \(E_{x0} = E_{q0}\) where subscript 0 denotes the steady-state value in the pre-fault state. On account of automatic voltage regulator requirement, a signal proportional to \((v_r - v_m)\) will be inserted into the exciter.
Therefore

\[ E_x = E_q + A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d\delta}{dt} + A_3 \frac{di}{dt} + \mu_v (v_r - v_m) \]

where \( \mu_v \) is the gain in the voltage channel.

The above equation can be written as follows:

\[ E_x = E_{q0} + (E_q - E_{q0}) + A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d\delta}{dt} + A_3 \frac{di}{dt} + \mu_v (v_r - v_m) \]

In the above equation the terms \((E_q - E_{q0})\) and \(A_3 \frac{di}{dt}\) are small as compared to \(A_1 \frac{d^2 \delta}{dt^2}\) and \(A_2 \frac{d\delta}{dt}\). If these are neglected, for simplification, the equation reduces to

\[ E_x = E_{q0} + A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d\delta}{dt} + \mu_v (v_r - v_m) \]

The input signal to the exciter is, therefore,

\[ E_x = E_{xo} + A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d\delta}{dt} + \mu_v (v_r - v_m), \quad (4.01) \]

since \( E_{q0} = E_{xo} \) as shown in Chapter II.

In eq. 4.01 \( A_2 = K_1 (T'_d + T_e) \) and \( A_1 = T_e T'_d K_1 \) as derived in Chapter III.
Chapter V

EXCITATION SYSTEMS (DIGITAL COMPUTER RESULTS)

Introduction

It is desired to study the transient stability of a synchronous generator, connected to an infinite bus, by means of a double circuit transmission line, when the input power of the generator is suddenly increased.

The results given in this chapter were obtained by simulating the system on a digital computer. Input power was increased suddenly until the system became unstable. The system was simulated with a conventional Automatic Voltage Regulator (A. V. R.) to maintain the machine terminal voltage constant. Curves were drawn from the data obtained with and without the use of auxiliary signals to make a comparison between the two cases. In this particular study, transient instability occurred when the input power was suddenly increased by 40 percent.

System Equations and Constraints

The investigation is based on the equations developed in previous chapters and for ready reference they are reproduced hereunder:

\[
P_u = \frac{v_b E_q'}{x_d' + x_e} \sin \delta + \frac{v_b^2 (x_d' - x_q')}{2(x_q' + x_e')(x_d' + x_e')} \sin 2\delta \tag{2.15}
\]

\[
E_q = \frac{E_q'}{(x_d' + x_e')} - v_b \cos \delta \frac{(x_q' - x_d')}{(x_d' + x_e')} \tag{2.22}
\]
\[ \frac{dE'}{dt} = \frac{1}{T'} \frac{dE}{dt} \quad (E_{ex} - E_q) \quad (2.5) \]

\[ \frac{dE_{ex}}{dt} = \frac{1}{T_e} (E_x - E_{ex}) \quad (2.26) \]

\[ E_x = E_{x0} + \mu_v (v_r - v_m) + A_1 \frac{d^2 \delta}{dt^2} + A_2 \frac{d\delta}{dt} \quad (4.01) \]

\[ A_1 = T_e T'_d o K_1, \quad A_2 = (T'_d o + T_e) K_1 \quad (3.16) \]

A typical machine was selected for the investigation with the constants as given below. All values are in per unit system.

\[ x_d = 1.25 \quad H = 3.0 \quad x_q = .70 \quad T'_d o = 5.0 \]

\[ x'_d = .28 \quad T_e = .15 \quad x_e = .5 \]

Exciter Field Ceiling voltage \[ Max = 3.5 \quad Min = -1 \]

The following operating conditions are assumed.

\[ P_u = 1.0, \quad v_b = 1.0 \quad v_m = 1.18 \quad f = 60 \text{ c/s}. \quad \text{Power Factor}=1.0. \]

The value of \( K_1 \) used in eq. (3.16) was \( .01 \times \frac{180}{\pi} \). This value was selected by plotting eq. (3.7) for this system as shown in figure 3.3.

By calculations, the initial values of \( E'q', \delta, E_{qd'}, \text{ and } E_q \) were found as follows:

\[ E'_q = 1.245 \quad \delta = 50.2^\circ \quad E_{qd} = 1.562 \quad E_q = 1.986 \]

\[ A_2 = .01(5 + .15) = .0515 \times \frac{180}{\pi} \quad A_1 = .01(5 \times .15) = .0075 \times \frac{180}{\pi} \]

The computer program is enclosed with the appendix.

**Computer Results:**

Figure 5.1 shows the Swing Curve, when the input power is raised by 30 percent. It can be seen that the auxiliary signals have a significant
damping effect on the oscillations.

Figure 5.2 shows the variation of terminal voltage with time during a transient. Without the derivative signals, the terminal voltage is higher but it has large fluctuations. With the derivative signals, the fluctuations have almost disappeared although the magnitude of the voltage is slightly less. The terminal voltage begins to rise as the oscillations of the rotor disappear. It is, therefore, observed that the quick damping of rotor oscillations is achieved at the expense of slight drop in terminal voltage during the transient.

Figure 5.3 shows the variation of field voltage with time. When auxiliary signals are used, the machine field voltage undergoes rapid changes, even reversing at a particular instant. This shows that to attain quick damping of rotor oscillations, the field voltage should be increased quickly to the maximum positive value reversed and again increased in the positive direction. This is an application of the classic "Bang-Bang" principle.

Figure 5.4 shows two swing curves drawn when the input power was 1.4 p.u. Both curves are for a system which includes the normal Automatic Regulator Action. It is observed that the auxiliary signals have significant effect on the damping of the oscillations.
Figure 5.5 is a graph of exciter field voltage and time. If there were no ceiling limit on the exciter field voltage, then the magnitude of this voltage would rise to as high a value as 9.5 per unit. Due to the ceiling limit, it does not go above 3.5 per unit. The results of this limit is that the damping of the oscillations is not as rapid as it would be if there was no ceiling on the exciter field voltage.
Input Power 1.3

Normal A.V.R in operation

Fig. 5.1 Swing Curves
Fig. 5.2 Sending End Terminal Voltage
Input Power 1.3 p.u.
Normal A.V.R. Control

Without auxiliary signals
With auxiliary signals

Time-seconds

Fig. 5.3 Machine Field Voltage
Input Power 1.4 p.u.

With auxiliary signals only and normal A.V.R. Control

No auxiliary signals but normal A.V.R. Control

Fig. 5.4 Swing Curve for input 1.4
Input Power 1.3 p.u. Normal A.V.R. Control

Voltage limited due to ceiling

Fig. 5.5 Exciter Field Voltage
CHAPTER VI
NORMALIZED STABILITY MARGIN

Introduction

Since digital computers are becoming commonplace in utility system operation, there is a strong possibility that during the next decade they will be used to make decisions and initiate corrective action during transient periods. It will therefore be convenient to have a numerical technique by which relative stability of a system can be conveniently determined. Baba et al suggested the use of phase-plane technique to develop an expression for stability margin of a synchronous generator. In this chapter the familiar equal-area criterion will be used to define stability margin and an expression will be developed for what the writer will call "Normalized Stability Margin", wherein the positive limit of stability margin will be unity. The concept of normalized stability margin will be used in Chapters VII and VIII to evaluate the effect of Phase-Shifting transformers.

Stability Margin

Figure 6-1 shows pre-fault and post-fault power angle curves. The amplitude of the post-fault curve is less due to the fact that post-fault reactance is higher, due to switching out of a line section. For simplicity, fault clearing time, being very small, is neglected but it will be taken into account later on.
According to equal area criterion,

if area $A_2 > area A_1$, the system is stable

if area $A_2 < area A_1$, the system is unstable.

\[
\text{Area } A_1 = \int_{\delta_1}^{\delta_2} (P_1 - P_{m2} \sin \delta) d\delta = P_1(\delta_2 - \delta_1) + P_{m2}(\cos \delta_2 - \cos \delta_1)
\]

\[
\text{Area } A_2 = \int_{\delta_2}^{\pi - \delta_2} (P_{m2} \sin \delta - P_1) d\delta = P_{m2}(\cos \delta_2 + \cos \delta_1) - P_1(\pi - 2\delta_2)
\]

Therefore $A_2 - A_1 = P_{m2}(\cos \delta_2 + \cos \delta_1) + P_1(\delta_1 + \delta_2 - \pi) \quad (6.1)$

It will be convenient to normalize expression 6.1 so that maximum stability margin will be unity which will occur at no load.

The lesser the load, the larger will be the area $A_2$ and smaller the area $A_1$; hence, maximum value of $(A_2 - A_1)$ will occur when $P_1 = 0$.

It will be convenient to designate the maximum value as unity; therefore, if the expression 6.1 is divided by maximum value of $A_2$ (or total area under the post-fault curve), the expression is normalized. The area under post-fault curve is equal to $2P_{m2}$.

Therefore normalized stability margin

\[
v = \frac{1}{2} \left\{ (\cos \delta_2 + \cos \delta_1) + \frac{P_1}{P_{m2}} (\delta_1 + \delta_2 - \pi) \right\} \quad (6.2)
\]
Fig. 6.1 - Determination of Stability Margin
Derivation of Formula for Stability Margin when Fault Clearing Time is Taken into Account

Let $P_{m1}$, $P_{m2}$, and $P_{m3}$ be the maximum power transferrable under pre-fault, during-fault and post-fault conditions. Refer to figure 6.2. Let $P$ be the input power in this case.

\[
\text{Area } A_1 = \int_{\delta_1}^{\delta_2} (P - P_{m2} \sin \delta) d\delta + \int_{\delta_2}^{\delta_3} (P - P_{m3} \sin \delta) d\delta
\]

\[
= P(\delta_2 - \delta_1) + P_{m2}(\cos \delta_2 - \cos \delta_1) + P(\delta_3 - \delta_2) + P_{m3}(\cos \delta_3 - \cos \delta_2)
\]

\[
\text{Area } A_2 = \int_{\delta_3}^{\pi} [P_{m3} \sin \delta - P] d\delta
\]

\[
= -P_{m3}[\cos(\pi - \delta_3) - \cos \delta_3] - P[\pi - \delta_3 - \delta_3]
\]

\[
= 2P_{m3} \cos \delta_3 - P[\pi - 2\delta_3]
\]

Therefore $A_2 - A_1 = P[\delta_3 + \delta_1 - \pi] + P_{m2}[\cos \delta_1 - \cos \delta_2] + P_{m3}[\cos \delta_3 + \cos \delta_2]$

Dividing it by area under post-fault curve equal to $2P_{m3}$

\[
\text{Stability Margin} = \frac{1}{2}[\cos \delta_2 + \cos \delta_3] + \frac{P_{m2}}{2P_{m3}}[\cos \delta_1 - \cos \delta_2]
\]

\[
+ \frac{P}{2P_{m3}}[\delta_1 + \delta_3 - \pi]
\]

(6.3)
Fig. 6.2 Pre-fault, During-fault, and Post-fault Power Angle Curves
Example 6.1

The following example illustrates the application of the concept of normalized stability margin.

The equations describing the power flow from a generator are

Before the fault $P_o = 1.735 \sin \delta$

During the fault $P_o = .42 \sin \delta$

After the fault $P_o = 1.25 \sin \delta$

where $P_o$ is the power output from the generator.

Assume fault removal when $\delta = 45^\circ$. Calculate the stability margin, if generator is delivering 1.0 per unit power.

\[
\begin{align*}
\varphi_{m1} &= 1.735, \delta_1 = \sin^{-1} \frac{1.0}{1.735} = 35.2^\circ = .615 \text{ radian} \\
\varphi_{m2} &= .42, \delta_2 = 45^\circ = .785 \text{ radian} \\
\varphi_{m3} &= 1.25, \delta_3 = \sin^{-1} \frac{1.0}{1.25} = 53.1^\circ = .928 \text{ radian}
\end{align*}
\]

Stability Margin = \[
\frac{1}{2} \left\{ .60 + .707 \right\} + \frac{.42}{2.50} \left\{ .818 - .70 \right\} \\
+ \frac{1.0}{2.50} \left\{ .61 + .928 - 3.141 \right\} \\
= .20 \text{ unit.}
\]

If the fault is removed at an angle $\delta = 51.6^\circ$, then stability margin

\[
= \frac{1}{2} \left\{ .60 + .622 \right\} + \frac{.42}{2.50} \left\{ .818 - .622 \right\} + \frac{1.0}{2.50} \left\{ .615 + .928 - \right. \\
- 3.141 \right\} = 0.0 \text{ per unit}
\]

Hence this corresponds to critical switching angle.
Chapter VII

PHASE SHIFTING TRANSFORMERS (BASIC ANALYSIS)

Introduction

Another method of improving transient stability in a power system is investigated. This method involves using phase-shifting transformers to shift the power angle curve by a certain angle to produce the desired control of swings. The phase-shift windings of the transformers can be short time rated which will reduce their cost.

Principle of the Method

Figure 7.1 shows a generator G supply power to an infinite bus through a double circuit transmission line. On the sending end there is a transformer T which has a phase shifting arrangement as shown in the figure. A is the main breaker, while B is an additional breaker. The two breakers have an interlocking device so that at any time when one breaker is closed, the other is open and vice versa. The generator is supplying power \( P_1 \). The voltage behind transient reactance of the generator is \( E'_q \) and that of the infinite bus is \( v_b \). \( X \) is the sum of the line reactance, the transformer reactance and the transient reactance. Therefore

\[
P_1 = \frac{E'_q v_b}{X} \sin \delta
\]

where \( \delta \) is phase angle between the two voltages.
Fig. 7.1 Application of P.S. Transformers in a Typical System
Fig. 7.2 Effect of Phase Shift on Power Angle Curve
The usual power angle curve is shown in figure 7.2. A fault occurs on one of the lines which is switched out due to the action of the fast-acting breakers. The loss of the line increases the reactance between the two voltages and operation moves from pre-fault to post-fault power angle curve. Since the duration of the fault is very short, no "during-fault" curve is shown for the sake of simplicity. As expected, the generator rotor begins to accelerate. In the case illustrated in the figure, the system is unstable since $P_1$ is greater than transient stability limit of the system.

If it is desired to control the swings, breaker A opens and B closes at the same time that the line is switched out. This is not difficult, since the relays which switch out the line can open A and close B also. Due to the phase-shifting property of the windings, the post-fault power angle is shifted towards the left by an angle $\theta$ as shown in figure 7.2. The result of this shift is the accelerating area is reduced while the decelerating area is increased, thereby increasing the stability margin. The magnitude of the swing is reduced too. If the angle $\theta$ is of correct magnitude, it is possible to bring the rotor to point $o$ such that $\frac{d\delta}{dt} = 0$ at $o$. Under that condition, if the phase-shift is removed when the operating point reaches $o$, there will be no back swing. Thus the use of phase shifting transformers has resulted in bringing the operating point to $o$ in a dead-beat non-oscillatory manner.
Phase-Plane Analysis of the Effect of Phase Shifting

To illustrate the basic effects of phase-shifting transformers, the phase-plane technique is used to analyze the swing equation. The following assumptions are made.

1. Damping is ignored.
2. The synchronous machine is represented by a fixed voltage behind the transient reactance.
3. Saturation is neglected.
4. The prime mover input remains constant during a swing.
5. The machine has a smooth rotor.

Therefore

\[ M \frac{d^2 \delta}{dt^2} = P_1 - P_{m1} \sin \delta \] \hspace{1cm} (7.1)

where \[ P_{m1} = \frac{E'_q v_b}{X} \]

The steady-state condition is given by

\[ P_1 = \frac{E'_q v_b}{X} \sin \delta_0 \] \hspace{1cm} (7.2)

The representative point is shown in phase-plane figure 7.3, by A. A 3-phase fault occurs on the system which reduces the output to zero, hence

\[ M \frac{d^2 \delta}{dt^2} = P_1 \text{ during the fault} \]

\[ M \frac{d\delta}{dt} = P_1 t + C \] \hspace{1cm} (7.3)
\[ t = \text{time after occurrence of the fault} \]

or \( M \omega = P_1 t + C \), where \( \omega = \frac{d\delta}{dt} \), \( C \) is a constant. When \( t = 0 \), \( \omega = 0 \), therefore \( C = 0 \). Therefore the phase-trajectory equation during the fault is

\[ M \omega = P_1 t \]

or \( \omega = \frac{P_1 t}{M} = \frac{d\delta}{dt} \) \hspace{1cm} (7.4)

Integrating again the above equation

\[ \delta = \frac{P_1}{2M} t^2 + C_1 \]. \( C_1 \) is a constant. When \( t = 0 \), \( \delta = \delta_0 \), therefore \( C_1 = \delta_0 \). Hence

\[ \frac{P_1}{2M} t^2 = \delta - \delta_0 \] \hspace{1cm} (7.5)

Eliminating the variable \( t \) in equations (7.4) and (7.5), the result is

\[ \delta - \delta_0 = \frac{P_1}{2M} \frac{\omega^2}{P_1^2} M^2 = \frac{M}{2P_1} \omega^2 \]

or \( \omega^2 = \frac{2P_1}{M} (\delta - \delta_0) \) \hspace{1cm} (7.6)

In the phase-plane, equation (7.6) is represented by a trajectory AB, which represents the "during fault" portion of the trajectory.

After a time \( t_1 \), the fault is removed and the swing equation becomes
\[ M \frac{d^2 \delta}{dt^2} = P_1 - P_{m2} \sin \delta \quad (7.7) \]

where \[ P_{m2} = \frac{E' q b}{X'} \] and \( X' \) is the post-fault reactance.

The initial conditions for equation 7.7 are: when \( t = 0, \delta = \delta_1 \).
\[ \omega = \omega_1, \quad \omega_1 = \frac{P_1}{M t_1}, \delta_1 = \frac{M \omega_1^2}{2 P_1} + \delta_0 \]
from equations 7.4 and 7.6.

Assume the new steady-state point in the phase plane to be \((\delta_2, 0)\).

Then
\[ P_{m2} \sin \delta_2 = P_1. \]

Also,
\[ \omega = \frac{d\delta}{dt} \]

\[ \frac{d^2 \delta}{dt^2} = \frac{d\omega}{dt} = \frac{1}{M} \left[ P_1 - P_{m2} \sin \delta \right] \]

or
\[ \frac{d\omega}{d\delta} = \left[ P_1 - P_{m2} \sin \delta \right] \frac{1}{\omega M} \]

or
\[ \omega d\omega = \left[ \frac{P_1}{M} - \frac{P_{m2}}{M} \sin \delta \right] d\delta \]

Integrating,
\[ \frac{\omega^2}{2} = \frac{P_1}{M} \delta + \frac{P_{m2}}{M} \cos \delta + C_2 \quad (7.8) \]

\( C_2 \) is a constant. For the transient stability limit \( \omega = 0 \) when \( \delta = \pi - \delta_2 \),
which follows from the application of equal area criterion to figure 7.2.
\[ C_2 = -\frac{P_1}{M}(\delta - \delta_2) - \frac{P_m2}{M} \left[ \cos (\pi - \delta_2) \right] \]

\[ = -\frac{P_1}{M}(\pi - \delta_2) + \frac{P_m2}{M} \cos \delta_2 \]

or \[ \frac{\omega^2}{2} = \frac{P_1}{M} \delta + \frac{P_m2}{M} \cos \delta + \frac{P_m2}{M} \cos \delta_2 - \frac{P_1}{M}(\pi - \delta_2) \]

\[ = \frac{P_1}{M}(\delta - \pi + \delta_2) + \frac{P_m2}{M} (\cos \delta + \cos \delta_2) \] (7.9)

This is the equation of the Seperatrix. If the fault is removed while the representative point is within the area enclosed by the seperatrix, the system is stable, otherwise it is unstable.

Suppose that the fault is removed when the representative point is at B, which has coordinates \( \omega_1, \delta_1 \). Substituting this condition in equation 7.8.

\[ \frac{\omega^2_1}{2} = \frac{P_1}{M} \delta_1 + \frac{P_m2}{M} \cos \delta_1 + C_2 \]

\[ C_2 = \frac{\omega^2_1}{2} - \frac{P_1}{M} \delta_1 - \frac{P_m2}{M} \cos \delta_1 \]

Therefore \[ \frac{\omega^2}{2} = \frac{P_1}{M}(\delta - \delta_1) + \frac{P_m2}{M} (\cos \delta - \cos \delta_1) \]

\[ + \frac{\omega^2_1}{2} \] (7.10)
Fig. 7.4 Determination of the Effect of Phase-Shift on Stability Margin
This is shown in figure 7.3 by post-fault trajectory. It will be seen that the representative point oscillates continuously around the new steady state point $\delta_2$.

Suppose that at the time of switching out of the line, the power angle is shifted towards the left by an angle $\theta$. Then the swing equation becomes

$$M \frac{d^2 \delta}{dt^2} = P_1 - P_m^2 \sin(\delta + \theta)$$

The apparent singular point is $\delta_3 = \delta_2 - \theta$. The trajectory equation (7.10) becomes

$$\frac{\omega^2}{2} = \frac{P_1}{M} (\delta - \delta_1) + \frac{P_m^2}{M} (\cos(\delta + \theta) - \cos(\delta_1 + \theta)) + \frac{\omega_1^2}{2} \quad (7.11)$$

The new trajectory is again oscillatory and is shown in the dotted lines. Figure 7.3 shows qualitatively how the trajectories are shifted towards the left by the action of the phase shifters.

If the phase shift is of such a magnitude that the point C passes through $\delta_2$, then it would be possible to remove the phase shift at that instant and the representative point will stay there without further movement, so that the transient is controlled in a dead-beat manner.

**Effect of Phase Shift on Stability Margin**

In chapter VI, the concept of stability margin was introduced as it gives a quantitative idea of stability. Referring to figure 7.4

$$\text{Stability Margin with phase shifting} = \frac{\text{area b'd'c'} - \text{area o ab'}}{\text{Total area under post-fault curve}}$$

$$\quad (7.12)$$
Stability margin without phase shifting

\[
\begin{align*}
\text{area bdc - area oab} &= \frac{\text{Total area under post-fault curve}}{\text{(7.13)}} \\
\text{Therefore increase in stability margin} &= (7.12) - (7.13) \\
&= \frac{\text{area oab - area oa'b'}}{\text{Total area under post-fault curve}}
\end{align*}
\]

(since area b'd'c' = area bdc)

\[
\begin{align*}
\text{area oab} &= \int_{\delta_1}^{\delta_2} (P_1 - P_{m2} \sin \delta) \, d\delta \\
&= P_1(\delta_2 - \delta_1) + P_{m2}(\cos \delta_2 - \cos \delta_1)
\end{align*}
\]

Similarly

\[
\begin{align*}
\text{area oa'b'} &= P_1(\delta_3 - \delta_1) + P_{m2}(\cos \delta_3 - \cos \delta_1)
\end{align*}
\]

Total area under post-fault curve = \(2 P_{m2}\) approximately with and without phase shift.

Therefore increase in stability margin

\[
\begin{align*}
\text{Therefore increase in stability margin} = \frac{P_1(\delta_2 - \delta_3) + P_{m2}(\cos \delta_2 - \cos \delta_3)}{2 P_{m2}}
\end{align*}
\]

Since \(\delta_3 = \delta_2 - \theta\),

\[
\begin{align*}
\text{Increase in stability margin} = \frac{P_1 \theta + P_{m2}(\cos \delta_2 - \cos(\delta_2 - \theta))}{2 P_{m2}}
\end{align*}
\]
Fig. 7.5 θ for Dead-Beat Control
Determination of $\theta$ for Dead-beat Transient Control

Figure 7.5 shows the usual pre-fault and post-fault power angle curves. The apparent power angle curve, while the phase shifters are energized, is shown in the dotted curve. It is desired to find the value of $\theta$ which moves the representative point from $a$ to $e$ in a dead-beat non-oscillatory fashion, if the phase-shift is removed when it reaches $d$.

The criterion for the above is that the area $abc = \text{the area cde}$ such that when the representative point reaches $d$, its velocity is zero. If the phase-shift is removed at $d$, it will occupy point $e$ and stay there permanently, since power input equals output.

Therefore,

$$\int_{\delta_1}^{\delta_3} [P_1 - P_m \sin(\delta + \theta)] d\delta = \int_{\delta_3}^{\delta_2} [P_m \sin(\delta + \theta) - P_1] d\delta$$

or

$$P_1(\delta_3 - \delta_1) + P_m \left[ \cos(\delta_3 + \theta) - \cos(\delta_1 + \theta) \right]$$

$$= - P_m \left[ \cos(\delta_2 + \theta) - \cos(\delta_3 + \theta) \right] - P_1[\delta_2 - \delta_3]$$

or

$$P_1[\delta_2 - \delta_1] = P_m \left[ \cos(\delta_1 + \theta) - \cos(\delta_2 + \theta) \right]$$

or

$$\frac{P_1}{P_m} \left[ \delta_2 - \delta_1 \right] = \cos \delta_1 \cos \theta - \sin \delta_1 \sin \theta - \cos \delta_2 \cos \theta + \sin \delta_2 \sin \theta$$

$$= \cos \theta \left[ \cos \delta_1 - \cos \delta_2 \right] - \sin \theta \left[ \sin \delta_1 - \sin \delta_2 \right]$$
\[
\cos \delta_1 - \cos \delta_2 = A \sin \alpha \\
\sin \delta_1 - \sin \delta_2 = A \cos \alpha
\]

Let
\[
\begin{align*}
\text{or } \tan \alpha &= \frac{\cos \delta_1 - \cos \delta_2}{\sin \delta_1 - \sin \delta_2} \\
A^2 &= (\cos \delta_1 - \cos \delta_2)^2 + (\sin \delta_1 - \sin \delta_2)^2 \\
&= 4 \sin^2 \left( \frac{\delta_1 - \delta_2}{2} \right) \\
&= 4 \sin^2 \left( \frac{\delta_1 - \delta_2}{2} \right)
\end{align*}
\]

or
\[
A = 2 \sin \left( \frac{\delta_1 - \delta_2}{2} \right)
\]

Then
\[
\frac{P_1}{P_{m2}} [\delta_2 - \delta_1] = [\cos \theta \sin \alpha - \sin \theta \cos \alpha]A
\]

\[
= A \sin (\alpha - \theta)
\]

\[
\alpha - \theta = \sin^{-1} A \frac{P_1}{P_{m2}} [\delta_2 - \delta_1]
\]

or
\[
\theta = \alpha - \sin^{-1} A \frac{P_1}{P_{m2}} [\delta_2 - \delta_1]
\]

or
\[
\theta = \tan^{-1} \frac{\cos \delta_1 - \cos \delta_2}{\sin \delta_1 - \sin \delta_2} - \sin^{-1} A \frac{P_1}{P_{m2}} [\delta_2 - \delta_1] \quad (7.16)
\]

Increase in Power Transmitted

Figure 7.6 represents a pre-fault, a post-fault without phase shift and a post-fault with phase shift power angle curves. Without phase shift control, the maximum power transmitted will be \( P_1 \).
Figure 7.6 Determination of Minimum $\theta$
which is determined graphically, satisfying the equal area criterion, such that area $oab = area bod$.

It is desired to transmit power greater than $P_1$ over the system by employing phase shifting technique. Suppose it is desired to increase $P_1$ to $P'_1$, then the power angle curve needs to be shifted by an amount $\theta$ such that area $o'a''b' = area b'c'd'$ which is possible for some value of $\theta$, as can be observed from the figure. Thus the phase shifting transformers have made it possible to raise the power transmitted to a level more than the transient stability limit $P_1$, in the case under consideration.

The maximum value of the power transmitted is $P_{m2'}$, the steady-state limit in the post-fault condition. It will not be practicable to raise the power to this value. The maximum value of the transmitted power will depend upon the maximum angle $\delta_2$ desired in the post-fault state, which should be less than $90^\circ$. Hence if $\delta_2$ be the maximum angle desired, then $P'_1 = P_{m2} \sin \delta_2$. The phase-shifting transformer windings will be designed in such a way that power $P'_1$ can be transmitted.

The minimum angle of phase shift will be necessary if area $o'a''b' = area b'c'd'$ or that full area above the power line is used in decelerating the rotor.

Referring again to figure 7.6, area $o'a''b' = area b'c'd'$ or

$$\int_{\delta_1}^{\delta_3} [P'_1 - P_{m2} \sin(\delta + \theta)]d\delta = \int_{\delta_3}^{\pi - \delta_3 - \theta} [P_{m2} \sin(\delta + \theta) - P'_1]d\delta$$

This gives
\[
\frac{p_1'}{p_{m2}} \left[ \frac{\pi - \delta_1 - \delta_3}{-\theta} \right] = \left[ \cos(\delta_1 + \theta) + \cos \delta_3 \right]
\]

But

\[
p_1' = p_{m2} \sin \delta_2 = p_{m1} \sin \delta_1
\]

and \(\delta_2 = \delta_3 + \theta\)

Therefore

\[
\frac{p_1'}{p_{m2}} \left[ \frac{\pi - \delta_1 - \delta_2}{-\theta} \right] = \left[ \cos(\delta_1 + \theta) + \cos(\delta_2 - \theta) \right] \quad (7.17)
\]

\[
\frac{p_1'}{p_{m2}} \left[ \frac{\pi - \delta_1 - \delta_2}{-\theta} \right] = \cos \delta_1 \cos \theta - \sin \delta_1 \sin \theta + \cos \delta_2 \cos \theta
\]

\[
+ \sin \delta_2 \sin \theta
\]

\[
= \left[ \cos \delta_1 + \cos \delta_2 \right] \cos \theta + \left[ \sin \delta_2 - \sin \delta_1 \right] \sin \theta.
\]

Let

\[
\cos \delta_1 + \cos \delta_2 = A \sin \alpha
\]

or \(\tan \alpha = \frac{\cos \delta_1 + \cos \delta_2}{\sin \delta_2 - \sin \delta_1}\)

or \(\alpha = \tan^{-1} \frac{\cos \delta_1 + \cos \delta_2}{\sin \delta_2 - \sin \delta_1}\)

and

\[
A = 2 \cos \frac{(\delta_1 + \delta_2)}{2}
\]

Therefore

\[
\frac{p_1'}{p_{m2}} \left[ \frac{\pi - \delta_1 - \delta_2}{-\theta} \right] = A \left[ \sin(\alpha + \theta) \right]
\]
or

\[ \theta = \sin^{-1} \frac{P_1' [\pi - \delta_1 - \delta_2]}{(\delta_1 + \delta_2)} - \tan^{-1} \frac{\cos \delta_1 + \cos \delta_2}{\sin \delta_2 - \sin \delta_1} \]

**Application to a Typical System**

Chapter VIII gives details of an actual system and computer
details when the technique of phase shift control was applied to a
typical system.
Chapter VIII

APPLICATION OF PHASE SHIFTING TRANSFORMERS

Introduction

This chapter gives details of an investigation of a typical system using the method of phase-shifting transformers as developed in the preceding chapter. The study which was done by digital simulation indicates that there is considerable potential in this technique and it is hoped that the electric companies will be encouraged to explore its application to their systems.

System Studied

The system consists of a synchronous generator supplying power to an infinite bus by means of a double circuit transmission line, as shown in Figure 8.1. A fault occurs at point P which results in opening of the circuit breakers adjacent to P thereby raising transfer reactance between the generator and the infinite bus. The equations for the power output of the generator are:

Before the fault \[ P_{m1} \sin \delta = 1.735 \sin \delta \]

During the fault \[ P_{m2} \sin \delta = 0.42 \sin \delta \]

After the fault \[ P_{m3} \sin \delta = 1.25 \sin \delta \]

Input Power = 1.15, hence \[ \delta_1 = \sin^{-1} \frac{1.15}{1.735} = 41.4^\circ, \]

\[ \pi - \delta_2 = \sin^{-1} \frac{1.15}{1.25} = 113^\circ. \]

The power angle curves are shown in figure 6.2.
Fig. 8.1 A Typical System Using a P.S. Transformer
Modern beakers are capable of operating in 3 cycles (clearing time .05 sec.). From standard curves, the clearing angle for this case is $45.3^\circ$.

A computer program was written to calculate the data for the swing curve. The constants of the generator were as follows:

$$H = 3.0 \quad G = 1$$

Therefore

$$M = \frac{GH}{180 \times f} = 2.78 \times 10^{-4} \text{ per unit}$$

A time interval of 0.05 sec. was selected for the calculations.

Figure 8.2 shows the results of different values of phase shift on stability of the generator. The figure shows three curves obtained for phase shifts of $0^\circ$, $5^\circ$, and $10^\circ$. The effect of $10^\circ$ phase shift is to decrease the maximum torque angle from $85^\circ$ to $68^\circ$, a decrease of $17^\circ$. Similarly, the phase shift of $5^\circ$ decreases the maximum torque angle by $10^\circ$. Therefore the amount of phase shift has considerable effect on the shape of the swing curve.

The computer program and the flow chart are enclosed in the appendix.

It may be mentioned here that there are many application problems to be analyzed before the techniques of phase shifting transformers can be used in actual systems.
Fig. 8.2 Effect of Phase-shift Transformers on Swing Curve
Chapter IX

SWITCHED SERIES CAPACITORS

(APPROXIMATE ANALYSIS)

Introduction

A very effective method of raising stability limits of long transmission lines is by means of fixed series capacitors. There are some excellent papers on this subject in the literature.

In 1965 Kimbark\(^6\) suggested the use of switched series capacitors which involves switching in a capacitor in series with the line as soon as a line section is switched out. The idea behind this scheme is that by switching in a capacitor, the increase in line reactance, caused by switching out of a section, is minimized. Kimbark's paper was followed by that of Smith\(^12\). He suggested that capacitors be switched in and out at appropriate instants to control the transients, the exact instants would be determined by a centralized control computer.

Scope of Investigation

Chapters IX to XI of this report relate to switched series capacitors as shown below:

Chapter IX. This chapter is concerned with an approximate analysis of the fundamental effects of switched series capacitors. As in Kimbark\(^6\) and Smith's\(^12\) papers, equal area criterion is the method used in the investigation. It is shown that the capacitive reactance needed to maintain stability can be minimized if dead-beat operation is
not desired and formulas are derived for capacitor sizes under different modes of operation. A simple system is simulated on a digital computer to find the effect of switching in different sizes of capacitors.

Chapter X. This chapter is a detailed report on some experimental work performed in the Power Systems Laboratory of the University of Michigan. The experimental results are compared with the results obtained by digital simulation of the system.

Chapter XI. This chapter is concerned with the study of the following problems arising in application of switched series capacitors.

i. Effect, on system stability, of a second fault which may occur while the capacitor is in circuit and the machines are swinging.

ii. Effect of Switching in a series capacitor on a circuit with distance relaying

Approximate Analysis

The aim of the approximate analysis is to investigate the fundamental effects of switched series capacitors. The investigation is restricted to a system which has two similar parallel lines such that the reactances of the two lines are equal.

Figure 9.1 represents a double circuit transmission line connecting a generator to an infinite bus. \( X_1 \) is the reactance of each circuit. \( X_c \) is the reactance of the switched series capacitor. Circuit
Fig. 9.1 A Typical System Using Switched Series Capacitors

Fig. 9.2 Power Angle Curves for the Above System
breaker S is normally closed and it opens when any of the circuit
breakers CB-1 through CB-4 opens. It is assumed that the same relay
initiates the tripping of the breakers S and any of the breakers CB-1
through CB-4.

When a fault occurs at point F, circuit breakers CB-1 and CB-2
trip, raising the line impedance from \( \frac{x_1}{2} \) to \( x_1 \). Since circuit breaker
S opens, the net reactance between \( A \) and \( B \) is raised from \( \frac{x_1}{2} \) to
\( x_1 - x_c \). It is therefore clear that if the value of \( x_c \) is chosen to be \( \frac{x_1}{2} \),
the line reactance will remain unchanged although one circuit has been
relayed out. It will however be shown that it is not necessary to have the
capacitive reactance \( x_c \) equal to \( \frac{x_1}{2} \), since a significant improvement in
the transient stability limit will be achieved even when \( x_c \) is much lower
than \( \frac{x_1}{2} \).

**Without Switched Capacitor**

Figure 9.2 is the power angle diagram for the above system.

Sine-curve no. 1 represents the prefault curve and no. 2 the post fault
curve. The during-fault curve is neglected since fault duration is very
small. \( P_{m1} \) and \( P_{m2} \) are the amplitudes of the power angle curves given by

\[
P_{m1} = \frac{E'_q E_2}{x'_d + 0.5 x_1}, \quad P_{m2} = \frac{E'_q E_2}{x'_d + x_1}
\]

where

- \( E'_q \) is the voltage behind the transient reactance of the generator
- \( E_2 \) is the voltage of the infinite bus
- \( x'_d \) is the transient reactance of the generator
- \( x_1 \) is the reactance of each line.
Since it is an approximate method, the assumption will be made that \( x'_d \) is small in comparison with \( x_1 \) and hence may be neglected in the derivation to follow.

Let \( P_m \) be the transient stability limit derived on the basis of the equal area criterion such that area \( abc = area \ cod \). \( P_m \) can then be determined from the following equation

\[
P_m[\pi - \delta_2 - \delta_1] = P_{m2} [\cos \delta_2 + \cos \delta_1]
\]  

(3.1)

where \( \delta_1 \) and \( \delta_2 \) are given by the following equations

\[
\delta_1 = \sin^{-1} \frac{P_m}{P_{m1}} \quad \delta_2 = \sin^{-1} \frac{P_m}{P_{m2}}
\]

Switched-Series Compensation

The effect of switched series compensation can be seen from figure 9.3.

It is assumed that the circuit breaker \( S \) is operated by the same relay which operates either circuit breaker CB1 or CB2, thus the capacitor \( x_c \) is inserted as soon as the faulty circuit is taken out and reactance will increase from \( \frac{x_1}{2} \) to \( (x_1 - x_c) \). Consequently, the operation will be transferred from curve 2 to 2'.

The operating point moves from 1 to 2 and 2 to 3. The rotor starts to advance along the dashed curve from 3 to 4 at which point
is maximum. From 4, the operating point moves along the dashed curve to 5 at which point \( \frac{d\delta}{dt} = 0 \).

If the value of \( x_c \) is chosen such that 5 and 6 lie on the same vertical line, it is sufficient to remove the capacitor at 5 and the operating point will move from 5 to 6 and stay there, since 6 is the steady-state point in the post-fault condition. This method was originally suggested by Smith\(^{12} \).

The operation has therefore moved from 1 to 6 in a dead-beat, non-oscillatory fashion with the aid of a switched-series capacitor. The value of the capacitor which will be required for the process will be unique for the loading \( P_m \).

It can be argued from the above that, by means of a switched series capacitor, it will be possible to increase the transient stability limit up to the value of \( P_{m2} \), with a maximum value of \( \delta = 90^\circ \). This is indicated in figure 9.4.

However, it would not be practicable to operate at \( \delta = 90^\circ \) and, therefore, the transient stability limit will be less than \( P_{m2} \).

Selection of Capacitor Size

There are two criteria for selecting the capacitor size as follows:

(1) To produce dead-beat non-oscillatory transfer from pre-fault to post-fault state.

(2) To maintain stability but with operation which is not dead-beat. (This will result in a smaller capacitor size.)
Fig. 9.3  Effects of Switched Series on Power Angle Curves

Fig. 9.4  Series Compensation for Maximum Transient Stability Limit
1. Dead-beat operation. Refer to figure 9.3. Select $\delta_2$ to be the maximum angle at which operation in the post-fault state may take place.

Then $\sin \delta_2 = \frac{P_m}{P_{m2}}$ or $P_m = P_{m2} \sin \delta$. Thus $P_m$ is the maximum amount of power which is required to be transmitted along the lines.

$$\delta_2 = \sin^{-1} \frac{P_m}{P_{m1}}$$

$$P_m[\delta_2 - \delta_1] = P_{m2} \int_{\delta_1}^{\delta_2} \sin \delta \: d\delta = -P_{m2}[\cos \delta_1 \cos \delta_2 - \cos \delta_1 \cos \delta_2]$$

$$P_m[\delta_2 - \delta_1] = [\cos \delta_1 - \cos \delta_2] \cdot P_{m2}$$

$$P_{m2}' = \frac{P_m[\delta_2 - \delta_1]}{[\cos \delta_1 - \cos \delta_2]}$$

$$\frac{E_2'}{x_2'} = \frac{P_m[\delta_2 - \delta_1]}{[\cos \delta_1 - \cos \delta_2]}$$

where $x_2'$ is the total reactance when the capacitor is in the circuit.
\[
\frac{E'_2 E_2}{x_1} \frac{x_1}{x_2} = \frac{P_m [\delta_2 - \delta_1]}{\cos \delta_1 - \cos \delta_2}
\]

\[
\frac{P_{m1}}{2} \frac{x_1}{x_2} = \frac{P_m [\delta_2 - \delta_1]}{\cos \delta_1 - \cos \delta_2}
\]

\[
x_2' = \frac{P_{m1}}{2P_m} x_1 \frac{\cos \delta_1 - \cos \delta_2}{\delta_2 - \delta_1}
\]

But \(x_2' = x_1 - x_c\)

or \(x_1 - x_c = \frac{P_{m1}}{2P_m} x_1 \frac{\cos \delta_1 - \cos \delta_2}{(\delta_2 - \delta_1)}\)

\[
x_c = x_1 \left[ 1 - \frac{P_{m1}}{2P_m} \frac{\cos \delta_1 - \cos \delta_2}{\delta_2 - \delta_1} \right]
\]

\[
= x_1 \left[ 1 - \frac{P_{m1}}{2P_m} \frac{\sqrt{\frac{P_m^2}{P_{m1}^2} - 1}}{\sqrt{\frac{P_m^2}{P_{m1}^2} - 1}} \right]
\]

\[
= x_1 \left[ 1 - \frac{P_{m1}}{2P_m} \frac{\sqrt{\frac{P_m^2}{P_{m1}^2} - 1}}{\sqrt{\frac{P_m^2}{P_{m1}^2} - 1}} \right]
\]
\[ x_1 \left( 1 - \frac{E_q' E_2}{x_1 \cdot P_m} \right) - \delta_2 = \sin^{-1} \left( \frac{P_m x_1}{2 E_q' E_2} \right) \]

Since \( E_q', E_2, x_1, \delta_2 \) and \( P_m \) are known, the value of capacitive reactance can be determined from the above mathematical relation.

2. Lowest Bound on the Capacitor Size.

As before \( \delta_2 \) is selected as the maximum operating angle in the post-fault state. Instead of being dead-beat the rotor is allowed to swing up to angle \( \alpha < \pi - \delta_2 \), see figure 9.5. The operating point moves from 1 through 2, 3, 4 to 5. At this stage the capacitor is taken out. The operation continues along curve 2 from 6 to 7 to 8 such that area 607 = area 789. The rotor oscillates around position 7 and finally settles down at 7 corresponding to angle \( \delta_2 \).

The value of \( x_c \) which corresponds to the case of the rotor swinging upto an angle \( \alpha \) is calculated on the next page.
As before, with reference to figure 9.5, using the equal area criterion.

\[ \int_{\delta_1}^{\delta_2} (P_m - P_m^i \sin \delta) \, d\delta = \int_{\delta_2}^{\alpha} (P_m^i \sin \delta - P_m) \, d\delta \]

or \[ P_m (\delta_2' - \delta_1') + P_m^i (\cos \delta_2' - \cos \delta_1') \]

\[ = P_m^i (\cos \delta_2' - \cos \alpha) - P_m (\alpha - \delta_2') \]

\[ P_m^i (\alpha - \delta_1') = P_m^i (\cos \delta_1' - \cos \alpha) \]

\[ E'_q E_2 \]

\[ x'_2 = \frac{E'_q E_2 \cos \delta_1' - \cos \alpha}{P_m (\alpha - \delta_1')} = x_1 - x_c \]

Therefore

\[ x_c = x_1 - \frac{E'_q E_2 \cos \delta_1' - \cos \alpha}{P_m (\alpha - \delta_1')} \]

If in the limit \( \alpha \) is allowed to approach a value of \( \pi - \delta_2' \), then the value of \( x_c \) in the above equation is the lowest bound on the size of the capacitor for system stability.
Application to an Actual System

Figure 9.6 represents a single machine connected to an infinite bus through parallel transmission lines. This system is studied with respect to the influence of switching in a capacitor in series with the transmission line. The numbers on the diagram indicate the values of the reactances in per unit. The breakers adjacent to a fault on both sides are arranged to clear simultaneously.

Resistance and shunt capacitance are neglected.

The following data are provided:

i. voltage behind transient reactance = 1.25 pu

ii. voltage of infinite bus = 1.0 pu

iii. power transferred = 1.0 pu.

The transient starts due to a 3 phase fault at P which is cleared by opening of the circuit breakers adjacent to P.

Positive sequence impedance diagram for the system is as shown in fig. 9.7.

Amplitudes of the power angle curves are: (See figure 6.2)

(A) Before fault = \( \frac{1.0 \times 1.25}{.72} = 1.735 \)

(B) During fault = \( \frac{1.0 \times 1.25}{2.98} = 0.420 \)

(C) After the fault = \( \frac{1.0 \times 1.25}{1.00} = 1.25 \)

therefore

\[ \frac{P_{m2}}{P_{m1}} = \frac{.42}{1.735} = .242 \]

\[ \frac{P_{m3}}{P_{m1}} = \frac{1.25}{1.735} = .720 \]
Fig. 9.5 Determination of Minimum Capacitive Reactance

Fig. 9.6 Single Machine System
\[ \delta_1 = \sin^{-1} \frac{1.0}{1.735} = 35.2^\circ = 0.615 \text{ Radians} \]

\[ \pi - \delta_3 = \sin^{-1} \frac{1.0}{1.25} = 126.9^\circ = 2.22 \text{ Radians} \]

\[
\delta_2(\text{critical}) = \frac{\cos^{-1} \left( \frac{1.0}{1.735} \right) (2.22 - 0.615) + 0.72 \cos 126.9 - 0.242 \cos 35.2}{0.72 - 0.242}
\]

\[ = 51.6^\circ \]

Since 3 cycle breakers (0.05 sec. clearing time) are commonly used, it would be interesting to find out if the power level can be raised above 1.0 p.u. Initial angles for various power levels are as follows:

<table>
<thead>
<tr>
<th>P</th>
<th>( \delta_1^\circ )</th>
<th>( (\pi - \delta_3)^\circ = \delta_m^\circ )</th>
<th>Clearing angle ( \delta_2^\circ ) (front standard curves)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1.0</td>
<td>35.2</td>
<td>127.0</td>
<td>38.6</td>
</tr>
<tr>
<td>1.05</td>
<td>37.2</td>
<td>122.8</td>
<td>40.7</td>
</tr>
<tr>
<td>1.10</td>
<td>39.3</td>
<td>118.0</td>
<td>43.9</td>
</tr>
<tr>
<td>1.15</td>
<td>41.4</td>
<td>113.0</td>
<td>45.3</td>
</tr>
<tr>
<td>1.20</td>
<td>43.7</td>
<td>106.5</td>
<td>47.7</td>
</tr>
</tbody>
</table>
Figure 9.7 Positive Sequence Networks
Knowing the clearing angle corresponding to the minimum clearing time of 3 cycles, it is necessary to draw swing curves to check whether power level can be raised from 1.0 pu. to 1.20 pu. or an increase of 20%.

The data for the swing curves as obtained from digital calculations are given and swing curves plotted in figure 9.8.

<table>
<thead>
<tr>
<th>Fault cleared in 3 cycles</th>
<th>Power</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>a-1.0</td>
</tr>
<tr>
<td>Time-sec</td>
<td>δ°</td>
</tr>
<tr>
<td>0.00</td>
<td>35.2</td>
</tr>
<tr>
<td>0.05</td>
<td>38.6</td>
</tr>
<tr>
<td>0.10</td>
<td>46.3</td>
</tr>
<tr>
<td>0.15</td>
<td>54.8</td>
</tr>
<tr>
<td>0.20</td>
<td>63.2</td>
</tr>
<tr>
<td>0.25</td>
<td>70.6</td>
</tr>
<tr>
<td>0.30</td>
<td>76.3</td>
</tr>
<tr>
<td>0.35</td>
<td>80.1</td>
</tr>
<tr>
<td>0.40</td>
<td>81.8</td>
</tr>
<tr>
<td>0.45</td>
<td>81.4</td>
</tr>
<tr>
<td>0.50</td>
<td>78.8</td>
</tr>
<tr>
<td>0.55</td>
<td>74.2</td>
</tr>
<tr>
<td>0.60</td>
<td>67.8</td>
</tr>
<tr>
<td>0.65</td>
<td>60.0</td>
</tr>
<tr>
<td>0.70</td>
<td>51.4</td>
</tr>
<tr>
<td>0.75</td>
<td>43.1</td>
</tr>
</tbody>
</table>
It is observed from the swing curves that cases d and e, corresponding to power transmitted of 1.15 and 1.20 respectively are unstable. So it has been possible to increase power by 10% by employing the fastest possible fault clearing.

Consider the effect of adding a suitable capacitor in series as soon as the line is tripped out. From the data provided, the amplitude of power angle curve during the post-fault condition is 1.25. Temporarily it is necessary to slightly increase this value to create stability for power levels of 1.15 and 1.20. Assume that the amplitude of the power angle curve during the post-fault condition is \( x \), and if \( P_m \) is the input power then from equal area criterion

\[
x = \frac{P_m (\delta_m - \delta_1) + .42 (\cos \delta_2 - \cos \delta_1)}{\cos \delta_2 - \cos \delta_m}
\]

The value of \( x \) as calculated from this equation is:

\[
P_m = 1.15 \text{ pu} \quad x = 1.30 \text{ pu}
\]

\[
P_m = 1.20 \text{ pu} \quad x = 1.36 \text{ pu}
\]

or

\[
P_m = 1.15 \text{ pu}, \quad \frac{E_d^* E_2}{E_a x - x_c} = 1.30 = \frac{1.25 \times 1.0}{1.0 - \frac{x_c}{x_c}}
\]

\[
x_c = 1.0 - \frac{1.25}{1.30} = 1.0 - .962 = .038 \text{ pu}
\]

\[
P_m = 1.20 \text{ pu}
\]

\[
x_c = 1.0 - \frac{1.25}{1.36} = 1.0 - .883 = .117 \text{ pu}
\]
The swing curves are again calculated after including the series capacitors in the circuit and the following results are obtained.

\[
\text{Power} = \begin{array}{cc}
1.15 (x_c = .038) & 1.20 (x_c = .117) \\
\hline
\text{Time-sec} & \delta^0 & \delta^0 \\
0.00 & 41.4 & 43.7 \\
0.05 & 45.3 & 47.8 \\
0.10 & 54.2 & 57.1 \\
0.15 & 63.9 & 66.9 \\
0.20 & 73.6 & 76.3 \\
0.25 & 82.3 & 84.5 \\
0.30 & 89.8 & 91.4 \\
0.35 & 95.9 & 96.8 \\
0.40 & 100.8 & 100.9 \\
0.45 & 104.5 & 103.8 \\
0.50 & 107.3 & 105.6 \\
0.55 & 109.2 & 106.4 \\
0.60 & 110.4 & 106.3 \\
0.65 & 111.0 & 105.3 \\
0.70 & 111.1 & 103.2 \\
0.75 & 110.6 & 100.1 \\
\end{array}
\]

The swing curves are replotted on the same graph as broken lines from which the stability achieved by using capacitors can be observed.
Figure 9.8 Effect of Switched Series Capacitor on Swing Curves

- Fault Cleared in 3 Cycles
- e': same as d but capacitor inserted
- e: same as e but capacitor inserted

0 - electrical degrees
0.1 0.2 0.3 0.4 0.5 0.6 0.7 0.8
time - secs.

120 100 80 60 40 20
Chapter X

SWITCHED SERIES CAPACITORS (EXPERIMENTAL STUDY)

Introduction

This chapter describes some experimental work, on Switched Series Capacitors, which was performed in the Power Systems Laboratory of The University of Michigan. The purpose of the experiment was to make a comparison between the digital solution of the machine equations and the experimental results. It will be seen that there was fairly good agreement between the computer and the experimental results.

Brief Description of the Experiment

The experiment was performed on a synchrounous Generator/D.C. Motor set. The D.C. Motor which was a compound wound machine acted as the prime mover for the generator. The generator was synchronized to the A.C. Supply system through two parallel transmission lines, one consisting of reactors to simulate a long transmission system. The other a simple direct connection essentially zero reactance. Under steady state conditions, power was flowing through both lines. The direct connection was opened suddenly which created a transient and the generator began to oscillate. The oscillations were measured by a specially designed measuring apparatus. Different sizes of capacitors were introduced in the system and the oscillations were recorded. It was observed that for a certain value of the capacitive reactance, the oscillations were minimal. The experimental results were
compared with the digital solution of the machine equations.

**Machine Specifications**

<table>
<thead>
<tr>
<th>Component</th>
<th>Specification</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Synchronous Generator</strong></td>
<td>Make Westinghouse</td>
</tr>
<tr>
<td>KVA</td>
<td>11</td>
</tr>
<tr>
<td>Volts</td>
<td>240</td>
</tr>
<tr>
<td>Amps</td>
<td>26.4</td>
</tr>
<tr>
<td>Phases</td>
<td>3</td>
</tr>
<tr>
<td>Power factor</td>
<td>0.9</td>
</tr>
<tr>
<td><strong>Excitation volts</strong></td>
<td>125</td>
</tr>
<tr>
<td><strong>Excitation amps</strong></td>
<td>5.6</td>
</tr>
<tr>
<td><strong>R. P. M.</strong></td>
<td>1200</td>
</tr>
<tr>
<td><strong>Cycles</strong></td>
<td>60</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Component</th>
<th>Specification</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>D. C. Motor</strong></td>
<td>Make Westinghouse</td>
</tr>
<tr>
<td>Compound Wound</td>
<td></td>
</tr>
<tr>
<td>HP</td>
<td>15</td>
</tr>
<tr>
<td>Volts</td>
<td>250</td>
</tr>
<tr>
<td>Amps</td>
<td>52</td>
</tr>
<tr>
<td>RPM</td>
<td>1200</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>Component</th>
<th>Specification</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>Auxiliary Generator</strong></td>
<td>Make Dayton Electric</td>
</tr>
<tr>
<td>Volts</td>
<td>115</td>
</tr>
<tr>
<td>Amps</td>
<td>3.7</td>
</tr>
<tr>
<td>Power Factor</td>
<td>1.0</td>
</tr>
<tr>
<td>RPM</td>
<td>3600</td>
</tr>
<tr>
<td>VA</td>
<td>1000</td>
</tr>
<tr>
<td><strong>Phase</strong></td>
<td>1</td>
</tr>
<tr>
<td><strong>Cycles</strong></td>
<td>60</td>
</tr>
<tr>
<td><strong>RPM</strong></td>
<td>3600</td>
</tr>
<tr>
<td><strong>VA</strong></td>
<td>1000</td>
</tr>
</tbody>
</table>
Operation of the System

Figure 10.1 is a schematic diagram of the system which was set up in the laboratory. The synchronous generator was connected to the infinite bus, which was the Detroit Edison supply in this case, through two parallel transmission lines. As shown in the schematic, line A was a direct connection but line B was made of lumped reactors to simulate a long transmission line. Line A had a contact 1 in series with it, which was controlled by coil CR1. Since CR1 was energized, this contact was closed; hence power was flowing from the synchronous generator to the infinite bus through line A. The coil CR1 had a push button 'a' in series with it which was normally closed. It was pushed by hand to de-energize the coil and open contact 1. Line A was thus switched out and power began to flow through line B. Since contact 3 was open, power was flowing through path CC' which was a series circuit comprising inductance and capacitance. The net reactance of path CC' was capacitive. It was necessary to keep a combination of inductance and capacitance in CC' since otherwise the value of capacitance required would have been very large.

Since coil CR1 was de-energized, it closed contact 1', which energized the timer. After a pre-set time delay, coil CR2 was energized which closed contact 2'. Closing of this contact energized coil CR3. As soon as this coil was energized, contact 3 was closed, shorting out the capacitor circuit.
Fig. 10.1 Schematic Diagram of the Laboratory System

D.C. Motor

Syn. Gen.

Line Impedances

NEMA-Type 1 Welding Timer

Adjustable Phase Shifter

Sanborn Recorder

Variac

Single-Phase Generator

Auto transformers
c
capacitors c'

1 1 1
After a certain time when oscillations had disappeared and steady state conditions realized, push button 'a' was released which energized coil CR1, closing contact 1, thus reinserting line A and returning to the initial conditions.

The capacitor insertion time was varied by varying the time setting on the timer. The above procedure was repeated for different loads, capacitors and the capacitor insertion times.

Figures 10.2 through 10.5 are the photographs of the system as set up in the laboratory.

Torque Angle Measurement

The principle of the measurement of torque angle can be illustrated by reference to figure 10.6. OB and OA are two phasors representing two voltage sources which are assumed to remain constant in magnitude. OB is the voltage of the remote generating station and OA is the voltage of the receiving station. If $\delta$ is the torque angle, then $\overline{AB}$ is the phase difference between the two voltages.

If $\overline{OB}$ is equal in length to $\overline{OA}$, then

$$\overline{AB} = 2 \overline{OB} \sin \frac{\delta}{2} = K \sin \frac{\delta}{2}$$

where $K$ is a constant or

$$\delta = -2 \sin^{-1} \frac{\overline{AB}}{K} \quad (10.1)$$
Fig. 10.2

Fig. 10.3
KEY TO FIGURES 10.2, 10.3, 10.4, 10.5

1. Sanborn Recorder
2. D.C. Driving Motor
3. Synchronous Generator
4. Artificial Transmission Line
5. Switched Series Capacitors
6. Single Phase Reference Voltage Generator
7. Gear and Belt coupling
8. Control Contactor
9. Infinite Bus
10. Welding Timer
11. D.C. Motor Starter
12. Synchronous Generator Field Regulator
13. Phase Shifter
14. Variac
Fig. 10.6  Principle of Torque Angle Measurement
If the phase difference $AB$ vs. time is recorded during a transient, by means of a recorder, it is possible to draw a torque angle vs. time curve by the equation 10.1.

The above principle was used to measure the torque angle during the transient. The voltage of an auxiliary generator was used as a reference signal. The auxiliary generator was coupled to the main shaft by gear and toothed belt drive as shown in figure 10.3, which was necessary since the auxiliary generator had a rated speed of 3600 rpm while the main shaft was running at 1200 rpm.

Since there was no load on this generator, its voltage remained constant during a transient, but its phase changed with the position of the rotor. This was the voltage $OB$ in figure 10.6. Voltage signal $OA$ was obtained from the Edison supply, through an auto transformer as shown in figure 10.1. The voltage signal $OB$ was adjusted by means of a variac such that $OA = OB$. A phase shifter was used to bring the two voltages in phase with each other when no power was transmitted across the line, so that $\delta$ was zero under no load condition.

Signal $AB$ was connected across a Sanborn Recorder. As soon as the transient occurred, the fluctuations in the magnitude of this signal were recorded by this recorder. The equation 10.1 was then used to draw a torque angle/time curve from the recorder tracing.
Experimental Results

During the experiment, a number of machine loadings, capacitor sizes and switching times were used. A typical case was selected for analysis and comparison with the digital computer results. The data obtained from the Sanborn Recorder were converted to torque angles by equation 10.1. Experimental torque angle curves were then superimposed on the theoretical curves as obtained from digital computer analysis, the details of which are given in the following sections.

Theoretical Analysis

A theoretical study of the system was made by digital computer simulation and results were obtained as follows:

The digital computer investigation was to find the time solution of the swing equation

\[ M \frac{d^2 \delta}{dt^2} = P_i - P_o \]

where \( P_i \) and \( P_o \) are power input and power output of the machine.

To find these quantities we proceed as follows:

Machine and System Parameters

Direct Axis Synchronous Reactance = 3.8 ohms per phase
Quadrature Axis Synchronous Reactance = 2.46 ohms per phase
Open Circuit Time Constant = 0.4 second
Direct Axis Transient Reactance = 0.89 ohm per phase
Line Reactance = \( 4 \times 0.39 = 1.56 \) ohms
Additional Reactance = 0.48 ohm
Per Unit System

The following calculations are done on 11 KVA and 240 volt base.

Per unit voltage = \( \frac{240}{\sqrt{3}} = 138 \) V

Per unit current = \( \frac{11000}{\sqrt{3} \times 240} = 26.5 \) amp.

Per unit impedance = \( \frac{138}{26.5} = 5.236 \) Ω.

On the above per unit system, the values of the parameters of the system are as follows:

\[ x_d = 0.725 \]
\[ x_q = 0.47 \]
\[ x'_d = 0.17 \]
\[ x_E (\text{Line Reactance}) = 0.300 \]
Additional Reactance = 0.092

Figure 10.7 shows the reactances in per unit.

Determination of Inertia Constant

The inertia constant of the machine can be found from the following formula.

\[ H = 0.231 \times W R^2 \times (\text{rpm})^2 \times \frac{10^{-6}}{\text{KVA}} \]

where \( WR^2 \) is moment of inertia in lb-ft\(^2\), \( H \) is the inertia constant in kw-sec/KVA and rpm is the speed of the machine in revolutions per minute.

The moment of inertia of the machine rotor and the coupled DC rotor was calculated experimentally and found equal to 32.9 lb-ft\(^2\).
Therefore
\[ H = \frac{0.231 \times 32.9 \times (1200)^2 \times 10^{-6}}{11} \]
\[ = 0.995 \text{ kw/sec/KVA}. \]

**Transient Stability Analysis**

The machine was operated at full load and at unity power factor and suddenly line 1 was opened by pushing the button a. Figure 10.8 shows the phasor diagram of the system prior to the transient. OA is the voltage of the infinite bus as well as the terminal voltage of the machine, since line 1 is closed. OA is also the direction of the current since the power factor is unity.

\[ v_m = v_b = 1.00 \ \begin{array}{c} 0 \\ 0 \end{array} \]

where \( v_m \) is the machine terminal voltage and \( v_b \) is the bus voltage.

\[ i = 1.00 \angle 0^0 \quad \varnothing = \text{power factor angle} = 0 \]

\[ \vec{E}_{qd} = \vec{v} + j \times q \vec{i} \]
\[ = 1.00 \angle 0 + \angle 90^0 \times 0.47 \times 1.00 \ \begin{array}{c} 0 \\ 0 \end{array} \]
\[ = 1.00 \angle 0 + .047 \angle 90^0 \]
\[ = 1.00 + j 0.47 = 1.105 \ \begin{array}{c} 25^0 \\ 25^0 \end{array} \]

\[ i_d = 1.00 \sin \delta = 1.00 \sin 25 = 0.423 \]

\[ E'_q = 1.105 - (x' \times d') i_d \]
\[ = 1.105 - (.47 - .17) 0.423 \]
\[ = 0.9781 \]
Fig. 10.7 Schematic Diagram of the System

Fig. 10.8 Phasor Diagram
\[ E_q = E_{qd} + i_d (x_d - x_q) \]

\[
= 1.105 + 0.423(.725 - .47) = 1.213
\]

A transient stability program was written and the swing curve was calculated by the help of this program. Figure 10.9 shows both the theoretical and the experimental results as explained below.

1. Experimental swing curve without capacitor insertion.
2. Theoretical swing curve without capacitor insertion.
3. Experimental swing curve with .1 capacitor inserted for 10 cycles.
4. Theoretical swing curve with .1 capacitor inserted for 10 cycles.

It will be observed from the figure that there is a reasonable amount of agreement between experimental and theoretical data.
Fig. 10.9 Comparison of Theoretical and Experimental Curves
without capacitor

with capacitor insertion

Fig. 10.10 Experimental Recordings
Chapter XI

SWITCHED SERIES CAPACITORS (FAULT STUDY AND RELAYING)

Introduction

In Chapter IX, an approximate analysis of switched series capacitors was presented. Chapter X related to the experimental demonstration of this technique. In this chapter some problems which will arise in the application of switched capacitors to practical systems will be investigated. These problems include the effect of a second fault while the capacitor is in circuit and the effect of the capacitor insertion on distance relaying.

System Description

Refer to figure 11.1 which is rather a generalized circuit although similar to the one set up in the laboratory. There are three lines in parallel which connect the synchronous machine to the infinite bus. Circuit no. 1 has negligible reactance as was the case in the laboratory set up, in order to get a good transient. Circuits no. 2 and 3, each consist of two sections. Each section has a reactance of 0.16 p.u. There are two Y-Δ transformers at the two ends of the line, which have solidly grounded neutrals. A capacitor of 0.1 p.u. reactance, which has a circuit breaker in parallel with it, is located at the center of the line, as shown in the figure. Each section has two circuit breakers at its ends which have distance relays on them.

It is admitted that since circuit no. 1 has been assigned a reactance of zero, this cannot be called a practical system.
Fig. 11.1 A Generalized Circuit

Fig. 11.2 A Typical By-pass Circuit
It was, however, selected in order to have as much similarity to the laboratory set up as possible.

The reactances of the generator are shown in the figure. The total line reactance is the same as in the experiment.

Since the parameters of the system are the same as the system in the laboratory had, the initial conditions for per unit power transfer are also the same.

Initially all these lines are in operation. Due to a fault line no. 1 is relayed out and the capacitor is inserted into the circuit. While the system is oscillating, there is another fault which is removed by switching out the corresponding line section.

The initial steady state conditions are the following. These are shown in figure 10.8.

\[ I = 1.00 \quad 0^\circ \]
\[ E'_q = 0.98 \]
\[ E_q = 1.21 \]
\[ v_b = v_m = 1.00 \quad 0^\circ \]
\[ \delta = 25^\circ \]

**Position and Type of Fault**

In figure 11.1, F₁, F₂, F₃ and F₄ are four different locations at which the fault will be assumed in order to get an overall picture of its effect. A three phase fault will be considered in the investigation since this results in the worst condition from the stability standpoint.
The following table shows the fault current through the capacitor for different locations.

<table>
<thead>
<tr>
<th>Fault Location</th>
<th>Current through the Capacitor</th>
</tr>
</thead>
<tbody>
<tr>
<td>$F_1$</td>
<td>7.70 pu</td>
</tr>
<tr>
<td>$F_2$</td>
<td>20.00</td>
</tr>
<tr>
<td>$F_3$</td>
<td>4.45</td>
</tr>
<tr>
<td>$F_4$</td>
<td>3.27</td>
</tr>
</tbody>
</table>

From the above table, it is obvious that as soon as a fault occurs at any of the locations $F_1$ through $F_4$, the capacitor will be bypassed due to excessive voltage across it. The fault will result in switching out of a line section by action of the associated relays and the breaker, which will take about 3 cycles. The capacitor will be fully reinserted in the transmission circuit within one half cycle after the overvoltage condition is removed.

**Capacitor Protection**

Should a second fault occur while the capacitor is in circuit, the current through the capacitor will be many times normal. It would, therefore, be necessary to bypass the capacitor to protect it against overvoltage. It is important that the capacitor bypass circuit be opened immediately after the capacitor voltage is reduced below the critical value.
A typical protective bypass circuit is shown in figure 11.2. This has been designed for use with fixed series capacitors and can equally well be utilized in the case of switched series capacitors. The breakdown voltage of the bypass circuit can be adjusted in the range of 2.5 to 3.0 times the rated voltage of the capacitor. To protect against moderate overvoltages, additional relay equipment has to be provided. Flow of current in the gap circuit causes operation of the electrically operated air valves and the air blast reservoir discharges a stream of air into the gap conducting area. After establishment of sufficient air flow, the arc in the gap space is quenched at each current zero and gap flashover occurs again on the next half-cycle of high voltage. Thus when the cause of the capacitor overvoltage is removed the gap arc is quenched at the first current zero thereafter and no restrike occurs. The capacitor is fully reinserted in the transmission circuit within one-half cycle after the overvoltage condition is removed.

Analysis of the Effect of Fault on Power Transfer

Power output from the synchronous generator to the infinite bus is given by the following equation

\[ P_u = \frac{E'_q v_b}{x'_d + x_e} \sin \delta + \frac{v_b^2 (x'_d - x_q)}{2(x'_q + x_e)(x'_d + x_e)} \sin 2\delta \]

\[ E'_q = .98 \]
\[ v_b = 1.00 \]
\[ \delta_o = 25^0 \]
\[ x'_d = .17 \]
\[ x_q = .47 \]

The transient proceeds as follows:

Initially all three lines were in operation and therefore \( x_e \), the external reactance was equal to zero (note this is the same condition as in the laboratory). Therefore

\[
P_u = \frac{.98 \times 1.0}{.17} \sin 25^0 - \frac{1 (.30)}{2 \times .47 \times .17} \sin 50^0 = 1.00 \text{ pu}
\]

As soon as a fault occurs on line no. 1, it is switched out and the capacitor is switched in. Therefore

\[ x_e = 0.07 + 0.08 - 0.10 + 0.08 + 0.07 = .20 \text{ pu} \]

Assume that while the capacitor is in the circuit, a second fault occurs at any of the locations \( F_1 \) through \( F_4 \), then the power output during the period of the fault is reduced to zero. The faulted section will be relayed out in 3 cycles.

For half a cycle, following the relaying out of the faulted section, the capacitor remains bypassed. Therefore the value of \( x_e \) during this half cycle = \( .07 + .16 + .08 + .07 = .38 \). Next the capacitor is inserted and therefore \( x_e = .38 - .1 = .28 \). The machine continues to swing until the oscillations are damped out.

The capacitor is inserted when \( \frac{d\delta}{dt} \) is positive and removed when it is negative. This technique requires a sophisticated control
equipment to switch the capacitor in and out of the circuit but this is a superior method of using a capacitor as compared to the method used in the laboratory where the capacitor was switched out after a certain number of cycles.

**Computer Study**

A computer program was written to simulate the above system. The value of $H$ used in the study was $.995$ as obtained in the experimental investigation.

The program makes the following simulation: The generator is operating in the steady-state supplying a power equal to $1.00$. A fault occurs which results in switching out of line no. $1$ and insertion of the capacitor. A second fault occurs immediately which drops the output to zero for a period of $3$ cycles. At the end of $3$ cycles, the fault is removed by switching out a line section. Since the capacitor will be out for another half cycle, the value of $x_e$ during the period immediately following the clearance of the fault $= .07 + .08 + .16 + .07 = .38$. Another half cycle afterwards, the capacitor is reinserted into the circuit and the value of $x_e = .38 - .1 = .28$. The capacitor remains in circuit when $\frac{d\delta}{dt}$ is positive and is out of circuit when $\frac{d\delta}{dt}$ is negative. The enclosed swing curve shows that the system is stable. It may be pointed out that the second fault was assumed to occur at the very start of the first swing which makes it rather a special case. In actual practice, the second fault may occur any time during the swing which fact should be considered when designing real systems.
\[ X_c = 1 \text{ p.u} \]

Capacitor inserted when machines are swinging apart, and removed when these are swinging towards each other.

Fig. 11.3 Swing Curve with Capacitor Insertion
Effect of Switched Series Capacitors on Distance Relaying

Refer to figure 11.4(a) which shows a fault at the location F_1. Since the capacitive reactance is subtracted from the total inductive reactance of the lines 1 and 2, the fault appears much closer to relay A which may cause operation of this relay. Thus instead of opening the faulty section 3, the unfaulted section 1 may be switched out.

In practice, due to the action of the bypass circuit, the capacitor will generally remain shorted out during a fault. In most cases, therefore, the operation of the distance relays will not be affected by capacitor insertion during fault conditions. An exception will be a "line-to-ground" fault where the fault current may not be high enough to short out the capacitor. In such a case the relaying will have to be properly designed to take care of the effect of the switched capacitor.

A capacitor insertion can have substantial effect on distance relaying during machine swinging. It would therefore be desirable to study the effect of capacitor insertion on the impedance "seen by the distance relay" during a swing.

Consider figure 11.4(a) again which can be simplified as shown in figure 11.4(b), if the parallel direct connection is omitted.

The following analysis is based on the assumption that the voltage behind the transient reactance remains constant, otherwise the calculations will become unduly complicated.
Fig. 11.4(a) Circuit for Distance Relay Study

Fig. 11.4(b) A Simplified Circuit
To investigate the problem, the following discussion will be helpful.  

Consider figure 11.5, which shows a simple two machine system. Here $V_A$ and $V_B$ are the voltages behind the transient reactances which are assumed constant in magnitude but varying in phase during swings or out-of-step conditions. $V_A$ leads $V_B$ by the variable angle $\delta$. The current anywhere in the series circuit is

$$I = \frac{V_A \delta - V_B}{Z}$$

where $Z$ is the impedance of the connecting circuit including the transient reactance of the two machines. The total impedance $Z$ is divided by the relay location M into two parts, $mZ$ and $(1-m)Z$ where $m$ is a real number less than 1. At point M the voltage $V$ is then

$$\overline{V} = (1-m) \frac{V_A \delta}{Z} + mV_B$$

The impedance "seen" by the relays at M is

$$\overline{Z}_r = \frac{\overline{V}}{I} = \frac{(1-m) \frac{V_A \delta}{Z} + mV_B}{\frac{V_A \delta}{Z} - \frac{V_B}{Z}} = \frac{(1-m) \frac{V_A \delta}{Z} + mV_B}{\frac{V_A \delta}{Z} - \frac{V_B}{Z}}$$

$$\overline{Z}$$
Fig. 11.5 A Simple Two Machine System
\[
\frac{Z_r}{Z} = \frac{(1-m) \frac{V_A}{V_B}}{\sqrt{\delta} - 1} = \frac{k(1-m)\sqrt{\delta} + m}{k \sqrt{\delta} - 1} = \frac{k(1-m) + m \sqrt{\delta}}{k - \sqrt{\delta}}
\]

where \( k \) is the magnitude of \( \frac{V_A}{V_B} \).

In the system under study \( k = 1.01 \) and if it can be taken as equal to 1, the above equation simplifies to

\[
\frac{Z_r}{Z} = \frac{(1-m) + m \sqrt{-\delta}}{1 - \sqrt{-\delta}} = \frac{-m(1 - \sqrt{-\delta}) + 1}{1 - \sqrt{-\delta}}
\]

\[
= -m + \frac{1}{1 - \sqrt{-\delta}} = \left( \frac{1}{2} - m \right) - j \frac{1}{2} \cot \frac{\delta}{2} \quad (1)
\]

This equation represents a vertical line, because the real part is constant while the imaginary part varies as a function of \( \delta \). If both sides are multiplied by \( Z \), lengths are multiplied by the magnitude of \( Z \) and the line is rotated counterclockwise through the impedance angle \( \theta \).

In the post-fault steady-state condition the capacitor will be out of the circuit and equation (1) will hold. However when the capacitor is inserted the value of \( m \) will change. The new value of \( m \) can be found as follows:
Let \( m = m_1 \) when the capacitor is inserted

\[
\text{Total Impedance} = Z - X_c
\]

Then \( m \cdot Z = m_1 Z_1 \)

\[
= m_1 (Z - X_c)
\]

or \( m_1 = \frac{Z}{Z - X_c} \cdot m \)

Thus the value of \( m_1 \) increases by an amount \( \frac{Z}{Z - X_c} \).

Applying the above analysis to the system of Figure 11.5(a) the following impedance is "seen" by the relay at location \( A \), without capacitor insertion.

\[
\frac{Z_r}{Z} = \left[ \left( \frac{1}{2} - m \right) - j \frac{1}{2} \cot \frac{\delta}{2} \right]
\]

\[
m = \frac{.17 + .07}{.17 + .07 + .08 + .08 + .07} = .24 = .51
\]

Thus the locus of \( \frac{Z_r}{Z} \) is a vertical line at a distance of \( \frac{1}{2} - .51 = -.01 \) from the origin. The ordinates of points on this line representing various values of \( \delta \) are given by

\[
y = -\frac{1}{2} \cot \frac{\delta}{2}.
\]

Table 1 shown calculation of \( y \) for 30° increments, from 0° to 180°.
Since the resistance of the line is neglected, \( Z = .47 \| 90^\circ \).

Hence the vertical line is turned counterclockwise by \( 90^\circ \) and since only one half current flows in the protected line, the loci is multiplied by \( .47 \times 2 = .94 \). The loci is shown in Figure 11.16.

As soon as the capacitor is inserted, the value of \( m \) becomes \( m_1 = \)

\[
m \cdot \frac{Z}{Z - X_c} = .51 \times \frac{.47}{.47 - .1} = .51 \times \frac{.47}{.37} = .646
\]

and \( Z_1 = .47 - .1 = .37 \). Therefore the loci jumps from A to B.

Assume that the first zone impedance element of the relay at breaker 1 reaches to 80% of the length of the line which it protects or to \( .80 \times .16 = .128 \). Let the second zone be set at .256 and the third at .384. The tripping characteristics of these elements are concentric circles of radii .128, .256 and .384 with centers at the origin. Let the directional
element have maximum torque at a phase angle of 45°. Its characteristic is a straight line through the origin and perpendicular to the line of maximum torque. The characteristic is shown in Figure 11.6.

From the figure it is observed that although insertion of the capacitor adversely affects the performance characteristics of distance relay no. 1, for the assumed conditions of the problem, it creates no serious difficulty as can be seen from the following discussion.

It has been noted earlier that the value of δ varies from 25° to 90°. The initial locus takes it into the zone Z₂. The new locus keeps it in the same zone. Therefore under conditions of "capacitor in" or "capacitor out" the worst condition is the same. The diagram shows that instantaneous tripping of this particular relay will not occur during swinging or out-of-step operation, but that delayed tripping will occur if the angle δ lies close enough to 90° for a sufficient time to close the contacts.

The above study demonstrates that while planning systems involving switched series capacitors care must be exercised to ensure that the protective relaying is not adversely affected.
Fig. 11.6 Impedance seen by Relay during swings.
Chapter XII

CONCLUSIONS

Three methods of improving transient stability of a synchronous generator have been investigated.

The first method involves the use of signals proportional to the derivatives of the torque angle $\delta$. It has been demonstrated that the transient stability of a synchronous generator is substantially improved if the derivative signals of proper magnitudes are used in the excitation system. A mathematical method has been demonstrated which develops formulas for amplification factors in various feedback channels. The values of line constants of the excitation system can be substituted in appropriate formulas to determine the magnitudes of these signals. Formulas have also been derived which takes saturation into account.

The second method involves the use of Phase-shifting transformers. A mathematical analysis of the effect of these transformers, on the transient stability, has been made. It has been demonstrated that this method should have considerable effect on improving the transient stability.

The third method uses switched series capacitors. By experiment and by theory the feasibility of this method has been demonstrated. Many technical problems associated with the use of the series capacitors have been thoroughly investigated.
In conclusion, it appears that there is considerable scope for improving the system stability during transient conditions by all these methods. While switched series capacitors and phase-shifting transformers are strong measures, due to economy these may have limited applications. There will, however, be situations where these will be the most economical techniques of controlling transient instability. On the other hand, excitation control continuously contributes to damping. With the development of static excitation systems, which have low time constants and much higher voltage limits, there is great potential for improving transient stability by using derivative signals.
REFERENCES


Appendix I

Mathematical Representation of a Synchronous Generator

The analysis to be presented here employs the concept of an ideal synchronous machine and is based on the classical work of Park.

An ideal synchronous machine has no saturation, hysteresis or eddy currents and all of its fields are sinusoidally distributed. The assumption of sinusoidal field distribution removes the harmonics of the airgap mmf and flux waves, from consideration. Under most operating conditions, these harmonics have a secondary effect on machine behavior, hence this assumption is justified.

The neglect of magnetic saturation is a serious approximation, particularly in steady state stability study. Hence the technique by which magnetic saturation can be taken into account will be discussed.

Magnetic hysteresis is negligible for materials used in the construction of modern machines and will be ignored here.

In the discussion to follow, a salient pole rotor is analyzed, rather than a nonsalient pole. The reason for this is that a hydroset is usually a slow speed set and hence has a salient pole rotor. Since
this research deals with the analysis of a single machine connected with
a large system by means of a long transmission system, it is more
representative of a remote hydro-station connected to a load center by
means of a transmission line.

It would be appropriate here to briefly describe the construction
of a typical synchronous machine. The salient-pole rotor has two axes
of symmetry. One, called the direct axis, is identical to the polar
axis, the other is coincident with the interpolar axis and is called the
quadrature axis. By definition, the positive quadrature axis is taken
ahead of the positive direct axis in the direction of rotation.

The stator has three, distributed phase windings, the positive
axes of which are equally spaced about the stator periphery at intervals
of $2\pi/3$ electrical radians. The three phases are lettered abc to indicate
the order in which their positive axes are encountered in skirting the
stator periphery in the direction of rotor rotation. A schematic
diagram of an idealized, two-pole synchronous machine is given in
figure A1-1. The angle $\theta$ is the electrical angle by which the rotor
reference axis is advanced on the stator reference axis in the direction
of rotation.

To develop the equations of motion for an idealized, three-phase,
salient pole synchronous alternator, linear system of coupled networks
will be used which is the usual method of treatment for such a study.
Saturation will be taken into account later on by means of certain
factors known as saturation factors. A set of inductance coefficients will be defined and the principal of superposition used in determining the total flux linkages of each of the several machine windings.

Double subscript notation will be used to identify the inductance coefficients, like subscripts designate a self-inductance coefficient, while unlike subscripts indicate the mutual inductance between the windings indicated by the subscripts.

Figure A-1.1 Two Pole Synchronous Machine
Fig. A-1.2 Y-Connected Synchronous Machine Windings
The induced voltage in a coil is calculated as

\[ v(t) = \frac{d \psi(t)}{dt} \]  

(1)

where \( \psi(t) \) is the coil's flux linkage. If current flows in the windings,

\[ v_a(t) = r_i(a(t) + \frac{d \psi_a(t)}{dt} \]

\[ v_b(t) = r_i(b(t) + \frac{d \psi_b(t)}{dt} \]

\[ v_c(t) = r_i(c(t) + \frac{d \psi_c(t)}{dt} \]

\[ v_f(t) = r_i f(t) + \frac{d \psi_f(t)}{dt} \]

(2)

Since saturation is not as yet taken into account, flux linkages can be assumed to be proportional to mmf's and superposition is applicable. Dropping \( t \) for convenience,

\[ \psi_a = L_{aa} i_a + L_{ab} i_b + L_{ac} i_c + L_{af} i_f \]

\[ \psi_b = L_{ba} i_a + L_{bb} i_b + L_{bc} i_c + L_{bf} i_f \]

\[ \psi_c = L_{ca} i_a + L_{cb} i_b + L_{cc} i_c + L_{cf} i_f \]

\[ \psi_f = L_{fa} i_a + L_{fb} i_b + L_{fc} i_c + L_{ff} i_f \]

(3)

\( L_{aa}, L_{bb}, L_{cc} \) are self inductances, \( L_{ab}, L_{ac}, \) etc. are mutual inductances.
$L_{aa}$ will have a maximum value when the direct axis is aligned with the axis of coil $a$ and $L_{ab}$ will have a maximum value when the rotor has moved $30^\circ$ clockwise from the center line of coil $a$. $L_{aa}$ has a minimum value when the quadrature axis is aligned with the axis of coil $a$. The variation of $L_{aa}$ between its extreme values is sinusoidal and can be expressed as

$$L_{aa} = L_s + L_m \cos 2\theta$$

where $\theta$ is measured from the center line of coil $a$.

$$L_{bb} = L_s + L_m \cos 2(\theta - 120^\circ)$$

$$= L_s + L_m \cos(2\theta + 120^\circ)$$

$$L_{cc} = L_s + L_m \cos(2\theta - 120^\circ)$$

The expressions for mutual inductances are

$$L_{ab} = L_{ba} = -[M_s + L_m \cos 2(\theta + 30^\circ)]$$

$$= -M_s + L_m \cos(2\theta - 120^\circ)$$

$$L_{bc} = L_{cb} = -M_s + L_m \cos 2\theta$$

$$L_{ca} = L_{ac} = -M_s + L_m \cos(2\theta + 120^\circ)$$

$$L_{af} = L_{fa} = M_f \cos \theta$$

$$L_{bf} = L_{fb} = M_f \cos(\theta - 120^\circ)$$

$$L_{cf} = L_{fc} = M_f \cos(\theta + 120^\circ)$$
Since the rotor mmf is driving flux across a constant air gap, $L_{ff}$ is a constant.

These expressions for the various inductances can be substituted in equation (3) to calculate flux linkages in terms of currents. The constant velocity of the rotor is taken into account by writing $\theta(t) = \omega t + \theta_o$.

The equation for $v_a$ is

$$v_a = [r - 2\omega L_m \sin(2(\omega t + \theta_o))] i_a - [2\omega L_m \sin(2\omega t + 2\theta_o - 120^\circ)] i_b$$

$$+ [2\omega L_m \sin(2\omega t + 2\theta_o + 120^\circ)] i_c - [\omega M_f \sin(\omega t + \theta_o)] i_f$$

$$+ [L_s + L_m \cos(2\omega t + \theta_o)] \frac{di_a}{dt} + [M_s + L_m \cos(2\omega t + 2\theta_o - 120^\circ)] \frac{di_b}{dt}$$

$$+ [M_s + L_m \cos(2\omega t + 2\theta_o + 120^\circ)] \frac{di_c}{dt} + [M_f \cos(\omega t + \theta_o)] \frac{di_f}{dt}$$

Similar differential equations are obtained for $v_b$, $v_c$ and $v_f$.

It is obviously very difficult to solve these equations.

**d-q Transformations**

It will simplify the calculations if the flux linkages due to the stator coils are resolved along the direct and quadrature axes of the machine. See Figure A1.1. Using the following Park's transformation

$$\psi_d = \frac{2}{3} [\psi_a \cos\theta + \psi_b \sin(\theta - 120^\circ) + \psi_c \cos(\theta + 120^\circ)]$$

$$\psi_q = -\frac{2}{3} [\psi_a \sin\theta + \psi_b \sin(\theta - 120^\circ) + \psi_c \sin(\theta + 120^\circ)]$$

$$\psi_o = \frac{1}{3} (\psi_a + \psi_b + \psi_c)$$

or
\[ \psi_a = \psi_d \cos \theta - \psi_q \sin \theta + \psi_o \]

\[ \psi_b = \psi_d \cos(\theta - 120^\circ) - \psi_q \sin(\theta - 120^\circ) + \psi_o \]

\[ \psi_c = \psi_d \cos(\theta + 120^\circ) - \psi_q \sin(\theta + 120^\circ) + \psi_o \quad (5) \]

Similarly

\[ v_a = v_d \cos \theta - v_q \sin \theta + v_o \]

\[ v_b = v_d \cos(\theta - 120^\circ) - v_q \sin(\theta - 120^\circ) + v_o \]

\[ v_c = v_d \cos(\theta + 120^\circ) - v_q \sin(\theta + 120^\circ) + v_o \]

\[ i_a = i_d \cos \theta - i_q \sin \theta + i_o \]

\[ i_b = i_d \cos(\theta - 120^\circ) - i_q \sin(\theta - 120^\circ) + i_o \]

\[ i_c = i_d \cos(\theta + 120^\circ) - i_q \sin(\theta + 120^\circ) + i_o \]

Substituting expressions for \( \psi_a, \psi_b, \) and \( \psi_c \) in (4), the result is

\[ \begin{align*}
\psi_d &= (L_s + M_s + \frac{3}{2} L_m) i_d + M_f i_f \\
\psi_q &= (L_s + M_s - \frac{3}{2} L_m) i_q \\
\psi_o &= (L_s - 2M_s) i_o \\
\psi_f &= \frac{3}{2} M_f i_d + L_{ff} i_f
\end{align*} \quad (6) \]

Let \( L_d = L_s + M_s + \frac{3}{2} L_m \). This may be called the direct axis synchronous inductance.

\[ L_q = L_s + M_s - \frac{3}{2} L_m \]

This is called the quadrature axis synchronous inductance.
\[ L_0 = L_s - 2M_s, \]

this is called the zero sequence inductance. Then

\[ \psi_d = L_d i_d + M_f i_f \]

\[ \psi_q = L_q i_q \]

\[ \psi_o = L_o i_o \]

\[ \psi_f = \frac{3}{2} M_f i_d + L_{ff} i_f \]

Transforming the voltage equations to the d-q reference frame.

\[ v_d = \frac{2}{3} \left[ v_a \cos \theta + v_b \cos(\theta - 120^\circ) + v_c \cos(\theta + 120^\circ) \right] \]

\[ = \frac{2}{3} \left[ (ri_a + \frac{d \psi_a}{dt}) \cos \theta + (ri_b + \frac{d \psi_b}{dt}) \cos(\theta - 120^\circ) + (ri_c + \frac{d \psi_c}{dt}) \cos(\theta + 120^\circ) \right] \]

\[ = \frac{2}{3} r[i_a \cos \theta + i_b \cos(\theta - 120^\circ) + i_c \cos(\theta + 120^\circ)] \]

\[ + \frac{2}{3} \left[ \frac{d \psi_a}{dt} \cos \theta + \frac{d \psi_b}{dt} \cos(\theta - 120^\circ) + \frac{d \psi_c}{dt} \cos(\theta + 120^\circ) \right] \]

\[ v_d = ri_d + \frac{d \psi_d}{dt} - \omega \psi_q \quad (8) \]

Similarly

\[ v_q = ri_q + \frac{d \psi_q}{dt} + \omega \psi_d \quad (9) \]

\[ v_o = ri_o + \frac{d \psi_o}{dt}. \quad (10) \]

Note that \(- \omega \psi_q\) and \(\omega \psi_d\) terms are large as compared to \(ri\) and \(\frac{d \psi}{dt}\) terms and correspond to the voltages induced as the air gap flux wave sweeps past the stationary stator coils.
The same equations can represent a generator if the sign of the currents is changed, which will give the following set of equations

\[ v_d = -ri_d + \frac{d}{dt} \psi_d - \omega \psi_q \]  
(12)

\[ v_q = -ri_q + \frac{d}{dt} \psi_q + \omega \psi_d \]  
(13)

\[ v_f = r_f i_f + \frac{d}{dt} \psi_f \]  
(14)

\[ \psi_d = -L_d i_d + M_f i_f \]  
(15)

\[ \psi_f = -\frac{3}{2} M_f i_d + L'_{ff} i_f \]  
(16)

\[ \psi_q = -L_q i_q \]  
(17)

For a balanced system,

\[ v_d = v \sin \delta \]  
(18)

\[ v_q = v \cos \delta \]  
(19)

where \( \delta \) is the angle between the voltage \( v \) and the quadrature axis.

Solving equations (15)-(17) for currents and substituting in (12)-(14)

\[ i_d = \frac{M_f}{L'_{ff} L'_{d}} \psi_f - \frac{\psi_d}{L'_{d}} \]  
(20)

\[ i_f = \frac{L_d}{L'_{ff} L'_{d}} \psi_f - \frac{3}{2} \frac{M_f}{L'_{ff} L'_{d}} \psi_d \]

where

\[ L'_{d} = L_d - \frac{3}{2} \frac{M_f^2}{L'_{ff}} \]

is called the direct axis transient inductance.
\[ v_d = \frac{d \psi_d}{dt} - \frac{r M_f}{L_f' L_d'} \psi_f + \frac{r \psi_d}{L_d'} - \omega \psi_q \]  
(21)

\[ v_q = \frac{d \psi_q}{dt} + \frac{r}{L_q} \psi_q + \omega \psi_d \]  
(22)

\[ v_f = \frac{d \psi_f}{dt} + \frac{\gamma_f}{L_f'} \frac{L_d}{L_d'} \psi_f - \frac{3}{2} \frac{r_f}{L_f' L_d'} M_f \psi_d \]  
(23)

If the derivatives of \( \psi_d \) and \( \psi_q \) are neglected, since these are small compared to the \( \omega \psi \) terms, this reduces (21) - (23) to

\[ v_d = -\omega \psi_q - r_i_d \]

\[ v_q = \omega \psi_d - r_i_q \]

\[ \frac{d \psi_f}{dt} = v_f - \frac{r_f}{L_f'} \frac{L_d}{L_d'} \psi_f + \frac{3}{2} \frac{r_f}{L_f' L_d'} M_f \psi_d \]

Let \( \frac{L_f'}{r_f} = T_{do}' \) = open circuit field time constant, then

\[ \frac{d \psi_f}{dt} = v_f - \frac{L_d}{L_d' T_{do}} \psi_f + \frac{3}{2} \frac{M_f}{L_d T_{do}} \psi_d \]

Let \( \omega M_f i_f = E_q \)

Multiply equation (14) by \( \frac{\omega M_f}{L_f'} \)

\[ \frac{\omega M_f}{L_f'} v_f = \frac{\omega M_f}{L_f'} r_f i_f + \frac{d}{dt} \left( \frac{\omega M_f}{L_f'} \psi_f \right) \]

Let \( E_q' = \frac{\omega M_f}{L_f'} \psi_f \).
\[
\frac{\omega M_f}{L_{ff}} v_f = \frac{\omega M_f}{L_{ff}} \frac{r_f}{r_f} v_f = \frac{E_{ex}}{T_{do}}, \text{ where}
\]

\(E_{ex}\) is the open circuit armature voltage.

Hence
\[
\frac{d}{dt} E_q' = \frac{1}{T_{do}} (E_{ex} - E_q)
\]

where \(E_q = \bar{v} + r\bar{I} + j\omega L_d \bar{I}_d + j \omega L_q \bar{I}_q\)

from equation (20)
\[
i_d = \frac{M_f}{L_{ff}} \psi_f - \frac{\psi_d}{L_d'}
\]

or \(\omega L' i_d = \frac{\omega M_f}{L_{ff}} \psi_f - \psi d\)

or \(\omega L' i_d = E_q' - v_q - ri_q\)

Hence \(E_q' = v_q + \omega L' i_d + ri_q\) \((24)\)

Since \(v_q = -ri_q - \omega L_d i_d + E_q\)

Substituting in (24) gives
\[
E_q' = -ri_q - \omega (L_d - L_d') i_d + E_q + ri_q
\]

But \(E_q = E_{qd} + \omega (L_d - L_q) i_d\)

Therefore \(E_q' = E_{qd} - \omega (L_q - L_d') i_d\)
or in phasor form

\[ \bar{E}'_{q} = \bar{E}_{qd} - e^{j\omega(t)} \left( L_q - L_d' \right) \bar{i}_d \]

If stator resistance is neglected, the equations can be summarized as follows:

\[ \frac{d \bar{E}'_{q}}{dt} = \frac{1}{T_{do}} \left( E_{ex} - \bar{E}_{q} \right) \]  

(25)

\[ E_q = E_{qd} + \omega(L_d - L_q) \bar{i}_d \]  

(26)

\[ \bar{E}_{qd} = \bar{v} + j\omega L_q \bar{i} \]  

(27)

\[ E_{q}' = E_{qd} - \omega(L_q - L_d') \bar{i}_d \]  

(28)

\[ \bar{E}_{q} = \bar{v} + j\omega L_d \bar{i}_d + j\omega L_q \bar{i}_q \]  

(29)

The phasor diagram corresponding to above equations is shown on the next page. In the diagram, \( v_b \) is the busbar voltage, \( v_m \) the machine terminal voltage and \( x_e \) the external reactance.
Fig. A-1. 3 Phasor diagram
Symbols Used- Program nos 1, 2 and 3

G = Machine Rating
H = Inertia Constant
DLTT = Time Interval
XD = Direct Axis Synchronous Reactance
XQ = Quadrature Axis Synchronous Reactance
XDP = Direct Axis Transient Reactance
XE = External Reactance
TDO = Open Circuit Time Constant
TE = Exciter Time Constant
VFO = Initial Value of Excitation Voltage
VR = Reference Voltage
MUV = Gain in Voltage Channel $\mu_v$
AMUD = $A_2$ = Gain in $\frac{d\delta}{dt}$ Channel
AMU2D = $A_1$ = Gain in $\frac{d^2\delta}{dt^2}$ Channel
VB = Busbar Voltage
XC = Input Power in the First Interval
YC = Input Power in the Second and Subsequent Intervals
EEXO = Initial Value of Exciter Voltage Referred to Armature Circuit
EQPO = Initial Value of Voltage Proportional to Field Flux Linkage
PU = Output Power

158
DF = \frac{d\delta}{dt}

DS = \frac{d^2\delta}{dt^2}

D(N) = \delta_n

V = Machine Terminal Voltage

F = Frequency
Program No. 3

\[ C = A \text{ Constant} = M = \frac{GH}{180f} \]

DLTT = Time Interval \( \delta T \)

PAI = Accelerating Power in the First Interval

DI = Initial Value of \( \delta \)

PI = Input Power

THETA = \( \theta \) the Angle by which the Power Angle Curve is Shifted
Computer Program No. 1

Chapter V: Excitation Systems (Digital Computer Results)

Title: To find transient Stability of a System using auxiliary signals in the excitation system.

```
DIMENSION D(50)
NAMELIST/NMEA1/F,G,H,DLTT,XD,XQ,XDP,XE,TD0,TE,
1FV0,VR,MUV,AMUD,AMU2D,VB,XC,YC,EEXO,EQPO
PU=1.0
READ(5,NAME1)
D(1)=50.2
PI=XG
EEX=EEXO
EQP=EQPO
T=0.
DF=0.
N=1
1 T=T+DLTT
 N=N+1
 DS=(PI-PU)*180.*F/(G*H)
 DDF=DS*DLTT
 DF=DDF+DF
 DD=DF*DLTT
 D(N)=D(N-1)+DD
 WRITE(6,3)T,D(N),EEX
3 FORMAT(2HT=,F6.3,2X,2HD=,F6.3,2X,2HEEEX=,F6.3)
 X=D(N)*3.14159/180.
 AA=VB*VB*(XQ-XDP)*SIN(X)*COS(X)/((XQ+XE)*(XDP+XE))
 BB=(XDP-XQ)/(XQ+XE)
 EQ=(EQP+BB*COS(X))/(1.+BB)
 VD=XQ*VB*SIN(X)/((XQ+XE)
 CC=(XD-XDP)/(XQ-XDP)
 DD=(XD-XQ)/(XQ-XDP)
 VF=CC*EQ-DD*EQP
 VQ=(XD*VB*COS(X)+XE*VF)/(XD+XE)
 V=SQRT(VD*VD+VQ*VQ)
 WRITE(6,10)V
10 FORMAT(2HT=,F6.3)
 EX=F V0-MUV*(V-VR)+AMUD*DF+AMU2D*DS
 IF(EX+1.)6,6,7
6 EX=-1.
 GO TO 4
```

continued
7 IF(Ex-3.5)4,5,5
5 Ex=3.5
4 CONTINUE
   AK1=DLTT*(Ex-ExEx)/Te
   AK2=DLTT*(Ex-ExEx+.5*AK1)/Te
   AK3=DLTT*(Ex-ExEx+.5*AK2)/Te
   AK4=DLTT*(Ex-ExEx+AK3)/Te
   Exx=Exx+(AK1+AK2+AK2+AK3+AK3+AK4)/6.
   DNN=(Exx-VF)*DLTT
   DNNM=(VF/EQP)*DLTT/2.
   DEQP=DNN/(TDQ+DNNM)
   EQP=EQP+DEQP
   PU=(EQP*VB*SIN(X)/(XDP+XE))-AA
   PI=YC
2 GO TO 1
R CONTINUE
END
Computer Program No. 2

Chapter VIII: Application of Phase Shifting Transformers.

Title: To compute data for a Swing Curve for a system using Phase Shifting Transformers.

```
1 DIMENSION D(25), DLTD(25), PO(25)
2 NAMELIST/NAME1/C, DLTD, PAI, DI, PI, DDI, THETA
3 READ(5, NAME1)
4 D(1)=DI
5 PO(1)=PAI
6 DLTD(1)=0.
7 T=0.
8 N=1
9,1 DDIO=DDI
9 T=T+DLTT
10 N=N+1
11 DLTD(N)=DLTD(N-1)+PO(N-1)*DLTT*DLTT/C
11,1 DLTD(N)=DLTD(N)+DDIO
12 D(N)=D(N-1)+DLTD(N)
13 WRITE(6,3)T, D(N)
14 3 FORMAT(2HT=, F6.3, 5X, 2HD=, F10.5)
15 X=((D(N)+THETA)*3.1416/180.)
16 PO(N)=PI-1.25*SIN(X)
16,1 DDIO=0.
17 IF(N-24)1,1,9
18 9 CONTINUE
19 END
# END OF FILE
```
Computer Program No. 3

Chapter X: Switched Series Capacitors (Experimental Study)

Title: To determine Transient Stability of a system using switched series capacitors. The capacitors are taken out after a fixed number of cycles.

```
DIMENSION D(30)
NAMELIST/NAME1/F,G,H,DLTT,XD,XQ,XDP,XEI,TDQ,TE,
1VF0,VR,MUV,AMUD,AMU2D,VB,XC,YC,EEX3,EQP0,D10,TF
PU=1.
READ(5,NAME1)
D(1)=D10
PI=YC
EEX=EEX3
EOP=EQP0
V=VR
T=0.
DF=0.
N=1
XE=XEI-XC
1 T=T+DLTT
N=N+1
DS=(PI-PU)*180.0/F/(6*H)
DDF=DS*DLTT
DF=DDF+DF
DD=DF*DLTT
D(N)=D(N-1)+DD
WRITE(6,3)T,D(N),V
3 FORMAT(2HT=,F6.3,2X,2H=,F6.1,2X,2H=,F6.2)
X=D(N)*3.14159/180.
AA=VB*VB*(XQ-XDP)*SIN(X)*COS(X)/((XQ+XE)*(XDP+XE))
BB=(XDP-XQ)/(XQ+XE)
EG=(EOP+BB*COS(X))/C(1.+BB)
VD=XQ*VB*SIN(X)/(XQ+XE)
CC=(XD-XDP)/(XQ-XDP)
DD=(XD-XQ)/(XQ-XDP)
VF=CC*EO-DD*EOP
VO=(XD*VB*COS(X)+XE*VF)/(XD+XE)
V=SQRT(VD*VD+VO*VO)
EX=VF0-MUV*(V-VR)+AMUD*DF+AMU2D*DS
IF(EX+1.)6,6,7
6 EX=-1.
G0 T0 4
7 IF(EX-3.5)4,5,5
5 EX=3.5
```

continued
6 EX=-1.
   G0 T0 4
7 IF (EX-3.5) A, B, C
5 EX=3.5
4 CONTINUE
   AK1=DLTT*(EX-EE)/(TE
   AK2=DLTT*(EX-EE+.5*AK1)/TE
   AK3=DLTT*(EX-EE+.5*AK2)/TE
   AK4=DLTT*(EX-EE+AK3)/TE
   EEE=EEE+(AK1+AK2+AK2+AK3+AK3+AK4)/6.
   DNM=(EEE-VF)*DLTT
   DNM=(VF/EQP)*DLTT/2.
   DEQP=DNM/(TD3+DNM)
   EQP=EQP+DEQP
   PU=(EQP*VB*SIN(X)/(XDP+XE))-AA
   IF (DF) A1, B1, C1
11 XE=XE1
   G0 T0 12
2 XE=XE2
12 IF (N-30) A, B, C
13 G0 T0 1
8 CONTINUE
END
Computer Program No. 4

Chapter XI: Switched Series Capacitors (Fault Study and Relaying)

Title: To determine Transient Stability of a system using switched series capacitors. The capacitors are taken out when the torque angle is decreasing and inserted when the torque angle is increasing.

```
DIMENSION D(30)
NAMELIST/NAME1/F,G,H,DLTT,XD,XQ,XDP,XEI,TDO,TE,
1VF0,VR,MUV,AMUD,AMU2D,VB,XC,YC,EEX0,EQ0,Q,XE2,DI0
PU=1.
READ(5,NAME1)
D(1)=DI0
PI=YC
EEX=EEX0
EQP=EQ0
V=VR
T=0.
DF=0.
N=1
XE=XEI-XC
1 T=T+DLTT
N=N+1
DS=(PI-PU)*180.*F/(G*H)
DDF=DS*DLTT
DF=DDF+DF
DD=DF*DLTT
D(N)=D(N-1)+DD
WRITE(6,3)T,D(N),V
3 FORMAT(2HT=,F6.3,2X,2HD=,F6.1,2X,2HV=,F6.2)
X=D(N)*3.14159/180.*
AA=VB*VB*(XQ-XDP)*SIN(X)*COS(X)/((XQ+XE)*(XDP+XE))
BB=(XDP-XQ)/(XQ+XE)
EQ=(EQP+BB*COS(X))/(1.+BB)
VD=XQ*VB*SIN(X)/(XQ+XE)
CC=(XD-XDP)/(XQ-XDP)
DD=(XD-XQ)/(XQ-XDP)
VF=CC*EQ-DD*EQP
VQ=(XD*VB*COS(X)+XE*VF)/(XD+XE)
V=SORT(VD*VD+VQ*VQ)
EX=VF0-MUV*(V-VR)+AMUD*DF+AMU2D*DS
IF(EX+1.)6,6,7
```

continued
4 CONTINUE
AK1=DLTT*(EX-EEX)/TE
AK2=DLTT*(EX-EEX+5*AK1)/TE
AK3=DLTT*(EX-EEX+5*AK2)/TE
AK4=DLTT*(EX-EEX+AK3)/TE
EEX=EEX+(AK1+AK2+AK2+AK3+AK3+AK4)/6
DNM=(EEX-VF)*DLTT
DNMM=(VF/EQP)*DLTT/2
DEQP=DNM/(TD3+DNMM)
EOP=EQP+DEQP
PU=(EQP*V3*SIN(X)/(XDP+XE)) / AA
IF(T-TF)2, 11, 11
11 XE=XE1
   G3 T0 12
2 XE=XE1-XC
12 IF(N=30)13, R, R
13 G3 T0 1
8 CONTINUE
END
TRANSIENT STABILITY STUDY

FLOW CHART I FOR THE COMPUTER Programs
Transient Stability Study
using phase shift transformers
in the system

Flow Chart II for Computer Programs